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There is no debating that the BASIC Stamp® I is a great little controller. Unfortunately, it has to sacrifice speed and performance to bring you the convenience of BASIC. Now, Micromint offers Stamp users the best of all worlds: assembly language speed, BASIC and C programmability, and enhanced performance.

Micromint introduces the three little PicStics. PicStics are BASIC Stamp | I I/O-compatible 14-pin boards

with a little extra. The primary enhancement is speed. Any program you have for a Stamp will automatically run faster on a PicStic using the PBASIC compiler. Besides getting at least a 15-times increase in execution speed and typically double the instruction space of the Stamp I, you also increase your code's security.

PicStics" also have more I/O lines and onboard features. The PicStic" 2 has a battery-backed hardware clock/calendar. Most people just call it a real-time clock because it allows you to trigger program events based on a particular time of day or day of the month. PicStic" 3 connects you to the analog world through its high-speed, 2-channel, 12-bit A/D converter.

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actual size 0.6" × 1.4"

	Micromint PicStic™ 1	Micromint PicStic™ 2	Micromint PicStic™ 3	BASIC Stamp® I
Price	\$29	\$39	\$39	\$34
BS1 software compatible	yes	yes	yes	yes
Reprogrammable EEPROM	yes	yes	yes	yes
I/O lines	10	9	10	8
Extra features	none	real-time clock	2-channel, 12-bit ADC	none
Programming language	PBASIC compiler, C, assembly lang.	PBASIC compiler, C, assembly lang.	PBASIC compiler, C, assembly lang.	PBASIC interpreter
RAM (variables)	22 bytes	22 bytes	22 bytes	14 bytes
Max. program length	150 inst*	150 inst*	150 inst*	80 inst*
Clock speed	4 MHz	4 MHz	4 MHz	4 MHz
Execution speed	50k inst/s*	50k inst/s*	50k inst/s*	2k inst/s
Max. serial I/O	2400 bps**	2400 bps**	2400 bps**	2400 bps
Packaging	14-pin SIP	14-pin SIP	14-pin SIP	14-pin SIP

*Maximum program length and execution speed are determined by instruction mix. Overall speed gain is typically at least 15 times while individual instructions may be 100 times faster.

PicStic[™] Development System includes:

PBASIC or C compiler, PicStic programmer, power supply, and cables.

Does not includes PicStic. \$149.00

Simply order PicStic by number: PS1, PS2, or PS3

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PicStic Most Asked Questions

1. Is PicStic BASIC Stamp compatible?

Using the PBASIC compiler, PicStics are 100% BASIC Stamp program compatible, but typically run 15 times faster (7 times faster than BS2) and allow more variables. If you have an existing BS1 program, it can be compiled using the PBASIC compiler and executed exactly as before. The only exception is that because PicStic uses a compiler, interactive features like debug are not compiled with the program.

2. I have a BASIC Stamp® development system. Gan I use it?

No. Parallax BASIC Stamps® use an interpreted BASIC with a tokenized BASIC downloader. PicStics use an EEPROM-programmable processor with a compiled PIC assembly language program. So, yes, you'll need a PicStic programmer with either the PicStic C or PBASIC compiler development system.

3. How fast is the PicStic ADC?

PicStic uses an LTC1298 12-bit ADC. In a typical application, the ADC can provide 1000 samples per second or more.

4. How much current can I draw from PicStic's onboard regulator?

For a 9-V input at 25°C, the 5-V output is typically 100 mA.

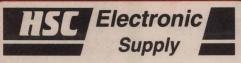
5. Why does PicStic have two more I/O lines and interrupts too?

PicStic has the same eight general-purpose I/O lines supported in PBASIC, but because PicStic uses an EEPROM PIC16C84, the two lines required for programming the Stamp aren't required on PicStic. These two I/O lines and the four interrupt sources are available using assembly language (or calls from PBASIC or C).

6. What about the 9600 bps or 10 MHz PicStics?

The best part about using PicStic and a compiler is that we can start by being 100% compatible and then add powerful new instructions as they are written. Next versions of the PBASIC compiler will allow 9600 bps, Peek and Poke, and increased clock speeds.

^{**}Comm speed is presently set for compatibility. New compiler versions will allow 9600 bps.



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Nuts & Volts Magazine encourages article submissions and queries. Send a SASE for a copy of our writer's guidelines.

All submissions should be on 5-1/4 or 3-1/2 inch diskettes and include hard copy as well. If return of materials is requested, include a SASE with your submission.

Deadlines should be discussed in advance with the editor, but generally all material should be submitted by the 1st of the month for the next month's issue.

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Dear Nuts & Volts:

I don't usually write to electronic magazines that I have followed since 1945, but someone should give Ronald Reis a pat on the back for such an excellent article on resistors "The Ubiquitous Resistor: All About the Most Widely Used Electronic Component of All Time" July '96.

I was Ham licensed in 1946, radio school USAAF 1944, and am now an extra, but it was a review that really impressed me.

Harry Tanzer Yucca Valley, CA

Dear Nuts & Volts:

The article entitled "The Ubiquitous Resistor" (July '96, p. 26) incorrectly attributes the invention of the Wheatstone bridge to Charles Wheatstone. In fact, the circuit was invented by Samuel Hunter Christie in 1833. Wheatstone merely encouraged and popularized its use in electrical measurements. Indeed, Wheatstone, in his own writings in the mid-ninteenth century, gave full credit to Christie for the circuit's invention.

> **Bill Johnston** White Sands, NM

Response:

Bill is right. Technically, Charles Wheatstone did not "invent" the bridge, which is named in his honor. He "developed" it in 1850.

Ronald A. Reis

Dear Nuts & Volts:

I am an electronic technology instructor at Suburban Technical School in Ronkonkoma, NY and I have found the articles to be a perfect introduction for my students into the hobbyist realm.

There are five instructors teaching five classes during the day. Each class runs 1,200 hours (about 10 months - throughout the year) and covers everything from DC and AC theory, semiconductors and op-amps, communications, digital, DOS, upgrading, and repair of PCs and networks. The four evening classes last 13 months and stress digital technol-

Although we try to allocate a third to a half of our time to laboratory applications and investigation of the lecture part of the course, we are always looking for real applications and fun projects. Your magazine fits the need and the students who have had a

chance to read my copy are motivated to try some of the designs or at least to enjoy the achievement of now being able to understand and to analyze the circuits. For most of these students, this is their first exposure to electronics; they were not brought up with a soldering iron in their hands and a burning desire to make electronic things.

Roger S. Cohen Holbrook NY

Dear Nuts & Volts:

In a recent Tech Forum column, I inadvertently called an EPROM and Electrically Erasable Read Only Memory (which, if you add Programmable, is the correct name for another device, the EEPROM). An EPROM is actually an Erasable Programmable Read Only Memory. The devices are functionally similar with both being field programmable and erasable; however, an EPROM can only be erased through exposure to ultraviolet light, whereas an EEPROM can be erased (a memory location at a time) while incircuit. Flash memories belong to this class; however, in this case, the entire device can be "flash" erased at once. Since I use both parts, the error was all too easy to overlook. My apologies for any confusion this may have caused. Aside from the name, the rest of the information was correct.

> Jay Rabkin Pasadena, CA

Dear Nuts & Volts:

As a subscriber, I want to thank you for posting Scott Edward's columns and info on your Web page. I hope that you see fit to expand this coverage to include other writers such as Reis, Weeder, Lancaster, and Lunt.

Rich Dunlap

Dear Nuts & Volts:

Haven't seen your magazine in years and when I get my issue I will be subscribing. Glad you have a Web site and find it very interesting. Your magazine was one of the best magazines I would read back when I was very interested in electronics. Now taken over by computers, your magazine is still going to be worth reading every month. Thanks for your hard work and taking the time to give us a Web site to visit.

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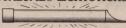
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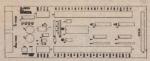


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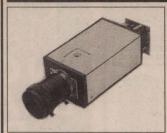
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PC Power Supply Repair











by TJ Byers

Pepairing a broken PC power supply is a lot simpler than you might think. Nine times out of ten you can do it yourself for under \$10.00.

It's 8:00 a.m., the neighbor's dog barked all night, your coffee tastes like weak tea, and the phone message light blinks frantically. Full of resolve, you flip on your PC's power switch, and ... presto — nothing! No lights, no beep, no fan, nada. Suddenly you realize, it's gonna be a really bad hair day.

While there's nothing I can do about the early hour or the coffee, I can probably help you get your PC back on its feet. The most common case of "Sudden PC Death Syndrome" is a defective power supply. The problem can come from many sources, like heat, power surges, and old age. While it's easy enough to replace a power supply by swapping the old for new, it's not always practical.

A case in point: I have an AST 486SX that died when a truck plowed into the corner power pole and caused a two-hour black out. When the power came back on, my PC didn't. A quick check showed the cause was a fried power supply. Unfortunately, a call to AST revealed, to my horror, that a replacement power supply costs \$150.00. Moreover, because of its unique case design, there's no generic substitute.

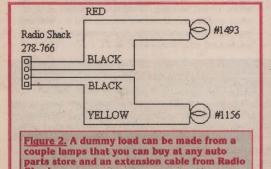
Fortunately, it's not difficult to fix PC power supplies. While they may look different on the outside, most PC power supplies use the same electronics on the inside. In this article, I'll show you how easy it is to fix a dead power supply.

The Basics

The power supply is a large metal box, mounted inside the PC that provides power to the mother-board and various peripherals. It's easily identified by a warning sticker on the case that reads "CAUTION! Hazardous Area" (or a similar high-voltage warning).

On the back of the power supply is an AC connector that plugs the PC into the wall. Often there's another AC connector that's used by some monitors. Most power supplies also have a voltage selector switch that lets it work with 110V or 220V power sources.

A typical PC power supply provides four DC output voltages: +5, +12, -5, and -12 volts. These voltages are available through four different types of connectors (Figure 1; 1-4). The color of the wire identi-



fies the voltage and its use (Table 1).

Getting Started

Alot of power supply failures are actually simple problems that are easy to fix. Obviously, the place to start is at the beginning – in other words, are you getting power from the wall to the PC? As stupid as it sounds, the first thing to do is look under your desk and see if the PC is plugged into the wall. If it is, move the plug to a different socket (they go bad, too, you know).

That done, pull the power cord from the back of your PC and see if the power is getting that far. You can do this using a VOM or a simple neon lamp circuit tester, like part number 22-102 from Radio Shack.

If there's no power, and you're plugged into a power strip or surge protector, the strip is probably the culprit. To test it, simply remove the PC's plug from the strip and

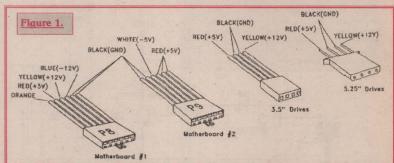
plug it into a wall socket. If the PC starts working, the problem is in the strip. Generally, the problem is a blown fuse or a tripped circuit breaker. You'll find both at the cord end of the strip. The last item you should test before popping the hood is the power cord itself; replacing it with another cord is the fastest and safest method.

Under The Hood

Still nothing? Now it's time to remove the cover.

Most covers are attached by five or six screws on the back. Before going any further, carefully read the instructions in the section called "Safety First."

The next logical place to look is at the power switch. Unfortunately, this may not be possible at this stage of the game. Many power supplies have a built-in power switch which isn't accessible until you dis-



The 5.25-inch drive connector is the easiest to access for testing. The motherboard connectors P8 and P9 are identical, and can be reversed. They plug into the motherboard with the black leads together on the inside.

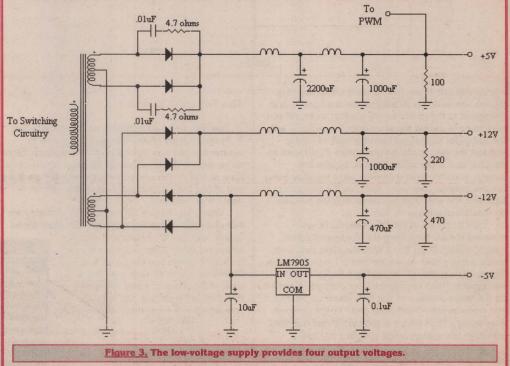
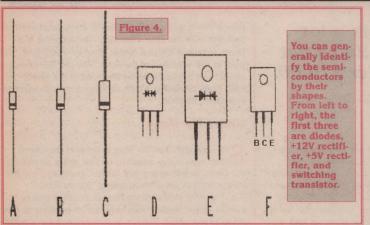


Table 1. Power Supply Color Codes Wire Color Voltage Use Motherboard, adapter cards, disk drives Red +5V Logic circuits (rarely used in modern PCs) Disk drive motors, RS-232 serial port, fans, -5V White Yellow +12V adapter cards Blue -12V RS-232 serial port, fans Orange n/a Power OK signal Black Ground (GND)



beginning with the floppies. Measure the +5- and +12-volt lines at each step. This will tell you whether or not the power problem is specific to a device. Don't forget to power off the system each time you disconnect a

device. With the hard disk(s) still connected, remove plugs P8 and P9 (Figure 1) from the motherboard.

Finally, it's time to deal with the unlikely possibility of a shorted hard disk. If you have more than one hard disk, start shedding them one at a time. When you're down to your last hard disk, unplug it and connect its power plug to the dummy load shown in Figure 2. (I don't recommend running a PC power supply without a load.) If the power supply is still dead, it's off to the drawing board.

If the supply was powered from the AC line within the last few minutes, the large electrolytic capacitors in the high-voltage section will most likely still have a charge in them that could give you a shocking awakening. If so, let the power supply rest for a while before you crack the case.

Each case has its own method of construction, but generally two sides of the enclosure are what protect the inside electronics. Remove the cover screws, taking care to watch out for attached leads, switches, and sharp edges. If you have to disconnect any leads (typically fan wires) or mechanical parts, note care-

fully how they go back together.

Give the electronics a good looking over, paying attention to any scorched or burned parts that may point to a failure. If you have a built-in power switch, now is the time to check it. Next, check the fuse. Is it blown? If in doubt, use the VOM to test for continuity (use the X100 range). If the fuse is blown, replace it with one of the same type and rating before going any further. It's possible the trouble is the result of metal fatique or mechanical failure of the fuse itself. To see if this solved the problem, connect the dummy load to one of the drive connectors and apply power.

If nothing happens, remove the dummy load and proceed to the resistance checks procedure. If the fuse blows with an explosion, go to the high-voltage repair section.

Resistance Checks

Referring to Table 2, perform a resistance measurement test. Keep the VOM's polarity correct, that is red to ground when testing a negative source, and wait for the filter capacitors to charge before taking a reading. The resistance values listed in Table 2 are only representative (the figures were gathered from actual measurements of several power supplies using a cheap VOM), so don't worry if your values are different from those listed.

However, if a resistance value is abnormally high or low, you have a problem. As a rule of thumb, a reading of 50 ohms or higher on the 5-volt and 12volt lines means the output is probably okay. A resistance value of 40 ohms or less indicates a short, generally in the rectifier diodes. The five-volt line is the most prone to failure because it carries the heaviest load (typically 20 amps). An extraordinarily high resistance reading indicates an open, probably a zapped board trace or a burned resistor. Both conditions are often harbingers of problems in the highvoltage section, but not necessarily. It depends on how fast the shutdown circuit reacted. But before we face that possibility, we first need to find the extent of the low-voltage damage.

Low-Voltage Repair

he low-voltage section of the power supply is a very simple rectifier, L-section filter design (Figure 3). Key to the success of this design is a multiple secondary power transformer. There is a 5-volt winding and a 12-volt winding. In high-power supplies (250 watts and larger), there are usually two five-volt windings that are paralleled for higher output current yet treated as a single winding.

High-Voltage Figure 5. The switcher section is the most common to fall. The power transistors have to have a breakdown voltage Common +1507 of 600 volts or more, and the damper dlodes have to be fast recovery (a IN4005 won't work) 1005 3901 Low-Voltage Outputs Switching OOluF Driver 390k To **PWM** -150V

mantle the unit. If you have a tower computer case, though, the switch is located on the front panel, and connected to the power supply via four wires. All you have to do is unplug the wires from the switch - with the computer unplugged from the wall, of course and test the switch with an ohmmeter. If you want to do a hot test of the switch (that is, bypass the switch), you can short the power wires together using two insulated jumper wires and plug the computer back into the wall. Just be careful that the jumpers don't touch anything.

Let's now look at the DC voltages. (If you removed the AC wires from the front-panel power switch, replace them first.) With the main switch off, locate a free power connector (the 5-1/4 inch version, Figure 1d, is preferred) or unplug a floppy drive to free up one. Don't unplug the hard disk; you'll need it for the entire duration of this test. Power up the PC, and measure the +5-volt (red) and +12-volt (yellow) lines using a VOM (black is ground). Make sure they fall within the voltage range specified in Table 2.

If they are out of range, power off the system and disconnect the mechanical drives one at a time.

The Drawing Board

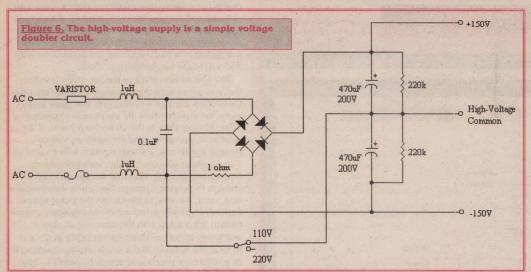
ow that we've done all that we can do with the power supply inside the cabinet, it's time to remove the unit and place it on the workbench. Since we've already disconnected all the power connectors, it's a simple matter of removing the mounting screws and sliding the power supply out of the cabinet, right? Well, hopefully.

Unless you have a tall tower, you'll probably run into obstacles, like adapter boards, disk drive signal

cables, and support brackets. If you're lucky enough to have a detailed user's manual, it shows you the procedure. Otherwise you're on your own. In either case, make notes of where everything is, how they're connected, and keep the screws with the items they came from.

WARNING: MAKE SURE THE PC IS DISCONNECTED FROM THE WALL **BEFORE STARTING DISASSEMBLY!**

Would you put a hairpin in an AC outlet socket? Not hardly! So why would you consider putting your finger in a power supply that is clearly labeled CAUTION!? Always unplug your PC before going under the hood. Once there, pay attention to my G! signs. I've done my best to make the troubleshooting processing as shock free as possible, but power has to be provided at various stages of the game. Be alert, don't be stupid, and if you don't know what to do next, stop now!



part, because you have to first locate the affected parts on the circuit board. Use the road map, "How To Find Waldo," to help you in your quest. An ohmmeter is a good way to probe suspected areas for shorted devices. Once the area is located, the real work begins because it's virtually impossible to tell the difference between a shorted diode and a shorted capacitor without removing one or the other. Since the rectifier is the most likely culprit and the easier to remove (the electrolytics are glued in place), I'd start there.

The +5-volt and +12-volt diodes are most likely nestled inside a transistor case mounted on a heat sink. The bigger one (Figure 4e) is the +5-volt rectifier, and the smaller one (Figure 4d) is the +12-volt rectifier. The negative-voltage rectifiers are individual diodes typically in a DO-41 case.

With the suspect rectifier or diodes in hand, do a resistance check of the defective voltage output line again. If the reading is within the normal range, trash the old part or parts and replace with new. (Helpful Hint: If the new diodes come in an axial-lead pack-

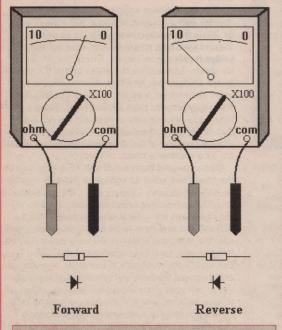


Figure 7. A cheap VOM is the best way to check translators and diodes. Why? Because the test voltage has to be enough to breach the barrier voltage of a silicon diode, typically 0.7 volts, and a lot of DVMs have a probe voltage of 0.3 volts and less.

Each winding has a grounded center tap to permit fullwave rectification using just two diodes (fullwave bridge rectifiers need four diodes). The direction of the rectifiers determines the polarity of the output voltage. Common cathodes are positive, and

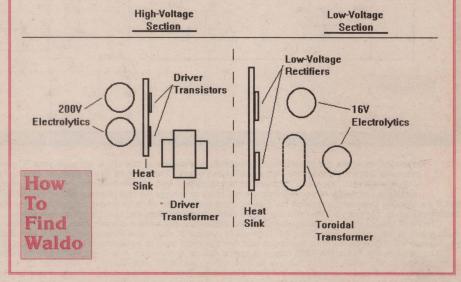
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Because of its highcurrent requirements, the +5-volt rectifier is usually an array of parallel Schottky diodes in a single package (Figure 4) that mounts on a heat sink. The -5-volt output is often derived from the -12-volt rectifier via an IC regulator (typically an LM7905

common anodes

are negative.



equivalent) rather than from the five-volt transformer winding. However, I've seen it done both ways.

The output of the rectifiers is filtered first by an inductor, called a choke, then by a heavy-duty electrolytic capacitor. In some designs, the five-volt line is double-filtered to reduce ripple by cascading two L-section filters on the output. Invariably, a bleeder resistor is placed the output to displace the consistence of the context to displace the

across the output to discharge the capacitors after power off.

The most common cause of low-voltage failure is a shorted rectifier. If one blows, so does its companion, which forces you to replace them as a package deal. Second on the hit list is a shorted capacitor, which usually does less overall damage. Most of the time, the failure is limited to just one output line, but there's no guarantee.

The first step is to locate the shorted components. For this operation you need access to the bottom side of the printed circuit board. This is the hard part, because no two supplies are alike. Use your

imagination, and be careful not to damage other components in the process. For example, twisting and turning the board too many times can cause attached wires to break loose.

Now comes the tricky

age, typically DO-41, solder them on the trace side of the circuit board instead of the component side. It's a lot easier.) If the output still shows a short, yank the electrolytic and check again. If the output is still shorted, make sure you're pulling the right teeth.

Exact replacement parts always cost more than generics, so go with the generic. You can get "universal" replacements from GE, RCA, and Philips ECG. (Infortunately, they're almost as expensive as the original. For the +5-volt rectifier, I recommend the MBR series from General Instruments and Motorola (available from Digi-Key and Allied Electronics, respectively). The +12-volt rectifier is a dual Schottky device that's available from several vendors, and generally sells for a buck or two. The negative voltage rectifiers must be fast recovery diodes, like a 1N4933. Replacement electrolytic capacitors are as close as your local Radio Shack.

When the voltage line has a three-terminal IC voltage regulator, check the resistance between both the input and the output (Figure 4) to ground. If only the output pin is shorted, the output capacitor is bad.

Table 2. Output Voltage and Resistance

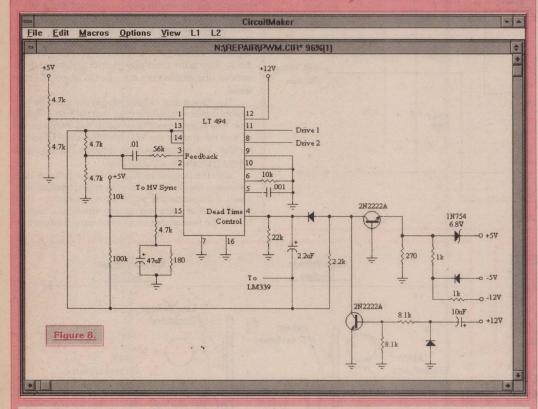
Nominal Voltage +5V -5V +12V -12V

Voltage Range +4.75V to +5.25V -4.75V to -5.25V +9V to +15V -9V to -15V 0V or +5V Resistance
>100 ohms
>100 ohms
>250 ohms
>250 ohms
1000 ohms

0 ohms

Wire Color Red White Yellow Blue Orange Black

Still Don't Work, Huh?



ine times out of ten, the troubleshooting techniques presented in this article will solve your PC power supply problems. But what if the power supply still doesn't work? There can be lots of reasons, ranging from a faulty transformer (good luck finding a replacement) to a bad solder connection. In most situations, I'd cut faulty transformer (good luck finding a replacement) to a bad solder connection. In most situations, I'd cut my losses and find a substitute power supply or try to salvage the motherboard for use in another system. But if you're really dead set on reviving the system, there is one more stage we haven't discussed — the PWM (pulsewidth modulator). But put your seat belt on, 'cause this is gonna be short and fast. It's not for everybody.

The PWM (Figure 8) is what drives the switching transistors, and when it doesn't work, nothing works. Consider it the brains of the power supply. The PWM is generally a single IC chip, most likely a Motorola TL 494. But before you replace the chip, let's see if it's working or not. For this you'll need an oscilloscope and a power supply.

The simplest way to test the PWM chip is to grab a disk drive connector and pump +12 volts into its yellow wire from an independent power supply. This can be done using another PC power supply or any other DC source (batteries work, too). Once power is applied to the PWM chip, observe the output waveforms on pins 8 and 11. Both outputs should be active squarewayes. If at first you don't succeed, ground pin 4 and try again, If the scope still shows

puts should be active squarewaves. If at first you don't succeed, ground pin 4 and try again. If the scope still shows nothing, replace the LT 494 chip. If the scope shows waveforms, the most likely culprit is the LM339 comparator. The LM339 is cheap, about a buck, and readily available, so it's worth a shot.

My method of replacing an IC is to clip the leads as close as possible to the body of the IC, leaving 14 or so metal pegs standing upright from the main board. Paying attention to direction, slip the replacement IC alongside the pegs and solder the new component in place.

If by now the power supply still doesn't work — chuck it.

If only the input pin is shorted, the rectifiers are bad. If both are shorted, the chance is both the diodes and the IC are shorted. To verify this theory, remove the IC and check the resistance again. If it reads okay, replace the semiconductors. The re-placement for the -5-volt IC voltage regulator is an LM7905.

High-Voltage Repair

If the new fuse blows when you apply power, there's a problem in the high-voltage section. We know this

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because the low-voltage section has an automatic shutdown circuit that reacts a lot faster than the fuse; that is, a low-voltage problem disables the power supply long before the fuse has time to blow. That doesn't necessarily mean the low-voltage outputs are okay, because failure of the -12-volt line can cause cascading damage that goes all the way back to the high-voltage section.

The high-voltage section is divided into two parts: the high-voltage power supply and the switching circuit. Most high-voltage failures occur in the switching circuit.

WARNING: COMPLETELY DISCHARGE THE INPUT CAPACITORS BEFORE WORKING IN THE HIGH-VOLTAGE SECTION!

If the fuse has a "mirrored" look to it, you can bet the farm that at least one of the two switching transistors is shorted (Figure 5). Typically they perish as a couple. These transistors are mounted on the heat sink(s) closest to the two largest electrolytic capacitors (see "How To Find Waldo"). With the red probe of a VOM on the collector of the first transistor, check the collector-to-emitter resistance, then the collector-to-base resistance (Figure 4). If a short is found, replace both the transistor and the damper diode that's across its emitter-collector. I normally use a Motorola MJE13009 for the power transistor and a 1N4937 for the damper diode.

You should also replace the low-value resistor that's in series with the transistor's base. This resistor is often used as a fusible link that goes puff when the switcher fails. Its purpose is to protect other components in the chain from harm. If the resistor is burned beyond recognition, you can replace it with any 1/4-watt resistor with a value of 1 to 10 ohms (the exact value isn't important). Sometimes, though, even the fusible isn't fast enough to prevent damage. So before installing the new parts, it's wise to check out the pulse shaper network (typically a resistor-diode-capacitor combination) associated with the base circuit, too. A quick way to test all three components at once is to treat the network like a single diode, checking it as a whole for shorts and opens (Figure 7). Now repeat the procedure for the second switching transistor.

The high-voltage supply is a simple voltage doubler with an output of about 300 volts (Figure 6). While this section rarely fails on its own, a shorted switching transistor can wipe out the bridge rectifier in an instant. Check the AC input for shorts, and replace the entire bridge if a short is found. Bridges can be either discrete diodes or a large, rectangular module, and you can find suitable replacements from Radio Shack. There's probably a one-ohm resister in line with the AC input that needs to be checked, too. On the outside chance that one of the doubler capacitors is shorted, do a resistance check of each.

When powered from a 220-volt AC power source, the capacitors serve as voltage dividers to provide an artificial ground. Consequently, the capacitance and ESR (equivalent series resistance) values of the capacitors are critical when operating from a 220-volt line and have to be evenly matched, otherwise the switching voltages will be uneven. As electrolytics age, both the capacitance and ESR changes. If the mismatch is too great, one voltage could exceed the limits of the switching transistor, which can start parts a-poppin'. You can check the voltage balance with a VOM. Always replace both capacitors, not just one, and use a good grade capacitor, like the Panasonic TSU series.

It's Showtime

If you've made it this far, you probably have a working power supply. But before you apply power, let's make sure we've covered everything.

- You did a final resistance check on the output voltage lines, and all are within the specifications of Table 2, right?
- You checked the resistance across the AC input (with the power switch on) and it measures 1 megohm or better, check?
 - You checked the fuse.
 - Any broken wires or burned parts?

Good! Then it's showtime. Re-assemble the power supply. Plug the dummy load into one of the disk drive connectors. Apply power.

If both lights light, congratulations! You've got yourself a working power supply, because the power supply itself needs the -5- and -12-volt lines to operate. Consequently, you don't have to test them, unless you're as curious as I would be. Now all you have to do is put everything back together and enjoy a more peaceful day - except for the coffee. Here I suggest ... NV

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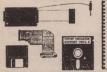
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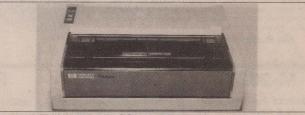
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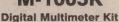
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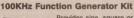
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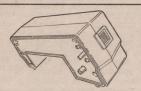
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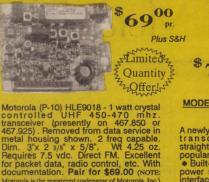
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ADC Motor PWM Speed Controller

by Lyndon Walker

ome time ago, I was building a project which used a small DC powered water pump. The pump was purchased from a surplus dealer and worked well, but ran a little too fast. I first thought of slowing it down by putting a power resistor in series with the motor leads, but later decided against that approach. My project, an automated plant watering system, used a 7.2V NiCd battery and although it was only on for a few seconds each day, I didn't want to waste power. The circuit I'll describe here was the solution to the problem.

There are many different ways to control a permanent-magnet DC motor, and they each have their advantages. Perhaps the oldest is by inserting a variable resistance in the power supply lead as shown in Figure 1.

Varying the speed this way is quite simple and very easy to do as long as you have a rheostat which can handle the required motor current. If just one speed is needed, the rheostat can be replaced by a single power resistor. Alternately, a group of power resistors connected in series can be used. A selector switch will then choose the desired tap point and so adjust the speed. This method, along with plenty of

oil to cool the resistors, was used for many years to control large motors. In fact, many radio-controlled models still use resistors as speed controllers. Inside the model, a servo motor turns a rotary switch which selects the desired speed.

In all these cases, the resistance will dissipate the unused power as heat, making this a very inefficient method of control. This is especially true if the circuit must run from battery power. However, because of its simplicity and ruggedness, resistance control is

Pulse-Width Modulation (PWM) is a much more efficient method of controlling motor speed. To do this, a constant stream of pulses at a fixed frequency and voltage is fed to the motor. The frequency is usually chosen so that it's outside the audible range, but low enough that the losses in the switching transis-

By varying the length of the pulses, we control the amount of time that the power supply is connected to the motor and thus control its speed. Changing a pulse train in this way — by varying the ON time of the pulses while keeping the pulse period constant - is known as Pulse-Width Modulation.

PWM settings are often specified by the resulting

Duty Cycle of the pulse train. The Duty Cycle is simply the percentage of time that the pulse is in an ON state. PWM is therefore considered as one form of Duty Cycle Modulation.

When the switch is in the ON state, the motor is connected to the power source. When it's OFF, there is no current flow from the power source. The switch has a low on-resistance and so it dissipates very little power as heat. Most of the heating in the switch comes as it is changing between ON and OFF states. By lowering the switching frequency and reducing the state transition time, these switching losses can be minimized. As a result, PWM control is quite power-efficient.

Because PWM requires only a fast switch and a means of controlling it, it has become very popular in computerized motor controls. In fact, judging from comments on the Internet robotics newsgroups, there are many people out there who believe PWM can only be done by programming a microcontroller or computer. Definitely not so. As the circuit in Figure 2 shows, an efficient PWM controller can be built at very low cost with readily available compo-

The central feature of this circuit is the NE556. This is a dual version of the popular NE555 or LM555 Timer IC. Each section of the '556 is configured differently. This makes it easy to modify the circuit. The first half (denoted on the schematic by legends ending in "1"), operates as a free running multivibrator. It generates short pulses at a fixed 19 KHz rate. These pulses are then used to trigger the second half of the IC which is configured as a monostable multivibrator.

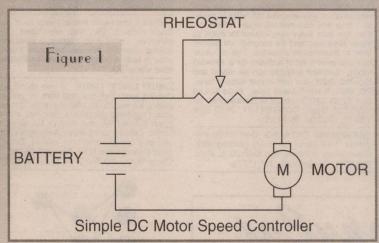
The monostable outputs a single pulse for each trigger pulse input. The length of the output pulse is controlled by T1, the variable resistor at pin 11, and is adjustable up to a maximum slightly less than the period of the trigger pulses. In this way, we create a train of 20 KHz pulses whose duty cycle we can control over a range of about 0 to 90%. At higher voltages, the limits are about 5% to 90%.

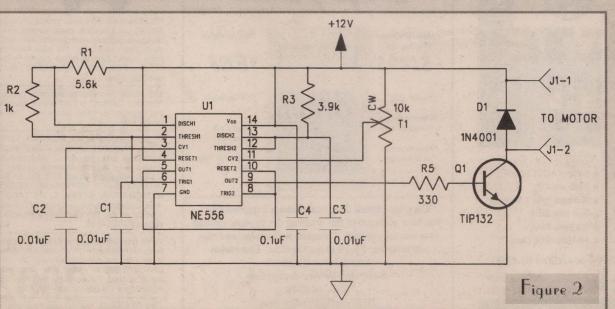
Pin 9 outputs the PWM pulses which control the TIP132 power transistor. This transistor is used as a "low side" switch which closes the motor circuit. The diode D1 suppresses the inductive kickback from the motor when Q1 switches off.

The circuit can be constructed on ordinary perfboard using wire-wrap or point-to-point techniques. Try to mount capacitor C4 as close to pin 14 of the 556 as possible. It is used for power supply decoupling and long wire lengths will diminish its effectiveness. Most component tolerances are not critical, but if you decide to substitute another transistor for Q1, be sure it can handle the motor current. The TIP120 available from Radio Shack is a good choice. The

TIP132 shown easily handles 3A with a small heatsink. You can also use one of the excellent International Rectifier IRFZ series of HEXFETs as long as the power supply is at least 12V. These HEXFETs switch rapidly and are available in a wide range of current ratings.

The circuit performs quite well over a supply voltage range of 5-12 VDC, so a regulator is not necessary. This makes it ideal for operation directly off the NiCd batteries often used to power hobby robots. The parts are inexpensive and available at Radio Shack or most of the surplus vendors which advertise in this or other magazines. This controller is an ideal small project for those rainy evenings. NV





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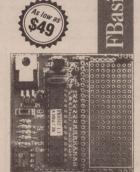
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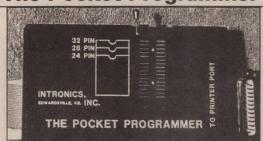
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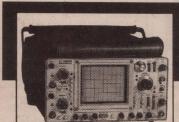
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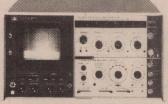
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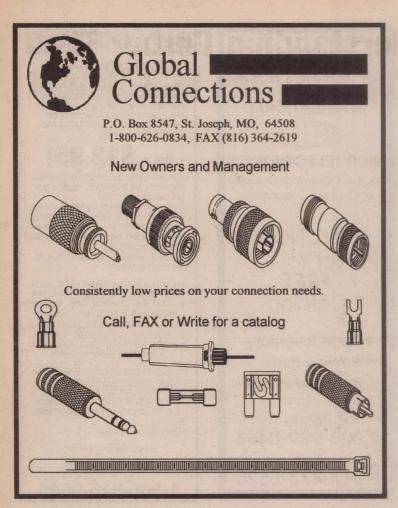
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by Scott Edwards

Understanding Gerber Files for Flawless PCB Output

You've designed and debugged a wonderful electronic widget using computerized circuit-board software, and now you want to have some printed circuit boards made. Should be simple enough, right? It can be, if you know how to generate error-free files for the board house.

This article will explore the mysteries of Gerber photoplotting files and Excellon drill files used in PCB manufacture. We'll look at strategies for getting the best results from the popular EASYTRAX and AUTOTRAX PCB CAD software, including the use of a free DOS program to take the drudgery out of producing flawless files.

Background

You can think of a printed circuit board (PCB) layout as consisting of two kinds of objects: tracks and pads. Tracks are solid lines of a particular width; pads are filled circles, rectangles, or other shapes. In the olden days, PCBs were laid out by draftsmen using black tape for the tracks and die-cut shapes for the pads. The layout would be photographed onto large

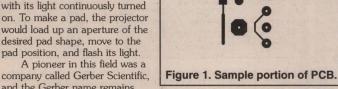
When computers came along, the drafting process was automated. However, the pen plotters used for regular drafting weren't entirely suitable for PCBs. For one thing, pens can not reproduce the full range of track widths required by PCB designers - at least not in a single stroke. And pens are very inefficient at making solid, filled objects like PCB pads. They have to spend time coloring in these shapes. Finally, a plot still has to be photographed onto film for use in the rest of the PCB process.

The problem was solved when it occurred to someone to replace a plotter's pen and paper with a light source and film. The light could be moved over the film and turned on and off to selectively leave images behind. Further, by making the light in the form of a tiny projector, the shape of the film images could be controlled. So, to make a track, the projector

would load up a circular slide or "aperture" of the appropriate size and move across the film with its light continuously turned on. To make a pad, the projector would load up an aperture of the desired pad shape, move to the

A pioneer in this field was a company called Gerber Scientific, and the Gerber name remains

associated with computer file format for the instructions that were used to control this kind of light plotter. The instructions themselves are set down in an Electronics Industry Association (EIA) standard, called RS274D.



Aperture Files

The idea behind the Gerber format is practical enough, but there was a complication: Not every photoplotter had the same set of apertures. And there was no universally agreed-upon correspondence of a particular Gerber code to a particular aperture shape and size. There couldn't be. We're talking about hundreds of commonly used combinations of aperture shape and size mapped to a relatively small number of codes (as few as 24).

So, the idea was to create two files: the first contained the plotting data, referring to apertures by a series of codes; the second contained the shape and size information for each code. For example, the plot file might say, in effect, 'move to XY coordinate soand-so, then flash aperture D23.' The aperture file would specify the size and shape of aperture D23. Here's where the fun begins: What if the aperture specified for D23 doesn't match what the PCB designer had in mind? Uh-huh, disaster!

Still harkening back to the old days, it was common practice to set up PCB designs according to apertures that your board shop had on hand. You'll see words to this effect in the instruction manuals of many older PCB CAD packages.

Gerber in the Modern Era

This stuff about mechanical plotters sounds quaint nowadays. You'd be hard pressed to find a board shop that uses one anymore. Mechanical plotters have long since been replaced by computerized raster photoplotters that are like laser printers on steroids (or, more accurately, like the photo-typesetters used in publishing). But they still use the Gerber file format.

Gerber has survived because of the inherent conservatism of business ("if it works, don't fix it") and

because of the format's versatility. A Gerber file is a text file that can be viewed and modified with a text editor or word processor. And it's relatively easy to buy or write programs that extract data from Gerber files for other manufacturing processes. So Gerber lives on.

There is an improved version of the Gerber format, referred to as Gerber-X or RS274X, in which the

aperture file becomes part of the plot file. If you own software that generates this kind of file, use it and skip the rest of this article. But the vast majority of inexpensive CAD packages, like AUTOTRAX and EASY-TRAX, are stuck in the Gerber-D time warp. So we still have to cope with separate plot and aperture files.

Plot File Format

A Gerber plot file is a plain-text file. If you open it with a text editor (like DOS EDIT or Windows Notepad) or word-processing program, you will see normal, printable letters-and-numbers text. The file is divided into 512-byte blocks consisting of 510 characters of data, followed by a carriage return and line

Blocks are divided into records. Records may vary

in length. The end of a record is marked with the asterisk character (*). A block contains as many complete records as will fit within its 510-character straightjacket; if there are unused characters, the end of the block gets padded with spaces.

Records are instructions to the photoplotter. For example, the record X200Y245* means move to the location with XY coordinates 200, 245. Dimensions are usually expressed in thousandths of an inch, referred to as "mils." Other examples of records:

Light on D02* Light off D03* Flash light

D23* Switch to aperture D23

Modern photoplotters typically allow a range of aperture codes from D10 to D269

The last two blocks of a Gerber file are special. The second-to-last block ends with the record M02*. The last block contains the backspace control character (ASCII 8), followed by 509 spaces, and ending with a carriage return and line feed.

Figure 1 contains a fragment of a PCB layout; Table 1 is the corresponding Gerber plot file.

Each Gerber file describes only one layer of the PCB, so you need a file for each layer used in the final PCB. For example, a two-sided board would require at least two Gerber files, one for top-layer copper, and one for bottom. But it could require more for such things as solder mask, parts-placement overlay, etc. To avoid confusion, there's general agreement on the DOS file extensions for each of the possible layers:

Top Layer	.GTL
Mid Layer 1	.GM1
Mid Layer 2	.GM2
Mid Layer 3	.GM3
Mid Layer 4	.GM4
Bottom Layer	.GBL
Top Overlay	.GTO
Bottom Overlay	.GBC
Power Plane	.GPW
Ground Plane	.GGD
Top Solder Mask	.GTS
Bottom Solder Mask	.GBS
Top Paste Mask	.GTP
Bottom Paste Mask	.GBP
Pad Master	.GPM
Keep Out Layer	.GKO
Drill Drawing	.GDD
Drill Guide	.GDC

Aperture File Format

The aperture file tells the photoplotter the size and shape of a pattern to be flashed (e.g., a pad) or dragged across the film (e.g., a track) for plotting. Aperture files are ordinary text files with entries as

shown in the ENTRIES sidebar.

The entries are separated by one or more spaces, and each line ends with a carriage return and linefeed. Table 2 is the aperture file from the Gerber file of Table 1. The meanings and usages of the entries are as follows:

D code. Just what it says: the D-code number, D10 through D269, assigned to this aperture in the Gerber plot file. (You should be aware that older photoplotters supported just 24 D codes mapped to fixed aperture numbers. This arrangement has gone the way of the Dodo.)

Shape. A word describing the shape of the aperture. There are quite a few recognized shape names, but not all shapes are supported by every photoplotter/board house. Here's a sampling of shapes used in aperture files:

CIRCULAR
OCTAGONAL
OVAL
SQUARE
RECTANGULAR
ROUNDRECT
CROSSHAIR
MOIRE
LINE
ANNULUS
THERMAL

Some pads can be described by more than one combination of shape and size specifications. A rectangular pad with identical X and Y dimensions is obviously a square, for instance. And some shapes are automatically converted to others by software or personnel at the board house. Although some systems allow a shape name of "LINE," circular apertures are generally used for both pads and line drawing.

X size and Y size. These are the aperture's dimensions in the X and Y coordinate system of the board. The sizes are usually expressed in mils, but may also be expressed in decimal inches. Note that if a board uses the same pad twice, but rotated 90 degrees, its aperture file will contain two entries for that pad with the X and Y size entries swapped.

Hole. Size of hole to drill in a given pad (usually mils, but may be decimal inches). This entry is frequently set to 0 (no hole) so that the hole sizes can be specified on a case-by-case basis through the drill and tool files (discussed later). The reason is that many

EN	TRIES	-						
	rmat:	D code D11	Shape CIRCULAR	X size	Y size 70	Hole 0	Use MULTI	

and tool-movement instructions. We won't go into it in detail, because there are several formats, and frankly, drill files almost never cause problems! It is worth the time to look at the companion tool files, though, because you may want to edit them.

Tool files (DOS file extension .TOL) are normal text files. Their contents can be as simple as a tool number, T01 through Tn (where n is the largest tool number used in the job) followed by the size of the hole to be drilled. Optionally, they may contain other information generated by your CAD program, including the number of holes drilled by a given tool, and even the total distance the tool will travel in moving from hole to hole of the PCB. The board house doesn't necessarily need this information, but may want it to factor into price quotes or job planning as this information can help in formulating an estimate of how long the board will take to drill. Here's a tool file from a PCB prototype I recently had made:

T01 .125 T02 .032 T03 .035 T04 .040 T05 .024

The drill sizes are in inches, so T01 is a 1/8" (0.125") bit, T01 is 0.032" (32-mil) bit, and so on.

Some board houses, particularly those specializing in fast turnaround, try to limit the number of tool changes. They usually do this by charging extra for drills not on their standard drill list. For example, I sent the board whose tool list appears above to AP Circuits, the Canadian prototyping house (Sources). Their standard drills in the range required by my prototype are: 0.028, 0.035, 0.042, and 0.125. To avoid the nonstandard drilling charge, I edited the tool file to map my "nonstandard" drills to AP's standard ones:

T01 .125 T02 .035 T03 .035 T04 .042 T05 .028 to the tool file. If you make a drill size larger, you have to ensure that it won't be too large for the smallest pad in which it will be drilled. Most board houses recommend a minimum of 10 mils (0.01)" of copper margin around a hole. When the pad is round, this copper is called the annular ring. To determine the annular ring for a given combination of pad and hole, calculate (pad-hole)/2. So if the pad is 62 mils and the hole is 35 mils, the annular ring is (62 - 35)/2 = 27/2 = 13.5 mils.

Generating Gerber Files in -TRAX Software

EASYTRAX is a free, DOS-based software package for generating PCB artwork; AUTOTRAX is its popular commercial counterpart. (EASYTRAX is available for downloading from the Nuts & Volts library on the Internet; go to www.nutsvolts.com and select the library option.)

These programs are of quite high quality, as they represented the state of the PCB-CAD art for personal computers at a time when a '286 processor was considered fast. But one area in which they really show their age is the generation of Gerber files. When these programs were current, the list of apertures available from your friendly photoplotter determined what shapes and sizes you could use in designing a PCB.

This expectation is hardwired into the way that the companion TRAXPLOT program generates Gerber files. The program combs through your aperture list, and tries to match the shapes used in your design to shapes available. You can set a "match tolerance" to allow the program to pick apertures that are close enough. Or you can override the program's match and select the apertures yourself. The program makes the outdated distinction between apertures coded for flash and stroke uses, and won't select an aperture for an "unauthorized" use without your permission.

All of this was necessary back when you had a limited set of apertures to work with. But with modern photoplotters, the opposite approach makes more sense: Use the shapes and sizes you need in your

Table 1. Gerber File for Figure 1 (spaces at end of file removed)

X0Y0*D11*X1725Y2700*D01*X1725Y2700*X1725Y2875*D02*X1825Y3000*D01*X1825Y3 000*X2000Y3000*D12*D02*X2000Y2900*D01*X2000Y2900*X1925Y2825*D02*X1925Y28 25*D01*X1925Y2825*X1925Y2775*D02*X1925Y2775*D01*X1925Y2775*X2000Y2700*D0 2*X1725Y2875*D01*X1725Y2875*X1725Y3000*D13*D02*X2000Y2900*D03*X2000Y2800* D03*X2000Y2700*D03*D14*X1725Y3000*D03*D15*X2000Y3000*D03*D16*X1825Y2800* D03*D17*X1825Y3000*D03*D02*M02*

board designers specify more than one hole size for pads of the same size, depending on the size of the wire or component lead that will be soldered to a particular pad.

Use. This entry can be used to direct the use of a particular aperture for flashing pads (FLASH), drawing lines (LINE), or both (MULTI). If there's no entry, the default is MULTI.

Drill and Tool Files

The situation with drill files (several formats; Excellon is the most common) and their companion tool files is similar to that of Gerber plot and aperture files. The main difference is that the drills still do physically change bits for each size of hole drilled, while the photoplotter just simulates the change of apertures.

The drill file (DOS file extension of .DRL or .NCD for numerically controlled drill) looks a lot like the Gerber plot file; it's a series of X and Y coordinates

Why edit the tool file rather than make the changes in the CAD program? Well, the board will eventually be manufactured, and those hole sizes are required for the components to fit properly. But on a quick, cheap prototype I can tolerate some slop. Editing the tool file makes the change completely temporary — when the board goes into production it will do so with the original, unmodified tool file.

Another reason for editing the tool file is to accommodate variations in procedure from one board shop to another. Some shops use the exact drill bits listed in your tool file. If the board is made with plated-through holes, the actual size of the holes will be several mils smaller than the drill due to the thickness of the plating. Other shops look at those drills as finished hole sizes. If the holes will be plated, they're drilled slightly oversize so that after plating they're the size you specify. You may want to consult your shop and find out which form of reckoning they use, and adjust your tool file accordingly.

Some caution is required when you make changes

Table 2. TRAXAN-Generated Aperture File for Figure 1						
D10	CIRCULAR	5	5	0		
D11	CIRCULAR	20	20	Ō		
D12	CIRCULAR	10	10	0		
D13	CIRCULAR	62	62	0		
D14	RECTANGULAR	50	120	Ō		
D15	DECTANGULAR	62	62	Ō		

OCTAGONAL

ROUNDRECT

D16

design, then compile an aperture list to match. When you run TRAXPLOT to create your Gerber files, just select your "perfect" aperture file, and run the process with 0 match tolerance. Since all of the apertures are there, the program will breeze through the matching process.

100

100

0

There are at least three ways to compile a perfect aperture list.

(1) Run TRAXPLOT once to find out what apertures you need, write them down, and manually type up a text file in the aperture format shown above.

(2) Open the -TRAX .PCB file with a text editor, read the database entries, mentally convert them to apertures, and manually type up an aperture file.

(3) Use my TRAXAN program to read your .PCB file and generate an aperture file.

To write TRAXAN, I had to figure out the format of the .PCB files generated by EASYTRAX and

Table 3. AUTOTRAX Native .PCB file for Figure 1

AUTOTRAX. Fortunately, the format turned out to be an extremely simple structured text document that vou can actually sight-read. Table 3 shows the .PCB file for Figure 1.

Let's look at how the -TRAX programs store the details that make up a track. They identify a track entry with the abbreviation FT or CT on a line by itself: FTs are 'free' tracks those not associated with a predefined component - while CTs are 'component' tracks usually partsplacement outlines. The next line sets down the characteristics that define the track:

Start X Start Y End X End Y Width Layer ?

Here's an example track entry:

1750 1700 1825 1700 8 6 1

That entry defines an 8-mil wide track that goes from XY coordinates 1750,1700 to 1825,1700 on the bottom PCB layer (layer 6). I never did determine what the ? entry is for, but I didn't try very hard, since all we need to know for our aperture file is the width!

The format for pads is similar. The program identifies a pad as free (FP) or component (CP), then specifies the information as shown in the LOCATIONS

Here's an example pad entry:

LOCATIONS -

X Location Y Location X Size Y Size Shape Hole Size ? Layer

1000 2000 60 80 4 30 1 13

That entry defines a type-4 (rounded-rectangular) component pad at XY coordinates 1000, 2000 with dimensions of 60 mils in the X direction and 80 in the Y. The pad has a 30-mil hole and is intended to be plotted on all copper layers (13). The layers are numbered as you might expect, from top (1) to bottom (6), with the number 13 used to mean multilayer. The pad shapes are numbered as follows:

- 1. Round
- 2. Rectangular or Square
- 3. Octagonal
- 4. Rounded Rectangular
- 5. Crosshair Target
- 6. Moire (concentric circular) target

So, you can see how it would be possible, if a bit tedious, to read the .PCB files and manually compile an aperture list. For tracks, you would look at the width entry and place a correspondingly sized circular aperture in the file. For pads, you'd convert the shape number to the appropriate shape-descriptive word (ROUND, RECTANGULAR, etc.) and enter the XY dimensions. You would, of course, avoid duplicate definitions; one aperture entry can be used for an unlimited number of pads or tracks: You'd also avoid giving two different apertures the same D code, since the format prohibits redefining the codes.

I wrote the TRAXAN program to automate this process. To use it, you just type "TRAXAN pathname" at the DOS prompt, where pathname is the drive, file name, and extension of your board, such as C:\PCBS\MY_BOARD.PCB. The program reads all of the pad and track entries in the PCB file and prepares an aperture file, which it saves to the same path as the PCB file, but with the extension .APT. In the example above, the output file would be C:\PCBS\MY_BOARD.APT.

By default, the dimensions in a TRAXAN-generated aperture file are in mils. You may change this to decimal inches by adding a slash (/) after the pathname: "TRAXAN pathname.PCB /." Note that this must be a forward slash (/) as DOS uses the backslash (\) as a separator for parts of a pathname. However, to prepare files for TRAXPLOT, omit the slash.

After running TRAXAN, use the TRAXPLOT program to generate your Gerber files as usual. Remember to specify the TRAXAN-generated aperture file, instead of TRAXPLOT's default STAN-DARD.APT. For all of the entries having to do with

matching apertures to shapes, accept the program's defaults. Since the TRAXAN-generated aperture file will have an exact match for every shape used in your design, the matching stuff will never come into play.

One note regarding TRAXAN: This program generates aperture files that make TRAXPLOT happy. Using the slash character (/), you can get it to output aperture files for certain other programs, like Gerberfile printing and viewing utilities. But TRAXAN aperture files won't please every program you encounter. I found that GCPREVUE, a free Gerber viewer found on the AP Circuits web site, wants a more elaborate aperture file. You can manually edit the TRAXAN aperture files, or modify the TRAXAN.BAS program to generate this format.

TRAXAN is available free from the Nuts & Volts library on the Internet in both executable (.EXE) and source (.BAS) formats. The files are combined into a single file, TRAXAN.ZIP, requiring PKUNZIP to unpack. TRAXAN is unsupported freeware - I've thrown it into the public domain for people to use, modify, and distribute without restriction. I've used the program to generate my own aperture files with complete success. However, TRAXAN was written quickly as a brute-force tool, so there's no support or documentation for it.

Summary

Working with Gerber files is easy, once you understand the relationship between plot files and aperture files. Despite changes in photoplotting technology, you can still produce excellent results with older PCB CAD software. NV

Sources

Inexpensive PCB prototyping: AP Circuits, #3-1112 40th Ave. N.E., Calgary, Alberta, Canada 72E 5T8; phone 403-250-3406. In addition to fast-turnaround, low-cost PCB prototyping services, APC offers an excellent collection of freeand shareware programs for PCB CAD, and documentation for novice PCB designers. Sign onto their dial-up BBS at 403-291-9342 or access their Internet web site at http://www.apcircuits.com

AUTOTRAX PCB CAD program: Protel Technology, Inc., 4675 Stevens Creek Blvd. Suite 200, Santa Clara, CA 95051; Sales toll-free: 1-800-544-4186; Callers outside the US/Canada: 408-243-8143; Fax: 408-243-8544.

TRAXAN program: Go to the Nuts & Volts web site (http://www.nutsvolts.com). Press the button to navigate to the library, and download the file TRAXAN.ZIP. This compressed archive file contains both an executable version of TRAX-AN (TRAXAN.EXE) and the QuickBASIC source code (TRAXAN.BAS). TRAXAN is unsupported freeware.

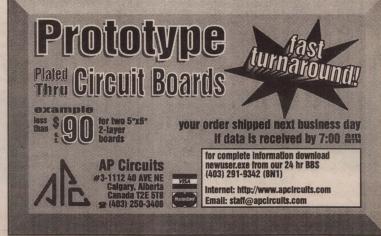
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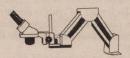


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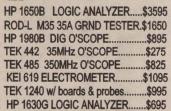
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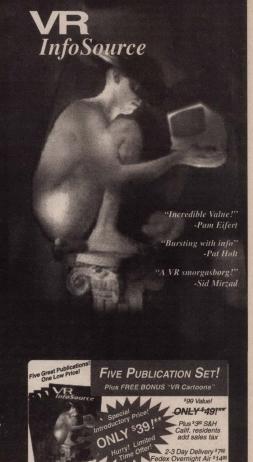
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Flame Speaker: An Experiment Plasma Acoust

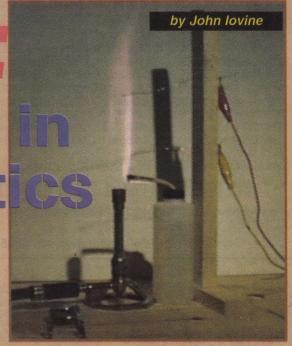
bout 25 years ago, I read an interesting article on three researchers - W. Babcock, K.L. Baker, and A.G. Cattaneo - who discovered an interesting phenomenon of plasma acoustics while working at the United

Technology Center in California.

The researchers were able to use a high-

temperature gaseous flame from a propane or butane torch that is seeded with potassium nitrate.

Any audio sound source may be used, but I recommend using music because it has more high-frequency components in it than speech (see Figure 1).



into two electrodes positioned into the flame. The flame is seeded with a water solution

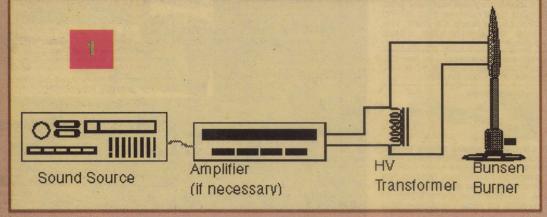
of potassium nitrate to induce lowtemperature ionization.

How it works

A water solution of potassium nitrate is fed into the flame by a alcohol lamp wick placed at the bottom of the flame. The potassium nitrate ionizes easily, by thermal ionization, inside the flame. lonization creates free electrons and positive potassium ions. When you supply a high-voltage electrical signal, the positive ions move toward the negative electrode, and electrons move toward the positive electrode. In the case of an audio signal, the electrical signal fluctuates (switching back and forth rapidly), causing the flame

to vibrate.

Flames, like a liquid, exhibit a surface tension. A gaseous membrane is created by a difference in density and high temperature of the torch flame to the ambient air. The membrane acts as a diaphragm to compress and rarify the air, and produce sound waves as the flame vibrates.



Wire Electrodes Seeder Transformer

temperature plasma as a speaker. They discovered that the response of the plasma speaker is better at higher audio frequencies than at low audio frequencies. This drawback, they theorized, may be a result of the small size plasma, rather than an intrinistic limitation.

Rather than trying to create a hightemperature plasma, we can use a lower

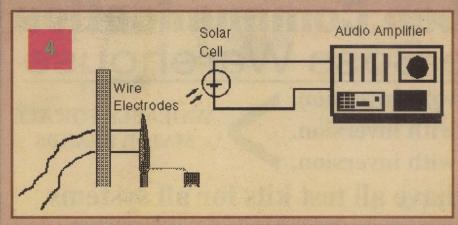
500 VDO Audio Source HV Bunsen Transformer Burner

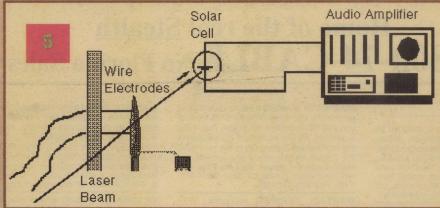
source (radio, tape, or CD player) must supply 10 or more watts of power to the HV transformer to be heard. If necessary, feed the audio sound through a power amplifier. The more power, the greater the volume. From the amplifier, the audio is fed into a step-up transformer.

The output of the step-up transformer is fed

Experimenting with plasma acoustics

Figure 2 illustrates a simple set-up for you to begin experimenting. The seeder I used was fashioned from a plastic bottle I had lying around, but just about any container will work.





Headphones

I drilled a hole in the top of the plastic bottle and fed the lamp wick through it. An alcohol lamp wick is used to continuously supply the potassium nitrate into the flame. The potassium nitrate strength isn't critical, dump a spoonful into any quantity of water and it will work fine.

The step-up transformer is a high-voltage transformer from

This is a threewire device. The two outside wires are the primary wires where the audio signal is fed. The center wire from the transformer is the high voltage out. The high voltage out is connected to one of the electrodes placed into the flame. The second electrode is connected to either one (but not both) of the primary wires. Be careful when operating the circuit; the

Images Co.

audio signal from the amplifier is stepped-up by the transformer and is capable of shocking you.

When running the plasma speaker, you'll notice a glow emanating from the tip of the electrodes. The glow keeps in step with the volume of the sound. If the electrodes start arcing, either move the electrodes further apart or turn down the volume on the amplifier.

Remember to turn off all equipment before adjusting or touching the electrodes. They are connected to a highvoltage transformer and are quite capable of shocking you.

Improving the

The experiment can be improved by using a larger and hotter flame. Anything that would do so would increase the quality of the sound. If possible, use nichrome or tungsten wire for electrodes. These wires will last longer in a high-temperature

Supplying a 500 DC bias to the circuit will improve the quality of sound (Figure 3). This allows the sound signal to travel on top of the DC bias. It

reduces the polarity fluctuation of the audio signal inside the flame and improves the fidelity of the

Other experiments

A solar cell or photo-transistor that is connected to an audio amplifier will output the sound signal fed into the flame when the light from the flame is directed onto the photocell by a lens (see Figure 4).

Another possibility is to direct a laser beam between the electrodes in the flame. This may be low cost to modulate a laser with information (see Figure 5). In the 1920s, Lee DeForest replaced a radio crystal detector with a flame (see Figure 6). NV

Parts list:

Bunsen Burner or Butane torch - Local Supplier

HV transformer \$19.95 Potassium Nitrate, 1/2 oz., packed in water \$6.00 Alcohol Lamp Wick \$1.00

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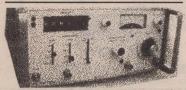
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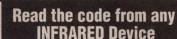
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\$149.00

C-BAND LNB





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Twenty very bright red LEDs mounted in a plastic reflector/lens housing. Measures 16.5" long by 1.2" wide. 12VDC @ 400mA opera-From auto brake light.

96L001 \$4.95 each

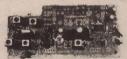


12VDC FAN

Measures 2" sq. x 0.625" thick. These are

96F003

\$4.95 each



6 METER FSK RECEIVER

Crystal controlled on 50.675 receive frequency. Super het with 10.7MHz IF. Three ICs and one transistor. Not a kit! New, with schematics and spec sheet for major compo-

92A067

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SIGNAMAIL*M

This transmitter/receiver pair is designed to alert you when your mail comes. The transmitter is activated by a tilt switch and broad-



casts coded tone that turns on a light and plays a tune in the receiver. These units have dozens of other uses such as perimeter or driveway alarms, gate warning, game trails, etc. The internal tilt switch is easily bypassed for remote switching. Documentation included.

\$14.95/Set 957012

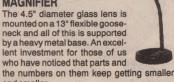


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Powerful two-stage, wireless mike kit transmits over half mile (up to 1Km in the open). Tunable to the upper part of FM broadcast band. Circuit requires the ability to tweak RF circuits by stretching or compressing coils. With mike and 9V battery clip. Operates from 6-12VEDC. 2.75" \times 0.625" \times 0.5". Sold for educational purposes only.

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and smaller. MC205 \$14.50 each

HEBTER

FIVE POUNDS OF FAMOUS BRAND SOLDER 0.125" diameter, 29% Tin, 61% Lead, solid

92Z012 5 Lb. Spool - \$19.95

QUACK MEDICINE KIT

Contains all parts and instructions necessary to build a "muscle stimulator." Runs on a 9V battery (not included). Produces a 90V pulse with approximately one second repeat rate. We suggest our Grey Project Box (92U062) for this kit. Sneak one into Biology Lab and make a dead frog's leg kick

Only \$9.95 each 94U002

DUKE'S DELIGHT

Five pounds of assorted electronic hardware fasteners, nuts, bolts, screws, brackets springs, standoffs, arms, gears, pulleys, and indescribable metal stuff scrounged from Silicon Valley's surplus to complement your hardware hoard. 92Z009

5 Lbs for \$4.95

PHONE LINE MANAGER

Better than Caller ID!

This device detects an incoming call. If you subscribe to caller ID, the unit will automatically receive and display the caller's telephone number. If not, the unit will ask the caller to enter their telephone



or personal ID number. The device will then display and compare this telephone or personal ID number to a memory bank of numbers that you've installed. If the system finds a match your phone will ring. If the system does NOT find a match, the caller will be directed to leave a message. All messages are digitally recorded. No tape or answering machine necessary. New, in factory cartons, with documentation.

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REVERSIBLE DC GEAR MOTOR

Usable voltage range approx. 12 to 28 VDC. Good to several hundred RPM. Measures 1.375" dia. by 2" long. Soho Part No. E-119-

96M001

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Recycled. 95W011

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WOW! May contain tape wound, powdered metal, ferrite material cores, different sizes 20 for \$4.95



TITE-FIT BOX CHASSIS

An orthodox box that has many uses and is engineered to deserve its name of Tite-Fit. Screw fastening. All screw fasteners are punched and tapped. Selftapping screws are not used. Available in Plain Aluminum.

Model#	W"	<u>D"</u>	H"	Price
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TF771	2-3/4	2-1/8	2-3/4	3.84
TF772	3-1/4	2-1/8	1-1/8	3.40
TF773	3-1/2	2-1/8	1-5/8	3.66
TF774	4	2-1/8	1-5/8	3.84
TF775	4	2-1/4	2-1/4	4.62
TF776	4	2	2-3/4	4.63
TF777	4-1/4	2-1/4	1-1/2	3.84
TF778	5	2-1/4	2-1/4	4.96
TF779	5	4	3	6.92
TF780	5-1/4	3	2-1/8	5.47
TF781	6	5	4	9.68
TF783	8	8	3-1/2	11.92
TF784	10	6	3-1/2	14.10
TF785	10	2	1-5/8	6.25
TF786	12	2-1/2	2-1/4	8.67
TF787	12	7	.4	19.59
TF788	17	5	4	21.51



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RESOURCE BIN

number fifty six

Creating your own web home page.

ur usual reminder here that the Resource Bin is now a two-way column. You can get tech help, consultant referrals, and off-the-wall networking on nearly any electronic, tinaja questing, personal publishing, money machine, or computer topic by calling me at (520) 428-4073 weekdays 8-5 Mountain Standard

I'm now in the process of setting up my new Guru's Lair web site you will find at (where else?) www.tinaja.com — This is the place you go for instant tech answers. Among the many files in our library, you will find complete reprint sets for all of the Resource Bin columns. Along with some new "rich text" global search features.

The site is still under construction, but I fully intend to move thousands of my files here. So, definitely come visit. But bear with me for a while till things really get into gear.

You will get the best results if you have both Netscape Gold and Acrobat Amber 3.0. You download these from www.netscape.com or www.adobe.com

Your Own Web Site

Last month, we looked into all the fundamentals of getting online. After you have your ISP and host software in place, and after you have surfed the net for a month, the chances are you might want to start up your own personal web home page.

Before you begin, though, take time to write down all the exact goals you want your site to accomplish. It is reasonable for a site to make others aware of you. Or to provide info on your products. It is reasonable for the site to focus on one topic.

But it is not reasonable to expect a site to generate scads of direct cash sales. It is not reasonable to expect thousands of daily visitors to your site. Unless you have an exceptional sole-source something to offer.

By far, your three most important features that you can offer on your site are content, content, and content.

Beyond lots of content, you should also secondarily aim to be a gateway site that provides great heaping bunches of well-annotated and working links to other quality web locations.

Your first and foremost rule is to study hundreds of existing sites to find out what does and does not work. I am not that convinced that presenting lots of highly personal information online is such a good

For instance, if you are female, your personal content will be treated as a porno site. At least by many viewers. If you must present a resume online, better to have a short summary plus a "click here for full E-Mail version."

And spare us the drivel. I couldn't care less about your parakeet.

Unless, of course, you've trained him to program in PostScript while he quests Mount Graham tina-

Note also that properly maintaining a useful web site demands

NEXT TIME: Don reviews fundamental resources for electronic servicing.

an absolute minimum of 20 hours of work per week. Sometimes much

You should find two types of sites: registered and non-registered.

A non-registered site is more or less free and is simply some space your ISP leases to you. If your ISP is aaron, your home page site address will be www.aaron/~your-

Being free is all a non-registered site is worth. Others can freely use your domain name. Getting your site indexed may be tricky. Your address is a free ad for your ISP. And if your ISP goes belly up (or you seek a better one elsewhere) you'll have to rename and repromote your site.

Instead, a registered site costs you \$50.00 per year. This gives you exclusive use of the domain name. Your name ends in .com when you are a for-profit venture. Or ends in

To find out if a domain name is still available, go to www.internic.net and use their whois feature. My first two choices of synergetics and guru were not available. So tinaja, of course, was the next logical choice.

To register your domain, you once again should contact www.internic.net - Then follow their online instructions. Note that registration can not be done by mail or fax. It must be done online.

Next comes the home page design. While you can hire others to do this, it is fairly easy to do yourself. You also get exactly what you want exactly how you want it this

There are all sorts of web designer tools. Adobe's new and free beta Web Presenter is fairly handy, along with their Page Mill. But all you really need is a free copy of Netscape's Navigator Gold. Netscape is all I ever use.

You may have heard of HTML. It is ludicrously absurd to call HTML a language. At best this is a crayon. A mesmerizing awful atrocity that goes out of its way to prevent your end user from seeing what you really want to show in exactly those forms and fonts you wish to show

Adobe's great Acrobat Amber 3.0 is infinitely superior to HTML. On all counts. It quickly and attractively lets you show any text and graphics and photos any way you want.

When it is properly installed into Netscape, you can instantly view files without needing a special

download or changing the program in use.

That anti-aliasing Adobe calls text smoothing makes all text ridiculously more legible. Their magnifier feature can instantly blow up schematics and such for greater detail.

Thus, a key web rule: Use an absolute minimum of HTML on your site! Favor Acrobat Amber instead.

Uh, you probably should pick up a book on HTML somewhere along the way. Jeff Duntemann has lots of them at his Coriolis Books.

Steal The Plans

Your simplest way to build a web page is to steal the plans. Find a page you like on the web and select the edit feature of Netscape Gold. Which will rapidly transfer copies of all files you'll need to duplicate the site.

Replace their text copy with yours. Then change the sizes and shapes of the graphics by using suitable editors. Grab the wallpaper into a GIF or JPEG editor and shift the color and density around. Mix and match the tools from several dif-

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Hardware Hacker II, III or IV	\$24.50
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PostScript Beginner Stuff	\$29.50
PostScript Show and Tell	\$29.50
Intro to PostScript Video	\$29.50
PostScript Reference II	\$34.50
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SYNERGETICS Box 809-NV Thatcher, AZ 85552 (520) 428-4073

Write in 39 on Reader Service Card.

RESOURCE BIN

SOME INTERNET START-UP RESOURCES

Adobe Acrobat 1585 Charleston Rd Mountain View CA 94039 (800) 833-6687

Boardwatch 8500 W Bowles Ave #210 Littleton CO 80123 (303) 973-6038

Coriolis Books 7339 E. Acoma Drive #7 Scottsdale AZ 85260 (520) 483-0192

Gale Research 835 Penobscot Blvd Detroit MI 48226 (313) 961-2242

Inter@ctive Week 100 Q Roosevelt Blvd #508 Garden City, NY 11530 (516) 229-3700

Internet Life One Park Avenue New York NY 11016 (212) 503-3500

Internet World 20 Ketchum Street Westport CT 06880 (203) 341-2872

Internet Yellow Pages 2600 Tenth Street Berkeley CA 94710 (800) 227-0900

Internic Registration Svcs 505 Huntmar Park Dr Hernon VA 22070 (703) 742-4777

LDS/WEB Trac 571 Central Avenue Murray Hill NJ 07974 (800) ASK-LDSI

The Net 1350 Old Bayshore #210 Burlingame CA 94010 (415) 696-1688

600 Community Drive Manhasset NY 11030 (516) 562-5000

Netscape Gold 501 E Middlefield Rd Mountain View CA 94043 (415) 528-3777

430 Princeland Court Corona, CA 91719 (909) 371-8497

Thatcher AZ 85552 (520) 428-4073

WEB Techniques 600 Harrison Street San Francisco CA 94107 (303) 661-1885

Po Box 55549 Seattle WA 98155 (206) 782-7733

WEB Week 20 Ketchum Street Westport CT 06880 (708) 564-1385

544 2nd St 3rd Fl San Francisco CA 94107 (415) 904-0664

Yahoo Corp 635 Vaqueros Avenue Sunnyvale CA 94086 (408) 328-3300

You'll have the best luck by cross linking sites that are the same size as yours or somewhat smaller. Always provide their link on your site

have a personal interest in a site. others will too. It is better to focus on stuff you can genuinely use.

And don't just provide a link. Tell your users what this link is and why you like it. Avoid using any text-only links. Instead, provide a button or some larger graphic to make for

By the way, if someone provides you with a free link, netiquette leaves you with two and only two options: Quickly give them a hassle free link in return or else send them an annual check for \$30,000.00.

in promoting your site. All the printed web magazines are always looking for new sites to review. So suggest yours. Several times. Any time you see a list of web sites anywhere, figure out how to contact the

Of course, you should prominently mention your site in any printed ads you may be running. And don't forget the power of the advetorial. Especially when it's combined with some subtle mention to visit my own web site found at http://www.

Promoting Your Site

before you display it.

So, once you get your web site up and running, how do you get anyone worthwhile to visit it?

so will loudly proclaim that "this site

sucks." Which strongly suggests

adding 52,761 to your web count

counters are useful or desir-

when they stick

and give you

obnoxious error

messages. Or

if they spill Java all over

an odometer

reading less than 50,000 or

Besides.

your screen.

Especially

able.

First off, it is probably not a good idea to promote your site too early. What you think you are going to have and what you really are offering are usually wildly different.

It is a good idea to have your site up for at least six weeks before you aggressively promote it. Before any promotion, it is also a good idea to already have so much content on your site that it can't be covered in one visit.

Second, you might not want just anyone visiting your site. Instead, you will probably want to focus all your promotion efforts upon those who are likely to share your views. Or those who may buy something from you.

Traffic for traffic's sake makes no sense whatsoever.

At any rate, your first promotion effort should be to link to the index services. Start with yahoo.com and use their add url service. Then use Yahoo's links to reach the other indexes.

Better yet, you might check out www.2com.com/~upfront/launch/ This free service can link you up to the top 20 or so index services.

Next, go to a really great site that wins lots of awards and boasts of it. The Science Hobbyist site you'll find at www.eskimo.com/~billb/ should work out just fine. Then mine their hot list of award issuing sites.

It is also a good idea to register your name and E-Mail address with such services as www.whowhere. com - Along with your site domain. And, of course, be sure to have a hot site link inside your E-Mail signature file.

But your most important and most useful traffic builder is the exchange link. You first provide a hot link to somewhere and then ask them to give you one back.

Threads in the web.

ferent sites.

Finally, get into an editor and view the actual HTML text. Then eliminate anything obviously proprietary. As a final test, do make certain your final appearance of the site does not look even remotely like the original.

There are also lots of web designer programs out there that include ready-to-use templates. Buttons, doodads, rules, and Java applets can be found on hundreds of web sites.

Speed, Not Glitz

A few site webmasters appreciate how infuriating it is to mope around waiting for a humongous and totally useless image to download. A bored or impatient visitor is a gone one.

If you must use images, compress them with Acrobat or JPEG. Never use GIF! Keep the images small.

I invite you to compare the access speed of my www.tinaja.com against anything else on the web. To run at this speed, I first use few and small graphics on the HTML

On the Acrobat side, I use Adobe's new Amber 3.0, which offers dense compression, webfriendly file order re-arrangement, and (eventually when the ISP gets around to it) faster page file block downloads.

Measuring Visits

How can you tell how many people are visiting your site?

You may have heard the term "hit." The hit is an obscenely bloated and a totally misleading measure for the popularity of a web location.

You see, a person does not really "visit" a web site. They instead ask for small packet bits and pieces.

In random order.

For instance, a visit to a home page may return the page itself in a single packet, an image or two in additional packets, and possibly a few artistic rules, buttons, or selected gimcracks. Dozens to hundreds of site "hits" may be involved in one access.

On top of that, some visitors arrive by mistake, take one look and leave. Yet others are those spiders and the tireless net crawlers from the index services. Some visitors may promptly return in a few minutes for one reason or another. Is this one hit or two? What if they return in an hour? In a day? In a week?

Instead of the "hits," you should be worried about how many important and key files of yours are getting fully downloaded. Say your catalog or a hot and unique new reprint.

There is one standard way to sort all of this out: Your ISP has two files related to your site called the log file and the error file. You should make arrangements with your ISP to gain routine access to these files.

Your log file line entry describes every access by who made it when. Repeated entries tell you how long the visitor stayed, which pages they accessed, and how they left. The error file is particularly useful to spot any internal links you have that are not working. Or to spotlight ISP traffic problems. Sadly, the error file can not spot a bad external link.

Once transferred to your host, these are plain old text files in a standard format. You can snoop around with any old word processor or editor. Or enter them into any spreadsheet or data base program.

What if you really want to know how many left-handed Brazilian cat fanciers visited your site at 2 am on Tuesdays? My own preference here is to stash the log file on my PostScript printer's hard disk. And then use PostScript as a super power general-purpose computer language to extract and present useful results. But you'll find lots of web tracking software out there. The free Webtrac site analysis software from www.lds.com is an older, but more or less typical code.

A pair of quick-and-dirty rules of thumb: To approximate your actual number of useful visitors, divide your hits by 60. Or log characters by 4500.

There are also "odometer" visitor number services. But I feel that this is odometer as in odious. I'm not all that convinced that these web first. The chances are that if you

easier mousing around.

Don't forget non-web resources

tinaja.com

Unpromoting Your Site

What if you receive too much site traffic? So much sheer volume that it is overloading your ISP? Or causing dropouts, delays, and other problems to those you really want visiting?

First, make sure your home page keywords are tightly focused. If you put sex, drugs, or rock and roll in your keywords, you'll be generating a lot of worthless traffic. As will getting yourself on a porno filter blacklist.

For every porno "bandstop" filter, you can also find a porno "bandpass" filter elsewhere on the net. These essential services clearly complement each other.

For every dry spot, there's a wet spot. Yin and yang.

Your ISP can install software filters for you that restrict unneeded traffic. These are sometimes called JavaScript applets. Or else use a CGI, short for Common Gateway

If, as I do, you are unable to offer any foreign sales or service, there's no point at all in welcoming any non-US country codes. If you are not busily ripping off the government, then a block of .gov domains might end up a good idea. To avoid non-buying students, block .edu.

Those indexing web crawlers can have a stick run through their

RESOURCE BIN

treads by proper use of a robots.txt

Certain traffic problems can be ISP caused. This can be the case if a lot of send timed out or read timed out errors appear in your log file. The solutions here are to complain loudly. Or to pick a better ISP.

Your E-Mail traffic can be very much reduced by ignoring most of it. And, above all, not unleashing a pen pal by repeated responses. I use this criteria as an E-Mail filter: Is there a 10 percent probability of a 10 cent net positive cash flow as a result of my answering?

The surest methods to dramatically reduce your visitors is to require use of passwords or charge an access fee. Which brings us around to ...

Showing A Profit

Very few web sites generate a net positive cash flow. But it is possible. If you are in it for the long term. If you give lots away to sell a little. If you use *pull* advetorials rather than *push* advertising. If you are seeking more indirect results (referral, awareness, contact) than any immediate hard cash sales. If you have totally unique products offered at a fair price.

And, above all, if and only if you have great heaping bunches of useful and unique site content.

There are four main ways to fund a site: by use of sponsors, advertisers, cash-and-carry, or subscription.

A sponsor pays you for a portion of your site to offer their catalogs or to save webmastering hassles. Or else because they believe in what you are doing. Or have some vested interest in your content.

Ferinstance, on a community site, you would provide all the upcoming events and the local school lunch menus (including the knockwurst ala king and the butterscotch pudding au jus) for free, but charge the businesses \$12.00 per month and churches \$5.00 per month for their listings and content.

An advertiser simply pays you for a billboard on your site. The fees range from \$5.00 to \$40,000.00, depending on your site popularity. Or sometimes paid on a per-visit basis. There are also net co-op advertising exchanges.

With cash-and-carry, a user pays so much per download or per hour so he can access your premium material. A patent file, technical reprints, or a map repository are examples.

With a subscription, the user pays a flat fee for unlimited site access over a given time period. I personally feel a single time long term charge makes more sense than by the month.

Your users will not view this as an ongoing obligation. And the payment hassles are few and far between. \$25.00 for six month's access sounds about right to me. You must, of course, have both a "free" and a "fee" area.

Reliable and secure electronic petty cash isn't quite here yet. But you can certainly expect it within a year. For now, visa-over-voice-phone or use of fax still seems the best route.

Subscription or cash-and-carry, you will have to provide lots of no-charge stuff, including free trials. Make it completely obvious *exactly* what *all* of the charged stuff is all about. Through annotated descriptions or abstracts.

All of the above combos can be used. As might cooperative zero cash exchange promotions between sites. While zero net cash, these can very much boost your relevant traffic.

This Month's Contest

I've once again summarized some web resources for you in this month's sidebar. Many of these services and links mentioned can also be reached at my www.tinaja.com Last month's column appears as file RESBN55.PDF on the Resource Bin library shelf.

As our contest this month, just tell me a webmastering trick or technique I am not yet aware of. Or else tell me about your web experiences.

There will be a largish pile of my new Incredible Secret Money Machine II books going to the dozen or so better entries, plus an all-expense-paid (FOB Thatcher, AZ) tinaja quest for two that will go to the very best of all.

Send all your written entries to me here at *Synergetics*, rather than to *Nuts & Volts* editorial.

Let's hear from you. NV

Microcomputer pioneer and guru Don Lancaster is the author of 33 books and countless tech articles. Don maintains his no-charge US tech helpline found at (520) 428-4073, besides offering all of his own books, reprints, and consulting services. Don also has two free catalogs full of his resource secrets waiting for you. Your best calling times are 8-5 on weekdays, Mountain Standard Time.

Don is in the process of setting up his Guru's Lair at http://www.tina ja.com

Full reprints and preprints of all Don's columns and ongoing tech support appear here. You can reach Don at Synergetics, Box 809, Thatcher, AZ 85552. Or send any messages to his US Internet address of don@tinaja.com

BOUGHT & SOLD

Ailtech 360D11, Frequency Syn01-2GHz \$1,000	HP 8569B, Spectrum Anyz. High Performance \$8000
Arritsu ME645A, Microwave Radio Test Set \$1,000 Arbiter 1040A, Meter Calibrator,	HP 8620C, Sweeper Mainframe
AC/DC/Voltage/Current\$1,200	HP 86220A, RF Plug-in, 10-1300MHz
Argosystems AS210, Frequency Calibration System \$3,500	HP 86222B, RF Plug-in, .01-2.4GHz, Opt. 02, 04 \$1,200
Astrosystems A1202S/R01, Decade Syncro Resolver Standard S400	HP 86241A, RF Plug-in, 3.2-6.5GHz
Austron 2100, Loran C Timing Receiver	HP 86260A RF Plug-in 12 4-18GHz \$400
Bafco 911A, Frequency Response Analyzer (unused)\$400 Ballantine 1627A, Scope Calib., with/acc. heads \$1,000	HP 86290B, RF Plug-in, 2-18GHz w/Scales\$1,400
Bird 43 Wattmeter 7 Shins S400	UD OCAND Cional Con E 1024MUs Oct 1 2 C2 000
I Boonton 102 C/E, Signal Gen., 1KHZ-512MHZ	HP 8654A Sinnal Gen 10-520MHz AM/FM \$500
Boonton 250A, RX Meter, Same as HP	HP 8656A, Signal Gen1-990MHz AM, FM, HPIB \$2,000
10MHz-1.2GHz, HPIB\$650	HP 8660A, Synthesized Signal Gen., Opt. 5,100
Bruel & Kjaer 1612, Bandpass Filter\$250	HP 86602B, RF Plug-in, 1300MHz\$500
Bruel & Kjaer 2107, Frequency Analyzer	
Cushman CE24B, Frequency Selective Voltmeter \$800	HP 8821A Medium Gain Bank Amp \$150
Cushman CE-6030, Com. Monitor, with Opt. 05, HPIB \$5,000	IFR FM/AM-1000A, Service Monitor
DDC SR-400, Syncro Resolver Simulator\$400 DR Thiedig MILLI-TO2, Ohmmeter,	inition cool room in Amplinor
1 Milliohm-2 Terraohms	Keithley 169, Digital Multimeter
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ESI 296, Auto LCR Meter\$650	01_19GHz e1 000
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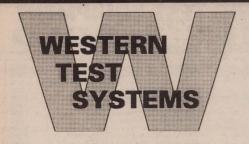
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W/TCXO ref., high res. audio HP 5340A-001,011 18 GHz Frequency Counter, \$1,9 CXO reference, HPIB HP 5340A-005,011 23 GHz Frequency Counter, HPIB HP 5340A-005,011 23 GHz Frequency Counter, HPIB HP 5342A-003,011 18 GHz Freq. \$2,4 HP 5342A-003,011 18 GHz Freq. \$2,7 Counter, +22 dBm, -20 dBm dynamic range,HPIB HP 5342A-20,04,05,11 24 GHz Frequency \$3,9 Counter, OCXO, DAC, and HPIB HP 5342A-225 MHz Frequency Counter \$2 STANDARDS HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz \$7 HP 5087A-031 Distribution Amp. \$1,2 0.1/1.0/5.0 MHz x4 channels each AUDIO & BASEBAND SPECTRUM ANALYZERS HP 8558A LF Section, 20 Hz-300 kHz \$5 DISTORTION ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling \$1,2 TEK DA4084 Programmable \$1,5 Distortion Analyzer, built-in low distortion osc. \$2,4 TEK DA4084 Programmable \$1,5	00.00 00.00 00.00 50.00 00.00
OCXO reference, HPIB HP 5340A-005,011 23 GHz Frequency Counter, HPIB HP 5340A-005,011 23 GHz Frequency Counter \$2,4 HP 5342A-003,011 18 GHz Freq Counter, +22 GBm_20 GBm dynamic range, HPIB HP 5342A-010,04,05,1124 GHz Frequency Counter, COXO, DAC, and HPIB HP 5382A 225 MHz Frequency Counter \$3,9 Counter, OCXO, DAC, and HPIB HP 5382A 225 MHz Frequency Counter \$2 STANDARDS HP 105A Quart Oscillator, 0.1/1.0/5.0 MHz HP 5087A-031 Distribution Amp, 0.1/1.0/5.0 MHz x 4 channels each AUDIO & BASEBAND SPECTRUM ANALYZERS HP 354A Distortion Analyzer, 514-600 KHz, -60 dB, auto nulling HP 334A Distortion Analyzer, built-in low distortion osc. \$2,4 TEK DA4084 Programmable Distortion Analyzer, 0.0025% THD, w/frame RMS VOLTMETERS	00.00 00.00 50.00 00.00
HP 5340A-005,011 23 GHz Frequency Counter, HPIB \$2,0 HP 5342A 18 GHz Frequency Counter \$2,4 HP 5342A-003,011 18 GHz Freq. \$2,7 Counter, +22 dBm, -20 dBm dynamic range, HPIB HP 5342A-01,04,05,11 24 GHz Frequency \$3,9 Counter, OCXO, DAC, and HPIB HP 5382A 225 MHz Frequency Counter \$2 STANDARDS HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz \$7 HP 5087A-031 Distribution Amp, \$1,2 0.1/1.0/5.0 MHz x 4 channels each \$1,2 0.1/1.0/5.0 MHz x 4 channels each \$1,2 0.1/1.0/5.0 MHz x 5 STANDARDS SPECTRUM ANALYZERS HP 8556A LF Section, 20 Hz-300 kHz \$5 STANDARDS SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling HP 339A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LE STANDARD SPECTRUM ANALYZERS HP 334A DISTORTION A	00.00 50.00 00.00 00.00
Counter, +22 d8m, -20 d8m dynamic range, HPIB HP 5342A-01,04,05,112 dFtz Frequency Counter, OCXO, DAC, and HPIB HP 5382A 225 MHz Frequency Counter STANDARDS HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz HP 5087A-031 Distribution Amp. 0.1/1.0/5.0 MHz x 4 channels each AUDIO & BASEBAND SPECTRUM ANALYZERS HP 8556A LF Section, 20 Hz-300 kHz DISTORTION ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling HP 339A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LEK DA4084 Programmable Distortion Analyzer, 0.025% THD, w/frame RMSVOLTMETERS	00.00
Counter, +22 d8m, -20 d8m dynamic range, HPIB HP 5342A-01,04,05,112 dFtz Frequency Counter, OCXO, DAC, and HPIB HP 5382A 225 MHz Frequency Counter STANDARDS HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz HP 5087A-031 Distribution Amp. 0.1/1.0/5.0 MHz x 4 channels each AUDIO & BASEBAND SPECTRUM ANALYZERS HP 8556A LF Section, 20 Hz-300 kHz DISTORTION ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling HP 339A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling LEK DA4084 Programmable Distortion Analyzer, 0.025% THD, w/frame RMSVOLTMETERS	00.00
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HP 5382A 225 MHz Frequency Counter \$2 \$TÄNDARDS HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz	50.00
HP 105A Quartz Oscillator, 0.1/1.0/5.0 MHz	
## 5087A-031 Distribution Amp	
0.1/1.0/5.0 MHz x 4 channels each AUDIO & BASEBAND SPECTRUM ANALYZERS HP 8556A LF Section, 20 Hz-300 kHz	
SPECTRUM ANALYZERS \$5 HP 8556A LF Section, 20 Hz-300 kHz \$5 DISTORTION ANALYZERS \$5 HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling \$5 HP 339A Distortion Analyzer, built-in low distortion osc. \$2,4 TEK DA4084 Programmable \$1,5 Distortion Analyzer, 0.0025% THD, w/frame RMS VOLTMETERS	
HP 8556A LF Section, 20 Hz-300 kHz	1000
DISTORTION ANALYZERS HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling \$5.4 HP 334A Distortion Analyzer, built-in low distortion osc. \$2.4 TEK DA4084 Programmable \$1,5 Distortion Analyzer, 0.0025% THD, w/frame \$1,5	
HP 334A Distortion Analyzer, 5 Hz-600 kHz, -60 dB, auto nulling \$1.4 Hz 339A Distortion Analyzer, built-in low distortion osc. \$2.4 TEK DA4084 Programmable \$1,5 Distortion Analyzer, 0.0025% THD, w/frame \$1,5 RMS VOLTMETERS	00.00
HP 339A Distortion Analyzer, built-in low distortion osc. \$2,4 TEK DA4084 Programmable \$1,5 Distortion Analyzer, 0.0025% THD, w/trame RMSVOLTMETERS	E0 00
Distortion Analyzer, 0.0025% THD, w/frame RMS VOLTMETERS	00.00
RMSVOLTMETERS	00.00
FLUKE 8920A True RMS Voltmeter \$7	
400 1470014 4011- 001111-	00.00
180 uV-700V, 10 Hz-20 MHz FLUKE 8921A True RMS Voltmeter, 180 uV-700 V, 2 Hz-2 MHz	00.00
HP 3400A RMS Voltmeter, 10 Hz-10 MHz\$2	50.00
OSCILLATORS	00.00
FLUKE 6011A Synthesized Generator,\$1,2 10 Hz-11 MHz, 10 Hz res.	00.00
HP 204D Oscillator, 5 Hz-1.2 MHz, 5VRMS, 80 dB step attenuator\$2 HP 209A Sine/Square Wave Generator, 4 Hz-2 MHz, 5VRMS max\$2	00.00
HP 239A Low Distortion Oscillator, 10 Hz-100 kHz\$4	
HP 652A Test Oscillator, 10 Hz-10 MHz	
ROCKLAND 5100 Synthesizer, 1 mHz-1.999999999 MHz	
5 Hz-500 kHz, 70 dB step atten., TM500	
TEK SG505-opt.01 Low Distortion Oscillator, \$1,0	00.00
MISCELLANEOUS	
HP 3575A-001 Phase-Gain Meter, 1 Hz-13 MHz, dual display	
HP 4437A Step Attenuator, 0-119.9 dB, S2 DC-1 MHz, 600 ohms unbal.	00.00
KROHN-HITE 3103 High/Low Pass Filter,\$5	00.00
10 Hz-3 MHz, 24 dB/octave KROHN-HITE 3202 / 3202R Dual	00.00
High-Pass/Low-Pass Filter, 20 Hz-2 MHz, 24 dB/oct	00.00
KROHN-HITE 3342 Dual HP/LP Filter,	
0.001 Hz-99.9 KHZ, 48 dB/octave KROHN-HITE 3750 LP/HP/BP/BR Filter,	00.00
0.02 Hz-20 kHz, 6/12/18/24 dB/oct. ROCKI AND 852 Dual Highpass/Lowpass Filter, 0.1 Hz-111 kHz\$1.0	

RF & MICROWAVE	
SPECTRUM ANALYZERS	
IP 11517A/18A/19A/20A Mixer,	\$675.00
12.4-40 GHz, w/adapters, for 8555A, 8565A, etc.	700
IP 11970AWR28 Harmonic Mixer, 26.5-40 GHz	
IP 11970Q WR22 Harmonic Mixer, 33-50 GHz	
IP 11970UWR19 Harmonic Mixer, 40-60 GHz	
IP 11971AWR28 Harmonic Mixer,	\$1,100.00
26.5-40.0 GHz, for 8569B IP 11971KWR42 Harmonic Mixer,	*4 400 00
18.0-26.5 GHz. for 8569B	\$1,100.00
18.0-26.5 GHz, for 85698 IP 5371A Modulation Domain Analyzer, 125 mHz-500 MHz	\$4,000,00
P 8406A Comb Generator,	
1/10/100 MHz increments, to 5 GHz	¥ 100.00
P 8444A Tracking Generator, 0.1-1300 MHz, for 8554B	\$800.00
IP 8444A-059 Tracking Generator,	
0.5-1500 MHz, for 8554,8568,etc.	
IP 8445B Preselector, 1.8-18.0 GHz, for HP 8555A	\$800.00
IP 8552B IF Section	\$750.00
IP 8553B RF Section, 1 kHz-110 MHz	
IP 8554B RF Section, 0.1-1250 MHz	\$800.00
IP 8554B/8552B/141T Spectrum Analyzer,	
100 kHz-1250 MHz, 100 Hz resolution	
IP 8555A RF Section, 0.01-18 GHz	\$950.00
IP 8555A/8552B/141T Spectrum	\$2,250.00
Analyzer System, 10 MHz-18 GHz, 100 Hz res.	
IP 8555A/8552B/8445/141 As above,w/1.8-18 GHz	\$3,000.00
Preselector, opt.003 freq.display	
P 8559A/853A Spectrum Analyzer,	\$5,500.00
10 MHz-21 GHz, 1 kHz res., w/display TEK 492-opt.1,2,3 Spectrum Analyzer,	
	\$8,000.00
50 kHZ-21 GHz, 100 Hz min.res.bw.	
EK 496P Spectrum Analyzer,	\$5,500.00
10 kHz-1.8 GHz, 30 Hz res., GPIB	
EK 7L13/7613 Spectrum Analyzer,	\$2,250.00
1 kHz-1.8 GHz, 30 Hz min.res.,w/frame	
EKTR502 Tracking Generator, 0.1-1800 MHz, for 7L12/13/14	
EKTR503 Tracking Generator, 0.1-1800 MHz, for 492/4/5/6	\$1,375.0
NETWORK ANALYZERS	\$375.0

HP 11590A-001 Bias Network, 1.0-18.0 GHz, APC7	\$450.00
HP 11664D WR28 Detector, 26.5-40 GHz, for 8755/6/7	\$750.00
HP 11665B Modulator, 0.15-18.0 GHz, for use with 8755/6/7 HP 11666A Reflectometer Bridge, 0.04-18 GHz, for 8755/8756	
HP 8407A/8412A/8601A Network Analyzer,	
100 kHz-110 MHz, with sweep generator	
HP 85050D APC7 Calibration Kit, for 8510 series HP 8505A-005/8503A Network An.,	\$1,100.00
0.5-1300 MHz, w/S-Parameter & phase lock	\$5,000.00
0.5-1300 MHz, w/S-Parameter & phase lock HP 8755Cr(3)11664A/182T Scalar	. \$1,750.00
Network An. w/3 detectors, 10 MHz-18 GHz & frame HP 8756A/(3)11664A Scalar Network Analyzer,	\$3 750 00
w/(3) detectors, 0.01-18 GHz	. 40,700.00
NARDA 7000A/7202/7206 Microwave	\$1,950.00
Multimeter System: scalar analysis 0.1-18 GHz WILTRON 640/G50/E(2)/7B50(2)	\$1.675.00
Network An., w/1-1500 MHz source, (2) log amp, (2) det.	. \$1,075.00
SIGNAL GENERATORS	
FLUKE 6060A/AN Synthesized Signal Gen.,	\$2,000.00
10 kHz-520 MHz, 10 Hz res, GPIB GIGATRONICS 600/10-18, opt.06	\$3,000.00
Synthesized Source, 10-18 GHz, 1 MHz res., OCXO, GPIB	
GIGATRONICS 875/50 Levelled	\$3,500.00
GIGATRONICS 875/86 Levelled Multiplier,	\$7,000.00
26.5-40.0 & 50.0-75.0 GHz outputs	
GIGATRONICS 910/12-18,opt6,14,	\$3,500.00
HP 85100V Frequency Mult.,	\$4,250.00
10-15 GHz in / 50-75 GHz out >0 dBm HP 8616A Signal Generator, 1.8-4.5 GHz	\$450.00
HP 8640B-001,002,003 Signal Gen.,	
0.5-1024 MHz, AM, FM, var. audio osc. HP 8654A Signal Generator,	
HP 8654A Signal Generator,	\$625.00
HP 8656A Signal Generator, 0.1-990 MHz, AM, FM, HPIB	\$2,900.00
HP 8660D/86603A/86632B Synthesized	
Signal Generator, 1-2600 MHz WAVETEK 907 Signal Generator,	\$1,000,00
7.0-11.0 GHz, 0 dBm levelled	\$1,000.00
SWEEP GENERATORS	
HP 8350A/86290B-H08 Programmable	\$5,500.00
Sweep System, 2.0-22.0 GHz, +4 dBm lvl'd HP 8600A Digital Marker, for HP 8601A	*400.00
HP 8601A Generator/Sweeper, 0.1-110 MHz, +20 dBm levelled	\$500.00
HP 8620C Sweep Oscillator Frame	\$550.00
HP 8620C-011 Sweep Oscillator Frame, HPIB programmable HP 86220A RF Plug-in, 10-1300 MHz, +10 dBm levelled	
HP 86222A RF Plug-In, 10-2400 MHz, +13 dBm levelled	\$1,375.00
HP 86230B RF Plug-in, 1.8-4.2 GHz, +10 dBm unlevelled	\$675.00
HP 86240A RF Plug-in, 2.0-8.4 GHz, +16 dBm levelled	\$1,000.00
2.0-8.4 GHz, +14 dBm lvld., 70 dB step att.	. \$1,200.00
HP 86241A-001 RF Plug-in, 3.2-6.5 GHz, +8 dBm levelled	
HP 86242D-004,008 RF Plug-In, HP 86245A RF Plug-in, 5.9-12.4 GHz, +16 dBm levelled	\$500.00
5.9-9.0 GHz, +10 dBm levelled	\$1,100.00
HP 86250D RF Plug-in, 8.0-12.4 GHz, +10 dBm levelled	\$675.00
HP 86260A RF Plug-in, 12.0-18.0 GHz, +10 dBm unlevelled HP 86260A-H04 RF Plug-in,	
10.0-15.0 GHz, +10 dBm unlevelled	
HP 86290A RF Plug-in, 2.0-18.0 GHz, +7 dBm levelled	\$1,750.00
HP 86290B RF Plug-in, 2.0-18.6 GHz, +10 dBm levelled	
WAVETEK 962 Sweep Generator,	
1.0-4.0 GHz, markers, +12 dBm univid.	
POWER METERS ANRITSU MP-81B WR10 Power Sensor,	\$1 975 00
75-110 GHz, for ML83A or ML4803A	
ANRITSU MP-82B WR8 (Pin Flange)	\$2,500.00
90-140 GHz Power Sensor BOONTON 42B/41-4E Analog Power Meter,	\$500.00
with 1 MHz-18 GHz sensor	
BOONTON 42B/42A-S3 Analog Power	\$375.00
Meter, with 1 MHz-8.4 GHz N(f) sensor GENERAL MICROWAVE 476/4240A	\$375.00
Power Meter & Sensor, 0.01-18 GHz, -35 to +10 dBm	
HP 432A/8478B Power Meter, 0.01-18 GHz, -20 to +10 dBm f.s HP 432C Autoranging Digital Power Meter, 10 uW-10 mW f.s	\$500.00
HP 435A/8481A Power Meter, 10 MHz-18 GHz,	\$1,000.00
-30 to +20 dBm	
HP 435A/8481H Power Meter, 10 MHz-18 GHz,	\$1,150.00
HP 435A/8482A Power Meter, 100 kHz-4.2 GHz,	\$1,000.00
-30 to +20 dBm	** 450.00
HP 435A/8482H Power Meter, 0.1-4200 MHz,	\$1,150.00
HP 435A/8483A Power Meter,	. \$1,000.00
0.1-2000 MHz, -30 to +20 dBm, 75 ohms	
HP 435A/8484A Power Meter, 10 MHz-18 GHz, -70 to -20 dBm HP 436A-022 Digital Power Meter, HPIB option	
HP K486AWR42Thermistor Mount,	
18.0-26.5 GHz, for 432 series	
HP Q8486A Power Sensor, 33.0-50.0 GHz, WR22, for 435/6/7/8 HP R486AWR28 Thermistor Mount,	
26.5-40.0 GHz, for 432 series	
RFMILLIVOLTMETERS	
RACAL 9303 TRMS Level Meter,	\$875.00
10 kHz-2 GHz, -77 to +23 dBm, GPIB AMPLIFIERS, MISCELLANEOUS	
AllTECH 7618E Noise Source,	\$500.00
10 MHz-18 GHz, SMA(m), 15 dB ENR	

BOONTON 82AD FM/AM Modulation Meter, 10-1200 MHz	\$800.00
HP 8447A-001 Dual Amplifier, 0.1-400 MHz HP 8901A-002,010 Modulation Analyzer,	\$500.00
150 kHz-1300 MHz, OCXO, int. cal.	\$3,300.00
M.P.D. LAB2-1020-2A Amplifier, 34 dB, 1.0-2.0 GHz, 2 Watts	\$800.00
M.P.D. LAB2-714-3A Amplifier, 34 dB, 0.7-1.4 GHz, 3Watts	\$800.00
MARCONITF2304 AM/FM Modulation Meter,	\$500.00
18-1000 MHz, FM dev.1.5-150 kHz	*225.00
MICROWAVE SEMI.CORP. MC5112 Noise Source,	\$325.00
COAXIAL & WAVEGUIDE	
Designation of the Party of the	
AMERICAN NUCLEONICS AM-432 Cavity	\$95.00
Backed Spiral Antenna, LHC, 2-18 GHz, TNC(f) *NEW* FXR/MICROLAB S3-02N Triple Stub Tuner, 200-1000 MHz,	****
100 Watts max., N(m/f)	\$90.00
GR 874-LTL Constant Impedance	\$450.00
Trombone Line 0-44 cm DC-2 GHz	
GR 900-Q GR900 14mm Interseries Adapters	\$125.00
HP 11691D Directional Coupler, 22 dB, 2-18 GHz HP 11692D Dual Directional Coupler, 22 dB, 2-18 GHz	\$800.00
HP 775D Dual Directional Coupler, 20 dB, 450-940 MHz	\$275.00
HP 776D Dual Directional Coupler, 20 dB, 940-1900 MHz	
HP 777D Dual Directional Coupler, 20 dB, 1.9-4.1 GHz	\$275.00
HP 778D Dual Dir.Coupler, 20 dB, 100-2000 MHz, N(m/f/f/f)	\$550.00
HP 778D-011 Dual Dir Coupler, 20 dB,	\$400.00
HP 8472A Crystal Detector, 10 MHz-18 GHz, neg. pol., SMA(m)	\$150.00
HP 8497K-004 Programmable	
Step Atten., 0-90 dB, DC-26.5 GHz, APC3.5	
HP 8498A-030 30 dB Attenuator, 25 Watts, DC-18 GHz, N(m/f)	
HP K422AWR42 Flat Broadband Detector, 18.0-26.5 GHz HP K532AWR42 Frequency Meter, 18.0-26.5 GHz	
HP Q752D WR22 Directional Coupler, 20 dB, 33-50 GHz	
HP R375A WR28 Variable Attenuator, 0-20 dB, 26.5-40 GHz	
HP R422AWR28 Flat Broadband Detector, 26.5-40 GHz	\$400.00
HP R532A WR28 Frequency Meter, 26.5-40.0 GHz	\$500.00
HP R914BWR28 Moving Load, 26.5-40 GHz	\$300.00
Directional Coupler, 10 dB, 40-60 GHz	\$450.00
NARDA 26298 20 dB Attenuator, 150 Watts, DC-1 GHz, N(f/f)	\$200.00
NARDA 3000-SERIES Directional Couplers	
NARDA 3090-SERIES Precision High Directivity Couplers	\$225.00
NARDA 369BNF / 369BNM High Power Termination, 175Watts, 0.7-18 GHz, N	\$325.00
NARDA 3752 Coaxial Phase Shifter, 0-180 deg./GHz, 1-5 GHz	\$1 250 00
NARDA 3753B Coaxial Phase Shifter,	\$1,250.00
0-60 deg./GHz, 3.5-12.4 GHz	
NARDA 4000-SERIES SMA Miniature Directional Couplers	\$75.00
NARDA 4203-6 Directional Coupler, 6 dB, 2-18 GHz, SMA(f/f/f)	
NARDA 4246B-10 Directional Coupler, 10 dB, 6-18 GHz, SMA(f)	
NARDA 5070-SERIES Precision Reflectometer Couplers	
NARDA 765-20 20 dB Attenuator, 50 Watts, DC-5 GHz, N(m/f)	\$135.00
NARDA 768-20 20 dB Attenuator, 20 Watts, DC-11 GHz, N(m/f)	\$125.00
NARDA 792FF Variable Attenuator, 0-20 dB, 2.0-12.4 GHz	
PAMTECH KYG1014WR42 Junction Circulator, 18.0-26.5 GHz SPACEK LARS DO-1 WR22 Flat Broadband Detector, 33-50 GHz	\$550.00
SPACEK LABS DQ-1 WR22 Flat Broadband Detector, 33-50 GHz TRGV510WR15 Precision	\$1,000.00
Rotary Vane Atten., 0-50 dB, 50-75 GHz	
TRGV551WR15 Frequency Meter, 50-75 GHz	\$600.00
TRGV559-10WR15 Directional Coupler, 10 dB, 50-75 GHz TRGW510WR10 Precision	\$400.00
Rotary Vane Atten., 0-50 dB, 75-110 GHz	\$1,000.00
TRGW551 WR10 Frequency Meter. 75-110 GHz	\$900.00
TRGW559-10WR10 Directional Coupler, 10 dB, 75-110 GHz	\$475.00
WAVELINE 822WR42 Precision	\$1,250.00
Rotary Vane Atten., 0-50 dB, 18-26.5 GHz WILTRON 60NF50 SWR Bridge,	******
5-2000 MHz, N(f) test port, 40 dB dir. WILTRON 87A50VSWR Bridge,	\$750.00
2-18 GHz, 35 dB dir., APC7 test port	De Co
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BOONTON 82AD FM/AM Modulation Meter 10-1200 MHz

HP 5005A Signature Multimeter	\$350.00
HP 8170A-002 Logic Pattern Generator,	\$1,200.00
2 MB/s, address driver option	
TEK 1240 Logic Analyzer, w/(36) 50 MHz channels	\$1,500.00

HP 4972A/18182A LAN Protocol Analyzer, with software	\$2,500.00
HP 59401A HPIB Bus Analyzer	\$700.00
TEK 1411R-opt.04 PAL Test Gen.	\$2,750.00
w/SPG12,TSG11,TSP11,TSG13,TSG15,TSG16	1 1 1 1 1 1 1 1 1
TEK 147A NTSC Test Signal Generator, with noise test signal	\$800.00
TEK 520A NTSC Vectorscope	\$1,200.00

MISCELLANEOUS	
P.A.R. 5202 Two-Phase Lock-in Amplifier, 0.1-50 MHz	\$3,000.00
P.A.R. 5205-94,95,96,98 Lock-In Amp,	\$2,750.00
P.A.R. 5206 Two-Phase Lock-In Amplifier, 2 Hz-200 kHz	\$2,500.00
P.A.R. 5208-92,94,97,98Two Phase Lock-In Amp.,	\$3,250.00
TEKTM5006 5000-series 6-slot Programmable Power Module	\$600.00
TEKTM503 500-series 3-slot Power Module	\$175.00
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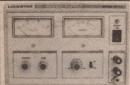
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Sometimes Two are Better Than One

ver the years, my experience in electronics — at both technician and engineer levels — has led me into a lot of work with sensors and transducers. That's why I was interested in the Speake Ltd. FGM-3 magnetic sensor, which we covered here last month. The US distributor of the sensor, Fat Quarters Software, provided me with a sample to evaluate. Discussions with the vendor showed me that the device could be used as a magnetic metal detector, especially for ferrous objects.

During my experiments, I wondered whether or not the resolution of such a detector could be improved. During a portion of my professional life, I worked on radar systems. Modern radars use a technique called monopulse resolution improvement (MRI), and I wondered whether or not the same method would work with other forms of sensor. I didn't have two magnetic sensors to experiment with, but I had some optical sensors, and if the method works for optical sensors, it will probably work with other sensors, as well. Before discussing the method, let's look at the idea of sensor resolution.

Sensor Resolution

Sensors designed to detect some other objects do not have a point-source detection zone, but rather they have a field of regard over which any object within it will be detected. Consider Figure 1A. Here we see a light sensor (e.g., photoresistor, photodiode, etc.) arranged to detect a light source (e.g., a light emitting diode) that travels along in front of the sensor over a straight line path.

If the sensor had an extremely thin field of regard, then we could know with high precision when the LED was opposite the sensor. At the instant of detection, its location would be well known. In real life, there is no such luck. A real sensor will have a detection profile more-or-less like Figure 1B. As the LED comes within the field of regard, the output voltage of the sensor begins to build, and will reach a peak when the LED is directly opposite the most sensitive spot on the sensor. After the LED passes this most sensitive point, the output voltage will begin to fall off until it is near zero again when the LED's path takes it out of the field of regard at the other end of the path. The output waveform is considerably fatter than the equivalent size of the "target" (i.e., our LED).

Hrumppffff! So what? The nature of the problem this causes is seen in Figure 2. Resolution, in this context, refers to the ability of the sensor to detect two identical targets spaced at some distance apart. The resolution is the smallest separation where the user of the instrument can detect the fact that two different targets are present, not just one. In the radar world, we use feet or meters between identical targets to define the resolution. Other sensors will use smaller or larger units, but the idea is the same.

In Figure 2, the problem is that the sensor field of regard is too broad to detect both

Open by Joseph J. Carr KAIPV Channel

targets. The circuitry will be designed to accept signals that are above some threshold, and reject signals below that level. In Figure 2, the

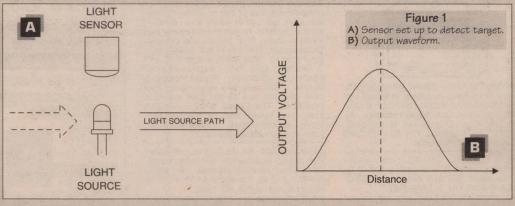
"response region" defines the area where signal detections are accepted, while the "insensitive region" is the portion of the curve where the circuitry will not see a detection. Note that both target "A" and target "B" are within the field of regard (which we called "beamwidth" in radar), so are seen as one target by the user ("apparent target" in Figure 7).

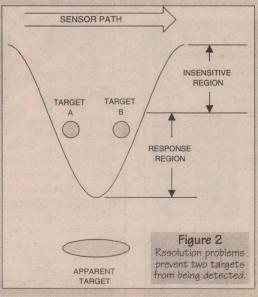
Now note what happens when the field of regard is narrower, as in Figure 3. In this case, no geometry of sensor and target will cause both targets to be within the field of regard at the same time, so the user of the instrument will see two distinct detections. Because the response of the sensor is not uniform as the sensor passes in front of the targets, the apparent targets will be a little oblong, rather than round.

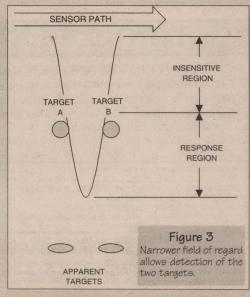
Whether your application is industrial process control, scientific instruments, or a

detector to find buried artifacts in a treasure field, the result of improving the resolution of your sensor system is that it becomes possible to: a) judge the approximate size of the detected object; and b) tell if more than one object is present.

Figure 4 shows how resolution is defined for most types of sensor. The curve shown is the "normal" curve, a.k.a. the "bell shaped curve." This curve is reasonably representative of many different forms of sensor, although be aware that it is not found on all sensors (some similar curve will be, however). Assume that we define the peak response as the zero decibel (0 dB) point. The resolution, which we will label "beamwidth" (BW) for lack of a better term, is the width of the curve at a point where the response drops off -6 dB for voltage (dBv) or current (dBi), or -3 dB for power. In the example of Figure 4, the peak response occurs at position X_o,







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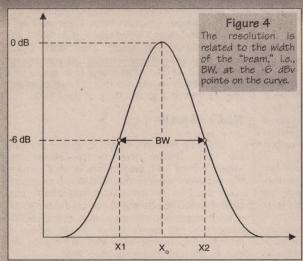


Figure 5 Sensor selected was a pair of OPT-101 devices. B-B **OPT-101**

while the beamwidth is the distance X2-X1. The detection circuitry is usually designed to respond only to signals that are larger than the -6 dBV level, i.e., signals that are produced when the target is positioned between X1 and X2 in Figure 4.

My Test Set-Up

Although my first inkling about using MRI to improve sensors other than radars involved speculating about using the Speake FGM-3 magnetic sensor in a high-resolution metal

detector, I didn't want to use that sensor because it was a bit costly. Although the FGM-3 prices are quite reasonable for magnetic sensors, it was simply too much to mess with on a hunch. So I decided to use an optoelectronic sensor. The sensor that I selected was the Burr-Brown OPT-101 device (two were needed). Figure 5 shows a block representation of the OPT-101. It comes in both SIP and eight-pin miniDIP forms. The pinouts shown in Figure 5 are for the eight-pin miniDIP version. When you see one of these devices it is like any other miniDIP chip, but is made of clear plastic, and has an optically sensitive region. The field of regard for this device is

The simplest connection for the OPT-101 is shown in Figure 5. Two

pins (3 and 8) are grounded, and one pin (1) is connected to a V+ source between +2.7 VDC and +36 VDC (I selected +12 VDC because that's what I have on the workbench). A 1 µF capacitor is used to decouple the V+ power supply connection. For the LED circuit I produced, the output voltage of the OPT-101 was from a few millivolts (mV) to +260 mV. Higher outputs would be realized with a stronger light source (mine was deliberately reduced in output). Also, by inserting a resistor between pins 4 and 5, one can increase the feedback resistance of the internal op-amp, and thereby produce more gain. With pins 4 and 5 shorted, as shown, however, the resistor is the internal 1 megohm value.

The next problem was to make a test jig. Two OPT-101 devices were mounted in an eight-pin IC socket on a piece of perforated board. When the OPT-101s are butted end-toend like that, the center-to-center distance between peaks was one centimeter (1 cm). That arrangement took care of the sensors, but what about the light source? The LED had to be pulled along in front of the sensors at the same distance from the plane of the sensor faces all along the path of travel. I solved that one from my ages-old junk box. I had an old AM car radio mechanical tuner. It had a worm gear drive and a dial. A small ruler (calibrated 0 to 100 mm, with large marks every 1 cm) was glued to the old dial scale. An LED and resistor were mounted on a small sliver of perforated board, and that assembly was soldered to the dial pointer of the car radio tuner. Neat, huh? I now had "micrometer-like" control over the position of the light source relative to the sensor assembly, and could measure the position by the markings on the ruler.

Figure 6A shows the geometry of my test set-up. The two OPT-101 sensors are designated OPTO-1 and OPTO-2, and they are spaced ΔX apart (in this case $\Delta X = 10 \text{ mm} =$ 1 cm). The LED was not as bright as it could be because it was connected to the +12 VDC source through a 1,200 ohm resistor, which means that 8 or milliamperes were flowing in the LED (15 mA is the normal brightness current for this LED).

The fields of regards of the two sensors are identical because the sensors were identical (plus or minus manufacturing tolerances). This produces the double-humped outputs shown in Figure 6B. The two curves represent the theoretical outputs of the two sensors as the LED was translated across their sensitive plane at a constant height. The curves should be identical in shape and amplitude. Note that the peaks occur at points equivalent to ΔX apart.

Several different detection schemes can be used to detect the position of the target. First, either curve can be used alone, and the peak noted. But this is not the best approach for the same reason that we don't radio direction find using the broad aspect of a loop antenna. The peak is simply too broad to be certain when the peak is present. We tend to have to go back and forth across it to settle on a good guess where the true peak is located.

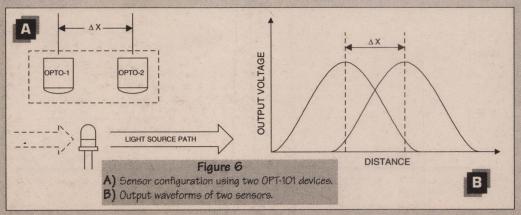
Another approach to detection is to sum the voltages and note where the dip between them occurs. Obviously, if we pass a peak and then come to a notch in the summed curves, then we have found the correct spot, right? Well, okay, I'll give that to you, but note in Figure 7 what often happens when the outputs of two sensors are algebraically added: the null tends to fill in and produce an actual null that is only marginally narrower than the peaks, if that.

So how do we solve the problem? Enter the Sensor Resolution Improvement (SRI) cir-

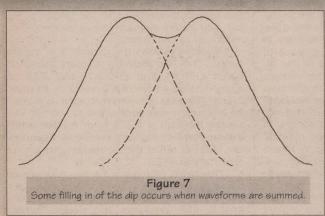
Sensor Resolution Improvement

The SRI circuit is devised from the radar world's MRI concept. In monopulse radar, the antenna is divided into two identical halves, each of which is fed the same transmit pulse. On receive, however, they work independently, to produce two signals (\$1 and \$2). These signals are processed to form new signals consisting of the sum (S1 + S2) and the difference (S1 - S2). The absolute value (i.e., without regard to polarity) of the difference signal - ABS(S1-S2) - is then created (a fancy radar guy's way of saying it was fullwave rectified). The system then takes the ratio (S1+S2)/(ABS(S1-S2)), and presents it. This resultant is a very much narrower version of the original. Two signals were, indeed, better than one in the resolution depart-

Figure 8 shows the signal processing flow needed to make SRI work for other forms of sensors (in this case, my optical sensors OPTO-1 and OPTO-2 are used). The outputs of OPTO-1 and OPTO-2 are labelled V1 and



Open Channel



V2, respectively. These signals are fed to inputs of a summing amplifier and a difference amplifier. The outputs of those amplifiers are V1+V2 sum signal and V1-V2 difference signal. The difference signal is fed through a fullwave precise rectifier (an opamp circuit), to create the ABS(V1-V2) signal. These signals are both fed to the inputs of an analog divider that produces an outout of:

$$V_o = \frac{K(V1+V2)}{ABS(V1-V2)}$$

(K is a scaling factor that all analog multipliers and dividers have.) If you plot the output voltage V_o against the position of the target, then you will note a startling increase in resolution. How much? Figure 9 shows the result of the trials with the LED. The data was collected, and then fed through an Excel spreadsheet to find the resultant seen in Figure 9.

You'll have to tune in next month to find out how it was done ... (I know, I'm a dirty rotten, #\$@#\$@, but there is only so much space and I have something else to cover.) In next month's column, we will take a look at some of the circuits that can be used to make this system work.

Save 144 MHz and 440 MHz!

The spectacular growth of the telecommunications and data communications industries has created a tremendous need for additional frequencies in the radio spectrum. Unfortunately, the spectrum is terribly limited. It seemed crowded when we had less complex telecommunications systems in the USA. A number of industry groups have proposed a reallotment of much of the spectrum to these new services. In the past, they overlooked the ham bands. But the success of one service in carving a little out of our 220 MHz band has led to an official position forcing reallocation of the 144-MHz (two-meter) and 440-MHz ham bands to other services.

Hams lost 11-meters to the Citizen's Band service in 1957 because they weren't using that band. They lost part of 220 MHz for the same reason, but two-meters, and to some extent 0.75-meters are well populated. We might lose those bands not for lack of use, but for the lack of political clout. The Federal Communications Commission was once populated with a large number of engineers,

many of whom were ham operators themselves. But today the "new FCC" is populated with lawyers ... and that's a bad sign for anyone at any time.

To make matters worse, there is a World Radio Conference coming up in 1997. The last WRC was 1979, and we gained some bands in the HF region in that one. But this time it might be different.

The official US position for WRC97 is being crafted at this time, and the FCC needs to know your opinion. Express it in three ways:

First, write your congressional representatives and let them know of your interest. Ask them to take a position in favor of ham radio and communicate it to the FCC.

Second, check into the FCC World Wide Web home page and give them your opinion (politely, of course, even if they are lawyers). The URL address is http://www.wrc97/gov

The third way to help is write or E-Mail the American Radio Relay League (ARRL) and

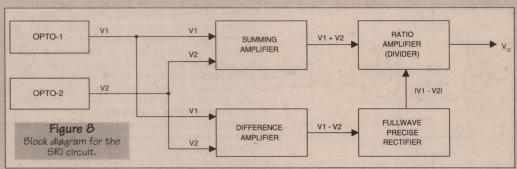
ask them how you can help. Their address is 225 Main Street, Newington, CT 06111 (tell them K4IPV sent you!). The ARRL E-Mail address is hq@arrl.org — In fact, if you are not a member of the ARRL, now is a good time to join, even if you are not a ham operator, you can lend a little clout of your own.

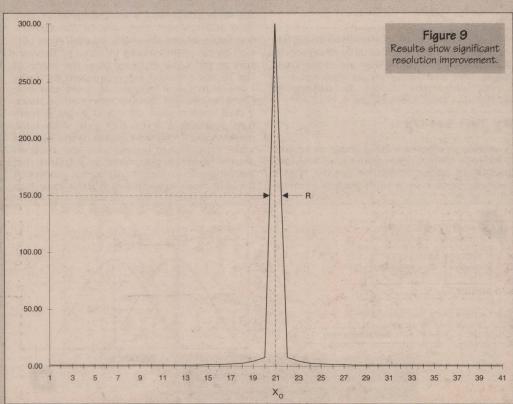
Next Month ...

In next month's Open Channel column, we will revisit the Sensor Resolution Improvement (SRI) project, show you some of the resultant waveforms, and some of the circuits used to do an actual trial of the concept. Sometime in the future, I hope to try it with magnetic sensors such as the FGM-3 devices. I still don't know if SRI works to improve metal detectors, but will try to find out in the months to come. **NV**

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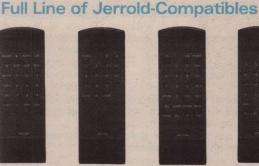


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he inductor, or coil, is the third of three passive components, the other two being the resistor and capacitor. Like the capacitor, it stores energy. Yet while the capacitor does so in electrostatic field, the inductor stores energy in an magnetic field. More importantly, an inductor acts to oppose any change in current. The ability to store energy and oppose a change in current (inductance) makes the inductor a useful component, one worth exploring.

Though not as widely used as capacitors, inductors, nonetheless, find extensive application in DC and AC circuits. In the former, we see them in solenoids, relays, speakers, the basic meter movement, and in electronic sensors. In AC circuits, they combine with other passive components to form simple low-pass, high-pass,

band-pass, and band-stop filters.

Figure 1b. Note, each type of inductor has a unique symbol.

Inductors, of course, exhibit inductance. The unit of inductance is the henry, symbolized by an H. In many applications, millihenries (Mh) and microhenries (Uh) are more common units.

Two characteristics determine the inductance of an inductor: (1) its permeability and (2) its physical parameters.

Coils are wound on either non-magnetic or magnetic materials. Non-magnetic cores consist of air, wood, copper, plastic, and glass. The permeability (ease with which a magnetic field can be established) of these materials is the same as for a vacuum. Magnetic materials are those such as iron, nickel, steel, or cobalt. These materials have permeabilities far greater than that of a vacuum. They are classified as ferromagnetic. Such cores provide a better path for the magnetic lines of force, thus permitting a stronger magnetic field.

Physical parameters, such as number of turns,

section, and travels out into the surrounding space, diminishing in strength. The magnetic field represents stored energy.

When current ceases to flow, the magnetic field that was established by it collapses back into the wire. This collapsing magnetic field induces a voltage into the wire. Thus, stored energy is returned to the circuit in the form of voltage, which, if the circuit is complete, will produce current.

If we wind a current-carrying wire into a coil, we have an inductor. The strength of the magnetic field surrounding that inductor is determined by the amount of current in the coil, the number of turns, how tightly the turns are wound, and the permeability of the core. All work to effect electromagnet field strength and the amount of inductance (L) in the coil.

Opposing a Change in Current Flow

An inductor has the ability to oppose a change

Inductors

by Ronald A. Reis

Store that Energy

In this article, we'll examine the physical characteristics of inductors, how energy is stored in a magnetic field, ways an inductor acts to oppose a change in current, and inductive reactance. We'll conclude by exploring resistive-inductive, RL, filters.

The Inductor: Coil that Wire

Inductors come in a wide range of shapes and sizes, as seen in Figure 1a. Some are tiny, no bigger than a 1/2-watt resistor. Others are larger than a baseball. Coils can be fixed or variable. They can be wound on an air core (actually a hollow cardboard tubing) or a core of powdered or laminated iron.

Schematic symbols for inductors are shown in

length, and cross-sectional area of the core, as shown in Figure 2a, also determine the inductance of a coil. Inductance is inversely proportional to core length and directly proportional to cross-sectional area. Furthermore, inductance is directly related to the number of turns squared. The mathematical relationship of these factors is shown in Figure 2b, where L is inductance in henries, N is the number of turns, μ is the permeability, A is the cross-sectional area in meters squared, and l is core length in meters.

Storing Energy in a Magnetic Field

The ability of a coil to store energy in a magnetic field is based on the principle of electromagnetism.

Every current-carrying wire produces a magnetic field that emanates from the wire's center, cuts through its cross-

Cross-sectional area, A

Core material

Number of turns, N

A $L = \frac{N^2 \mu A}{l}$ B

Figure 2

in current. If current is increasing, the inductor will act to oppose the increase. If current is decreasing, the inductor acts to oppose the decrease. In the former, current eventually increases to a constant value, but, because inductance is present, it will take longer to do so. In the latter, current inevitably decreases to a set value (or to zero), but again, will take longer to do so because of circuit inductance.

The role of inductance to oppose a change in current can also be summarized by Lenz's Law:

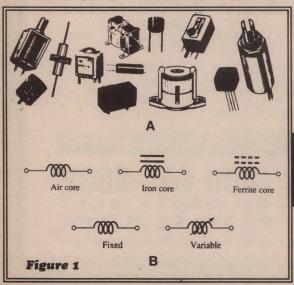
"When the current through a coil changes, an inducted voltage is created as a result of the changing magnetic field and the direction of the inducted voltage is such that it always opposes the change in current."

In other words, a changing current, producing a changing magnetic field (expanding or contracting), induces a voltage into the very coil in which current is flowing. This is called self-inductance. The induced voltage will create its own current in the coil. Polarity of the induced voltage, and thus direction of the resulting, "induced" current, will act to provide an opposing or aiding current to the original. When the original current increases, an opposing "induced" current is created. When the original current decreases, an aiding "induced" current is produced.

Lenz' Law is illustrated in Figure 3. Let's examine the figure closely.

In Figure 3a, the current is constant, being determined by battery voltage and resistor R1. (For purposes of explanation, we assume the coil has zero resistance.) Remember, a voltage is induced only if current, and the resulting magnetic field, is changing. If current is constant, the magnetic field, while present, is not changing. Thus in Figure 3a there's no induced voltage.

In Figure 3b, the switch is quickly closed. Total circuit resistance decreases (for resistors in parallel, total resistance is less than the lowest



• INDUCTORS • INDUCTORS • INDUCTORS • INDUCTORS • INDUCTORS •

resistor) and current attempts to increase. But as current increases, the magnetic field begins to expand. This expanding (changing) magnetic field induces a voltage, and thus current, that opposes the increase for an instant.

After switch closure, the rate of magnetic field expansion decreases (Figure 3c). Eventually current in the coil reaches a steady, constant value determined by battery voltage and total circuit resistance (Figure 3d). While the current is now greater than before the switch was closed, it took time to reach the new value.

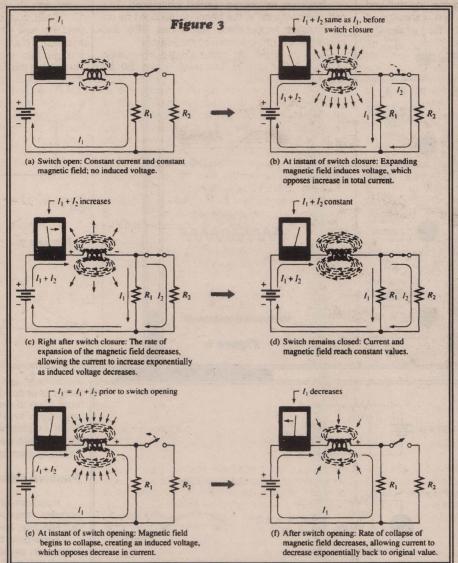
In Figure 3e, the switch is suddenly opened. With more resistance in the circuit, less current "wants" to flow. As current decreases, the magnetic field begins to collapse, inducing a voltage into the coil. The resulting "induced" current opposes this decrease. We say the "induced" current is aiding the original current, flowing in the same direction as it tries to help maintain the original current value.

Eventually, though, current decreases to the original value determined by battery voltage and R1 (Figure 3f).

To summarize: "The polarity of induced voltage is opposite battery voltage for an increase in current and aids battery voltage for a decrease in current."

Inductive Reactance

Opposition to current flow in a DC circuit is known as



resistance. It's measured in ohms. Opposition to current flow in an AC circuit is called reactance. It, too, is measured in ohms.

When an AC circuit has inductance, opposition to current flow exhibited by the inductor is known as inductive reactance. The letter X identifies reactance. When used in conjunction with the subscript L, the term X_L denotes a value of inductive reactance.

If alternating current flows in an inductor, that current is much less than the resistance of the coil windings alone would allow. Why? Because, as we saw above, current variations in the inductor, producing a varying magnetic field, causes voltage across the inductor (self-inductance) that at times opposes the applied voltage. The result is inductive reactance, the additional opposition to a sinewave alternating current.

The amount of X_L can be expressed by a simple formula:

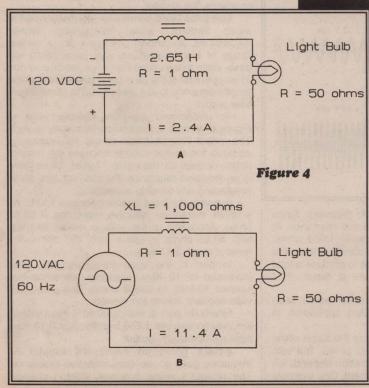
$$X_L = 2\pi fL$$

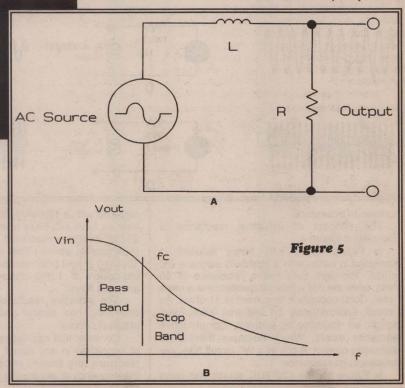
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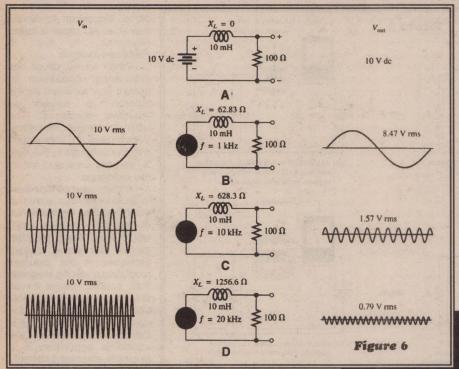
 X_L equals inductive reactance in ohms. 2π is a constant of 6.28 for a sinewave.

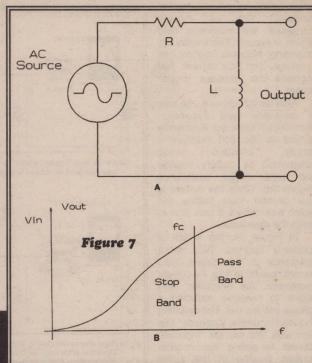
f equals frequency in hertz. L equals inductance in henrys.

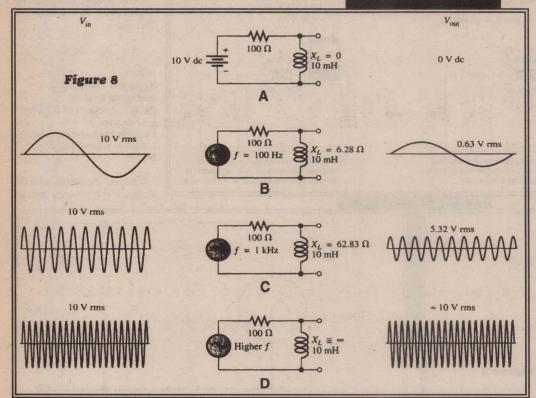
Note, X_L increases with an increase in frequency and an











increase in inductance.

The concept of inductive reactance is

illustrated in Figure 4.

In Figure 4a, a 2.65 henry inductor is connected in series with a light bulb across a DC source. The light bulb has a resistance of 50 ohms, while the coil winding's resistance is a mere 1 ohm. Total opposition to current is 51 ohms. As a result, approximately 2.4 amperes flows. Even though an inductor is present, no inductive reactance exists. Why? Because there's no alternating current. A coil in a DC circuit has only resistance, no reactance.

In Figure 4b, the 120-VDC source has been

replaced with a 120-VAC, 60-Hz generator. Since now we have AC, there is inductive reactance. At 60-Hz, inductive reactance equals 1,000 ohms. (See formula above.) That reactance is added to 50 ohms of light bulb resistance to produce a total opposition of 1,050 ohms. As a result, 11.4 amperes flows.

The inductive reactance effect we've been examining has almost unlimited application in practical circuits.

Consider that resistance has the same ohms of opposition in any circuit, DC or AC. But with reactance, the amount of opposition depends on the frequency of alternating current. One example would be to use X_L where it's desired to have high ohms of opposition to alternating current, but little opposition to direct current. Or, to use X_L for more opposition to a high-frequency AC compared to lower frequencies. We take advantage of these factors when we design simple passive filters with inductive elements. Let's take a look.

Passive Filters

Inductors are used with other passive components, resistors, and capacitors, to form simple, but effective, frequency filters.

A low-pass filter, consisting of an inductor and resistor in series across an AC signal source, is shown in Figure 5a. The output signal is taken off, or across, the resistor.

Low-pass filters permit only low frequencies to through while attenuating frequencies. As shown in Figure 5b, a certain range of low frequencies (the pass band) will pass, while those above the range are filtered out. The upper end of the pass band is identified as the cutoff frequency, fc. It's always 70.7% of maximum filter output.

To understand low-pass filtering, refer to Figure 6. In part a, the source frequency is zero. The output voltage, taken across the resistor, is equal to the source voltage because the resistor drops, or uses up, the entire 10 volts. Since there is no inductive reactance, the coil has zero ohms (neglecting any winding resistance).

In part b, the input frequency is now 1 KHz. As a result, there's an inductive reactance of 62.83 ohms. According to the voltage divider formula, with 62.83 ohms in series with 100 ohms, the

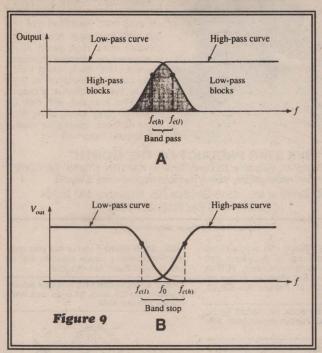
resistor drops 8.47 volts.

In part c, the input frequency has been increased to 10 KHz. Inductive reactance has jumped 10-fold, to 628.3 ohms. Now only 1.57 volts appears across the resistor.

Finally, in part d, with a 20 KHz input signal, the inductor exhibits 1,256.6 ohms. Just 0.79 volts appears across the output.

What's going on here? It's simple! As frequency goes up, so does inductive reactance. The resistor's value, however, doesn't change.

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Since the source voltage (10 volts) remains | produces an inductive reactance of 6.28 ohms.

constant, with each increase in frequency there's less voltage to be dropped across the resistor, the circuit output. Remember, in a series circuit, voltage drop is proportional to resistance. As the voltage across one resistive element increases (inductor, in this case), it decreases across the other resistive element (resistor).

A high-pass filter, also consisting of an inductor and resistor in series across an AC signal source, is shown in Figure 7a. Note, in this case, the signal is taken off the inductor.

High-pass filters, as you would expect, permit only high frequencies to pass through while attenuating lower frequencies. As we see in Figure 7b, a range of high frequencies (the pass band) will pass, while those below the range are filtered out. As with a lowpass filter, the cut-off frequency, fc, is 70.7% of maximum filter output.

Referring to Figure.8a, with a source frequency of zero, no inductive reactance is present and there's zero voltage across the inductor, and, thus, the output.

In part b, a 100 Hz source signal

The output voltage is 0.63 volts.

In part c, a 1 KHz source signal results in a 62.83-ohm inductive reactance. Output voltage is now 5.32 volts.

And in part d, with a high enough frequency from the source, inductive reactance reaches infinity. In that case, all the input voltage would appear across the output.

What we're seeing, of course, is the reverse of a low-pass filter. Essentially, the only difference between the circuits of Figures 6 and 8 is the point at which we take the output signal. For a low-pass filter, the signal is taken off the resistor. For a highpass filter, it's taken across the inductor.

By combining low-pass and high-pass filters, we create a band-pass filter, one that allows a certain band of frequencies to pass and attenuates, or rejects, all frequencies below and above the pass band (Figure 9a). Furthermore, a band-stop filter can also be created from low- and high-pass filters. In terms of response, this type of filter is essentially the opposite of a band-pass filter. It allows all frequencies to pass except those lying within a certain stop band (Figure 9b)

Too bad we don't have space to delve deeply into band-pass and band-stop filters, as well as numerous other inductor applications. But I think we've covered enough to give you the message. An inductor may be just a coil of wire wound on a core. Yet it has some remarkable uses. NV

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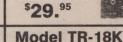
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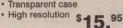
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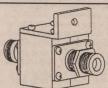
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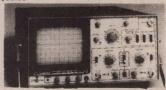
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Stamp Putting the Spotlight on BASIC Stamp Projects, Hints, and Tips Applications

This month, we're going to blast our way through two frequently requested applications: a circuit for connecting the BS2 to the phone line to take advantage of the DTMFout instruction, and an all-in-one LED

In BASIC for Beginners, we'll discuss binary-coded display for the BS1. decimal (BCD) numbers, which just happen to be useful for both DTMF and LED-display applications. But first, a word from our sponsor regarding Internet access to this column and

many of the goodies discussed

Apps on the Net

You can now retrieve past columns of Stamp Applications from the Nuts & Volts web site, www.nutsvolts.com - Click on the button for the library, and your browser will take you to ftp.nutsvolts.com and the directory /pub/nutsvolts/library. Go into the subdirectory /stampaps and you'll see a listing of Adobe Acrobat files for numbers 1 through whatever of this column (this is number

If you're interested in some of the Stamp-related products sold by my business, Scott Edwards Electronics, the information is just a couple of levels up on the same ftp site. The path is /pub/nuts volts/scott (of ftp.nutsvolts.com). Again, the files are in Acrobat format (now supported directly by many browsers) and include the current catalog, user's manuals to our most popular products, and samples of our AppKit documentation.

Now back to our regularly scheduled pro-

Connect BS2 to Phone Lines. Give the BS1 an LED Display

Project double-header for the BASIC Stamps I and II

Phone-Line Interface

The BS2 can generate telephone "touch" tones with its DTMFout instruction. However, you can't just hook the BS2 to the phone line and start dialing. There are several reasons:

· Audio from most sources, including the BS2, is single-ended (also known as ground-referenced). The signal voltage is the difference between a single wire and ground. Since phone lines travel far and wide - connecting locations with vastly different ground potentials and picking up noise along the way - they use differential signaling. A signal voltage is the voltage difference between the two wires that make up a phone-line pair.

> · Phone systems use the presence of a load on the line to determine when to go off hook and present a dial tone to local phone or other equipment and a busy signal to others trying to dial in.

· The phone line carries more than just audio signals; a direct-current (DC) supply voltage, a high alternating-current (AC) ringing voltage, and dangerous voltage spikes all show up on the phone lines.

· Despite its rugged signals, the Federal Communications Commission (FCC) considers the phone system to be tender as a newborn babe. They insist that their baby be protected from voltages that might confuse or hurt it.

For all those reasons, you need an interface to connect anything, including the BS2, to the phone line. If you're manufacturing a product, the FCC requires that you use an approved Data Access Arrangement (DAA). Because of the expense involved in getting regulatory approval for DAAs, they tend to be priced out of proportion to the cost of the components involved - \$25.00+ in small quantities. DAAs are also generally not available from the usual electronics parts outlets.

If you just want to experiment with the BS2's DTMF capabilities, a full-blown DAA is probably overkill, anyway. Using a simple circuit I found in the Encyclopedia of Electronic Circuits, Volume 5, by Graf and Sheets (TAB/McGraw Hill, 1995; ISBN 0-07-011077-8), I had a BS2 up and dialing in just a few minutes. My adaptation of the circuit

appears in Figure 1, complete with part numbers and sources.

The circuit meets all the basic requirements for a phone-line interface. transformer The converts singleended audio from the BS2 into differential signals. Sidactor and backto-back zeners on either side of the transformer clamp voltage transients to 280V on the phone side, and approximately 4.5V on the Stamp side.

Once you're hooked up, dialing the phone is simple. The DTMFout instruction just needs to know the BS2 pin number to use and a list of digits to dial. For example, to dial 555-1234 through pin 0:

DTMFOUT 0,[5,5,5,1,2,3,4]

DTMFout sends each tone for 200 milliseconds (ms) followed by 50 ms of silence. This gives the phone-company equipment plenty of time to detect and respond to the tones. However, you can also customize the durations of the tones and silences as follows:

DTMFOUT 0, <tone>, <quiet>, [st>]

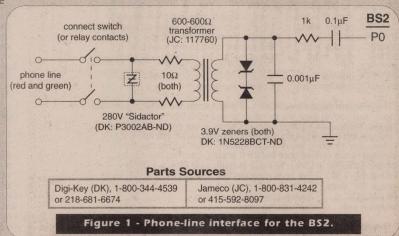
The tone and quiet times are in ms. So, to dial 555-1234 with 1-second tones and 0.5-second silences, you'd program:

DTMFOUT 0,1000,500,[5,5,5,1,2,3,4]

The obvious application for DTMFout is to build a super-deluxe telephone dialer. But don't forget all of the other services you can now control from the telephone keypad - like pagers, voice mail, menu routers, automated information systems, etc. For example, you could have the BS2 dial your pager if it detected an intruder in your home, or a burst pipe in your basement, or a crashed computer at your office. If you frequently access your bank's account-information system, you could automate the menu-walkthrough with the BS2.

Integrated LED Display/Driver

An earlier column on the MAX7219 LED dis-



Stamp Applications:

play driver (no. 10, Dec. '95; available from the Nuts & Volts web site as ST_AP10.PDF) drew a lot of interest.

The MAX chip makes it easy to talk to sevensegment LED displays up to eight digits long. Unfortunately, using the chip requires quite a bit of wiring to get all those LEDs connected.

Some readers asked for an integrated version of the MAX, incorporating LEDs and driver into a single module. I searched, and found the Three-Five System's TSM6755. Its display drivers aren't as fancy as the MAX7219 - you have to tell them specifically which LED segments to turn on - but the additional programming effort is more than balanced by the simplicity of the hardware. Figure 2 shows the complete hook-up

The TSM6755 module is available from my order desk as an AppKit with code on disk for the BS1, and PIC microcontrollers

> (Sources) and full manufacturer's docs. The module is available by itself from Farnell Components (1-800-718-1997).

Although the TSM6755 is easy to use, it shares one drawback with other LED displays: it draws a lot of current. With all 34 LEDs on at 10 milliamperes

(mA) each, the display can draw 340 mA. That's more than the 50-mA rating of the BS1's voltage regulator, or the 100-mA rating of the Counterfeit (BS1 kit; Sources).

There are at least three solutions to this problem: (1) Use a separate five-volt power supply with a rating of at least



The TSM6755 AppKit is available for \$25.00 from:

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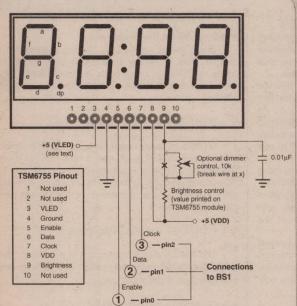
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Internet http://www.parallaxinc.com

500 mA. These show up in the electronics catalogs from time to time at bargain prices. (2) Use the circuit in Figure 3 to convert any 7- to 12-volt DC supply to regulated five volts for VLED. (3) If you're using a Counterfeit with its 100-mA regulator, you can test the module at reduced brightness. Just use a brightness-control resistor of at least triple the value printed on the TSM6755 module.

On the software side of things, the program listing that accompanies this column tells the tale. The subroutine IIIV write converts a 16bit value between 0 and 9999 into decimal digits; converts the digits into patterns of LEDs corresponding to the individual numerals; and sends these patterns to the TSM6755 by a clocked (synchronous) serial method. The subroutine allows the program to specify the location of a decimal point, and to determine whether or not to blank leading zeros. Leading zeros are the implied zeros to the left of the first non-zero digit of a fixed-length display. Huh? Okay, suppose you want to display "123" on a four-digit display. Do you want to



Note: The two digits on the lefthand side of the display have additional vertical segments to form a + sign in the center of the digits. However, in the TSM6x55 modules, these segments are not fitted with LEDs, and will not light.

Figure 2 Connecting the TSM6755 LED display module.

Listing 1: Using the TSM6755 with BS1

'Program: IIIV_LED.BAS

'This program controls TSM6x55 LED displays. It demonstrates the basics of communicating with the TSM6x55s by displaying the value of a counter on the LEDs.

Hardware interface with the TSM6x55:

SYMBOL DATA_n = 1
SYMBOL DATA_p = pin1
SYMBOL CLK = 2
SYMBOL Enable = 0

Bits are shifted out this pin # to display.

'Data valid on rising edge of this clock pin. 'Activates TSM6x55 to accept data.

SYMBOL initLED = bitO

SYMBOL colon = bit1 SYMBOL bitPat = b1

SYMBOL clocks = b2

SYMBOL digit = b3 SYMBOL DP = b4 SYMBOL DP = b4 SYMBOL dispVal = w3 SYMBOL counter = w4

Flag to trigger initialization of display.
Flag to turn on colon in the middle of display.
Byte to be sent to the display.

Bit counter used to clock out bits.

Loop counter for clocking out bytes.

Position of decimal point Value to be displayed on the LEDs.

'Counter for demo. The program begins by clearing all pins except the TSM6x55 enable to 0, and setting their I/O direction to output. It then writes 0s

to the TSM6x55's internal registers to clear out any garbage that

may be left from a previous write to the display. If this were not done, any 1s left in the display registers would cause the display to load newly arrived data before it should, garbling the display. let port = \$FF01 'Dirs = \$FF (all outputs) and Pins = 1 (low). et initLED = 1 'Set up to initialize the display. let port = \$FF01

let initLED = 1 gosub IIIV_write let initl FD = 0 let colon = 0

'Clear display registers.' Switch to normal operation.' Turn off the colon.

'The decimal-point variable serves three functions: (1) If it's

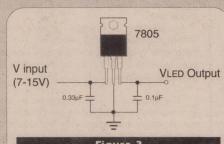


Figure 3 Voltage regulator for VLED.

in the range of 0 to 3, it turns on the decimal point of the corresponding digit (numbered right to left). (2) If it's in the range of 1 to 3, it also turns off leading-zero blanking of the display, allowing numbers like "0.123" to be displayed. (3) If DP is 0 or greater than 3, leading-zero blanking is on, so that numbers like "26" don't display as "0026." When DP is 0, the rightmost decimal point is on _with_ leading-zero blanking, so that single-digit numbers display like so: "4." For our demo, we'll turn off the decimal points and turn on leading-zero blanking.

let DP = 255

'No decimal point; no leading zeros.

MAIN PROGRAM LOOP

Now that the display is properly initialized, we're ready to send it data. The loop below increments a 16-bit counter and displays the lower 4 digits on the TSM6x55. A subroutine takes care of converting

the value into digits, suppressing leading zeros, and sending data

to the display.

let dispVal = counter gosub IIIV_write let counter = counter+1 pause 200

Copy counter into dispVal. Write it to the display. Increment the counter.

Slow things down a bit.

Do it again.

TSM 6x55 SUBROUTINE

'This routine converts the value in dispVal into a series of bit 'patterns that light up the appropriate LEDs on the display to 'show individual digits. It sends these bytes to the display via 'clocked (synchronous) output in which a bit is placed on the data 'line, then the clock line is pulsed to shift that bit into the display. IIIV_write:

low enable. high DATA n pulsout CLK,1

' Activate the display.
' Send a start bit (1) to the display.
' Clock out the start bit.

Stamp Applications:

see "123" (leading zeros blanked) or "0123" (leading zeros allowed)?

Apart from the leading zeros, extracting the digits of a number is straightforward. Just take the remainder of the division by 10 (value // 10 in PBASIC) to get the ones-place digit. Then divide the value by 10 and do it again. I point this out because digit-extraction is a common requirement for driving displays. It's important enough that Parallax has added a digit-extraction function called DIG to the BS2.

BASIC for Beginners

In past columns, we discussed several common numbering systems: decimal, binary, and hexadecimal. If you recall, hexadecimal (or hex, for short) is the preferred shorthand for binary, since each digit neatly expresses a nibble (four bits). Hex digits range from 0 through F, representing decimal values of 0 through 15.

There's another important application for hex numbers: providing a reasonable way for a computer to store and manipulate decimal digits. Hex numbers used this way are referred to as binarycoded decimal (BCD).

The idea is this: instead of using hex numbers' full range of values (0-15), restrict them to the values that overlap with the decimal system (0-9). Each hex, excuse me BCD, digit fits neatly into four bits. When it's time to display those digits, no conversion is required. Just put the value of each BCD nibble on the corresponding digit of the dis-

Contrast that to the approach required for converting pure binary values into decimal digits. The one's digits is extracted by taking the remainder of division by 10. The binary value is then divided by 10 and the one's digit extracted again until all digits are done.

We take built-in division capabilities for granted with the Stamps, but division is difficult for lim-

ited micros. If division can be avoided, the resulting program will be smaller and faster than one that uses division.

How do you decide when BCD might be the better way to go? This depends on your application, and which Stamp you're using. The BS1 does not have nibble variables, so its ability to deal with BCD is limited.

The BS2 does have nibble (nib) variables, and the ability to define arrays - variables with multiple cells that can be addressed by providing an index value. It has hex format modifiers for debug and serout instructions that also work for displaying BCD digits.

Listing 2 is a BS2 program that demonstrates some basics of BCD by implementing a counter that counts from 0

```
This program implements a counter to demonstrate the basics of working with binary-coded decimal (BCD) numbers with the BS2.
  Variables used in the program.

3CDcnt var nib(6) 'Six-digit BCD counter.
var nib 'Index into array of BCD digits.
nonZ var bit 'Flag used in blanking leading zeros.
BCDcnt var
nonZ
'You can enter a starting value for the counter by changing the 'digits stored in BCDcnt below.

BCDcnt(5) = 0: BCDcnt(4) = 0: BCDcnt(3) = 0

BCDcnt(2) = 0: BCDcnt(1) = 0: BCDcnt(0) = 0
                       Main Loop of Demonstration Program
  All of the real action takes place in the subroutines—all this
  main program loop does is to display the BCD number, increment it, and repeat. Forever.
  gosub BCD_display
gosub BCD_inc
                                                ' Show the BCD number.
                                                'Add 1 to the BCD number.
goto Main
                                                 'Do it again.
                            BCD Subroutines
        ===BCD_display
  This routine displays the six BCD digits stored in BCDcnt() on
 the screen using the debug instruction. It incorporates logic to eliminate leading zeros (i.e., to display values like "001234" as "1234"). A zero is defined as a leading zero if it is to the left of the first non-zero digit. We also have to consider the case in which _all_ digits are zero; here we don't want to blank the zero in the ones place.
  In the routine, a flag (nonZ) is set to 1 by the first non-zero digit. The if/then instruction right below the label "skip1" takes care of the rest. It says, "if this digit is 0, AND the first non-zero digit hasn't come yet, AND this isn't the ones place, then don't print this digit."
BCD_display:
nonZ = 0
for i = 5 to 0
                                                                'Turn on leading-zero blanking.
'For each digit..
     if BCDcnt(i)= 0 then skip1
                                                                 'In effect: "if digit is not 0...
     nonZ = 1
                                                                 '..then nonZ = 1
                                                                  See comments above for next line.
     if BCDcnt(i) = 0 and nonZ = 0 and i <> 0 then skip2 debug hex1 BCDcnt(i) 'Place the digit or
                                                                  Place the digit on the screen.
  skip2:
                                                                  Process all 6 digits.
  next
                                                                 'Send a carriage return.
  debug cr
                                                                'Return to program.
 return
  ====BCD_inc
This subroutine increments (adds 1 to) the 6-digit BCD number stored in BCDcnt. It works like grade school arithmetic: add 1 to the ones
  digit. If the result is 10, put zero into the ones digit and carry the 1 to the 10s digit. The routine uses a loop to carry the 1 all
  way up to the 6th digit when the number to be incremented is 999999.
BCD_inc i = 0
                                                                'Start at lowest digit.
  incLoop:

if i > 5 then done

BCDcnt(i) = BCDcnt(i) + 1

if BCDcnt(i) < 10 then done
                                                                  Stay within the 6 digits, 0-5.
                                                                  Increment digit.
If no carry-the-one, we're done.
Otherwise, zero this digit..
     BCDcnt(i) = 0
                                                                    and increment the next digit.
      i = i + 1
                                                                   Keep going until out of digits...
   goto incLoop
 done:
                                                                     or no more carries.
                                                                 'Return to program.
return
```

999,999 on the debug screen. That points out yet

another handy use of BCD: to construct counters

that exceed the limits of built-in word (16-bit) vari-

Listing 2: Demonstrating BCD Counting with the BS2

Program: BCD_DEMO.BS2

```
'Get the ones digit of dispVal.
en noZblank 'No 0 blanking if 0 < DP < 4.
     bitPat = dispVal//10
     if DP < 4 AND DP > 0 then noZblank
'The line below blanks 0s in the leftmost positions of the display.
'A digit is defined as a leading zero by these three rules:
'(1) The digit must be a 0 (bitPat = 0), and
'(2) It must be the leftmost digit of the number (dispVal = 0), and
'(3) It must not be the only digit of the number (digit<>0).
'If a digit meets these three rules, then it is changed from 0 to blank.
if bitPat = 0 AND dispVal = 0 AND digit <> 0 then blank
noZblank: 'Look up LED pattern matching digit value.
lookup bitPat,(111,40,93,124,58,118,119,44,127,62),bitPat
'To turn on the decimal point for a particular digit, its bit pattern 'must be ORed with 128 (%10000000). So the two lines below are meant 'to work as: "If DP = digit then bitPat = bitPat OR 128." The backwards 'logic is required because PBASIC can only go to a label as the 'outcome of if/then. Below are several more examples of negative
```

Convert and send digits 0 - 3 (1s to 1000s).

```
logic to work around this limitation.
 if DP <> digit then skip0
bitPat = bitPat | 128
                                       'Turn on bit 7 of LED (decimal point).
skip0:
  if initLED = 0 then skip1 "If initLED = 1 then bitPat = 0"
blank:
  bitPat = 0
                                                      ' Send all zeros to display.
skip1:
 for clocks = 1 to 8
let DATA_p = 0
IF bitPat < $80 then skip2
                                                      'Send eight bits.
'If msb of bitPat = 1, then let
                                                      "...DATA_p = 1, else DATA_p = 0.
    let DATA_p = 1
                                        Pulse the clock line.Shift bitPat one bit to the left.Continue for eight bits.Divide dispVal by 10 to get next digit.
    pulsout CLK,1
    let bitPat = bitPat * 2
next clocks
dispVal = dispVal/10
next digit
DATA_p = colon
pulsout CLK,1
pulsout CLK,1
                             Now put the colon bit on the data line.
                            And clock it out to the display.
This takes 3 clocks; 2 for the LEDs of
                            the colon, plus 1 to load the display.
Disable the TSM6x55.
Return to program.
pulsout CLK,1
```

high enable

return

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Questions & Answers

TECH FORUM

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QUESTIONS

I would like information on the unique electrical and mechanical characteristics of the electrical switch gear (switches, relays, contactors, etc.) used in diesel-electric locomotives. I found an oblique reference to this unique equipment in an old locomotive maintenance manual, but I haven't been able to locate any details. Any pointers to existing literature or references on these subjects would be greatly appreciated.

9961 Joseph F. Richmond Joppa, MD

I'm looking for some information on an electronic timing advance to replace a mechanical centrifugal advance. This electronic device should replace the distributor completely. Newer GM vehicles have such features. I'm trying to replace the distributor for '79 Honda Civic CVCC and '88 Toyota Tercel. Is this possible? Where can I get more information?

9962

James Lee via Internet

Is it feasible to modify digital clocks so that the hour or minutes can be set back individually? There are some clocks that have reverse setting, but they affect hours and minutes simultaneously.

9963

Ramon Eller Pittsburgh, PA

I need information on higher power laser beams. I do not understand the difference from 1 MW to a 100 MW. Could you please educate me? Where can I purchase such products?

9964

Brent Wessock Las Vegas, NV

I have a Professional Computer-Basic made by Texas Instruments. I

have two problems.

First I need to find the bootup disks and documentation. It uses two diskettes to start up, and I have lost them along with the owner's manual. I have tried Radio Shack repair, but they don't have an idea where to get them. My second problem is getting beyond the error codes. How do I find out what the error codes mean?

9965

Ray Kuharec Bridgeport, CT

I have searched quite a while for a simple solution to a problem. I have a PC controlled stepper motor driver system (Arrick Robotics) that controls two size 23 five-volt, one-amp, six-wire stepper motors. I would like to replace

one or two of these motors with similar motors with more torque. These type of motors usually have higher current needs, such as three to five amps and sometimes lower voltage. Can this be done using a separate power supply and perhaps opto isolators and Darlington transistors? Can anyone supply a schematic? I don't have an EE!

9966

Nicholas I. Oshana Jr.

Bristol, CT
I have a small 8-cylinder gas

engine. Is there some way for me to activate an auxiliary electric cooling pump when the engine drops below 1000 RPM?

9967

Bob DeVoe Grosse IIe, MI

Do you know of a low-cost, tunable, line powered, 88-108 MHz telephone transmitter with a range of about 100 feet maximum? Can you show me a circuit of how to build one?

9968

J. Carroll Denver, CO

ANSWERS

ANSWER TO #8962 - AUGUST 1996

There were two questions actually asked here. To solve the power problem for the walkman, you can purchase what you need at Radio Shack. You can purchase the following units depending on the current requirements of the walkman:

Part# Current rating
270-1560 up to 300 mA
270-1562 up to 400 mA
273-1801 up to 1 amp

Depending on the brand of walkman, you may also need to purchase 270-1563 for the currect size adapter. To solve your reception, you can use a longer piece of audio cable coming out of the headphone jack on the walkman. The audio cable is actually the antenna on a walkman. You will get better reception if you have the wire horizontal and not vertical. Also, the higher the cable is from the ground, the better reception you'll get.

Mike Dunmire Lemoyne, PA

ANSWER TO #89611 - AUGUST 1996

There are a few places where you can still find parts and software for the Tandy Coco3. The place that I'd recommend first would be Radio Shack, since Tandy is the manufacturer. They have a list of software available from Ft. Worth. I can guarantee that you'll have to special order the items, but they may be able to ship directly to your address and save you a return trip. You can also try 1-800-THE-SHACK and if

you have internet access, they can be found at Tandy.com

The other place I would recommend is *PCM Magazine*. They usually have a good selection of retailer advertisements for older Tandy computers.

I am pretty sure that you will not be hooking a CD-ROM up to it, however check your local Radio Shack for those data cassettes for the CCR-81, they may have some at discontinued prices, or may be able to special order them

> Mike Dunmire Lemoyne, PA

ANSWER TO #8968 - AUGUST 1996

The HTX-100, as stated, is a transceiver, meaning that it can transmit and receive. Because it takes considerably more power to transmit than receive, the display blanks out after a few seconds to conserve vital battery power. This feature can also be found on the TRC-225 walkie-talkie. The original question doesn't state if this occurred from the beginning or is a developed "problem," so I assume that this is the correct answer to the question.

Mike Dunmire Lemovne, PA

ANSWER TO #8963 - AUGUST 1996

Dongles are hardware security keys which plug into the parallel port of a computer and allow a particular piece of software to operate. Typically, when you start the software application, the security key is interrogated and information is read back into the program. The information may be a single byte or several hundred bytes and must match exactly the same data generated by an algorithm within the software. For example, a specific pseudo random code that is millions or billions of bits long can easily be generated with a few flip-flops and gates in the key and a few lines of code by the software. When the program is invoked for the first time, the key is initialized to the first bit in the enormously long, but known sequence. At various times, the software application will interrogate the key again and the next byte(s) in the sequence are sent and compared with that generated by the software. If at any time there is a discrepancy, you will be unceremoniously exited from the application. Often, after the application is initially started, the key can actually be removed since it is common to only interrogate the key at startup and then anytime thereafter that you try and save your work. Obviously, the key would have to be re-inserted prior to saving.

Since each key may work differently from any other, it is difficult to build a "universal dongle" for all appli-

ANSWER INFO

 Include the question number that appears directly below the question you are responding to.

you are responding to.
• Payment of \$25.00 will be paid within four weeks of publication if your answer is printed.

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 Due to space limitations, we can not reprint the original questions with the answer. The question number and the issue it appeared in are printed above the answer.

QUESTION INFO TO BE CONSIDERED FOR PUBLICATION

All questions should relate to one or more of the following:

1) Circuit Design 3) Problem Solving

2) Electronic Theory 4) Other Similar Topics

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Be brief but include all pertinent information. If no one knows what you're asking, you won't get any response (and we probably won't print it either).

 Write legibly (or type). If we can't read it, we'll throw it away.

 Include your Name, Address and Phone Number. Only your name will be published with the question, but we may need to contact you.

cations. There are, however, some things which you can do depending on whether the key is available (perhaps you can borrow one temporarily) or not. The following methods may provide varying degrees of success:

(1) First disassemble the software (using, say, DEBUG) and attempt to rewrite the code to jump over that part which accesses the protection routine. This will require a fair amount of knowledge of assembly coding, port addresses etc.

(2) Attach a logic analyzer to the key and monitor what it is doing. You

TECH FORUM

may not be able to determine the sequence that it is using. But you will be able to record the information from, say, the first dozen interrogations. A simple ROM based circuit could then defeat the key. This approach will only prove useful when a single byte or a few bytes are sent per activation of the key (which must also be available.

(3) Purchase a piece of software specifically designed to eliminate the need for the key. Some companies such as Safesoft Systems Inc. (204)669-4639, write software which will work for individual applications for people who have "lost" their own legal key.

The software works in one of two ways. The first type is run only one time during which it will permanently rewrite your original application to eliminate the need for the key. The second type must be activated prior to invoking your program which it does not modify. It simply remains resident in your computers memory and emulates the operation of the hardware key. If they have a "softkey" for your particular application, you can buy it for about \$200.00, but it is not universal.

(4) You can purchase the "Hard Lock Analyzer" from A-GO Technologies Inc. (604)736-0588 or E-Mail: mchung@iceonline.com for about \$200.00. It is a hardware board (7" by 4.3") which plugs into a PC slot in your computer and can acquire up to 64K bytes of sequence information.

With the original key attached, you place the board in "learn" mode. Afterwards, it will operate in emulate mode. Depending on the number of bytes that the key is required to supply each time that it is invoked will determine how many saves you will be able to do before having to restart the application and initialize the sequence.

The legality of any of these methods is an open question, but not, I'm sure, to the software manufacturers! In any event, these software keys are definitely a nuisance especially when you have half a dozen or more to contend with. Some can be "stacked" end-to-end while others simply refuse to operate unless they are the sole proprietor of the port. It is for this reason that many software manufacturers seem to be moving away from these keys to protect their software.

Jay Rabkin Pasadena, CA

ANSWER TO #89610 - AUGUST 1996

How simple converting a two meter rig to 100 MHz less for both transmit and receive depends on the rig. Generally speaking, it's not practical because of the great frequency change. Going from 144 to 44 would require changing all the non-IF tuned circuits, and assuming it's not a synthesized system, swapping crystals. Synthesized rigs would be even rougher since there's computer code to go along with it. It's not something I'd recommend trying.

recommend trying.

Question 2, on changing the scan mode from 1-400 to 400-1, this would rquire a change to the operating code for the synthesizer of the Pro 2004. If you could obtain the necessary chips (blank) hack the code, rewrite it, burn a new chip, and re-install it successfully, you could probably do it. But whether it

scans up or down, it's still going to find the active channels in the same amount of time.

> Richard M. Nelson Newport News, VA

ANSWER TO #8961 - AUGUST 1996

Although your problem may be a cable as you suggest, I would first make sure that the printer select switch on the computer is set correctly. Unlike "real" Apple II computers, the Laser 128 has both parallel and serial ports built-in. Either one can be used as the standard printer port (i.e., accessed via the PR#1 command), therefore there is a switch on the unit to select which port you use to connect to your printer. Since you stated that the computer hadn't been used for a while, I'm betting someone changed the setting of this switch.

Jeff Mazur

ANSWER TO #8964 - AUGUST 1996 Cordless Telephone Frequencies

Chan	555 161	Base	Handset
1		43.72	48.76
2		43.74	48.84
3		43.82	48.86
4		43.84	48.92
5		43.92	49.02
6		43.96	49.08
7		44.12	49.10
8		44.16	49.16
9		44.18	49.20
10		44.20	49.24
11		44.32	49.28
12		44.36	49.36
13		44.40	49.40
14		44.46	49.46
15		44.48	49.50
16		46.61	49.67
17	В	46.63	49.845
18	C	46.67	49.86
19	-	46.71	49.77
20	D	46.73	49.875
21	A	46.77	49.83
22	E	46.83	49.89
			49.93
23		46.87	
24		46.93	49.99
25		46.97	49.97

Channels A-E handset frequencies are also used for baby room monitors and low-power handie-talkies.

Effective June 5, 1995, channels 1-15 are authorized and the original 10 channels are renumbered 16-25.

Jerome Kowalewski Milwaukee, WI

ANSWER TO #89613 - AUGUST 1996

Modems, internal or external are turned on by communications software. I like Quick Link II which I use for both FAX and data. There are lots of others, but Quick Link II has automatic search for the port, IRQ and address that the modem is set to. Also, it is menu driven and almost trouble free. You may have to experiment with modem speed until you get garbage and back down to the last that worked well, if you don't know the speed.

Travis Brackeen St. Louis, MO

ANSWER TO #8965 - AUGUST 1996

You want to build a thyristor controller with output that responds to a DC input voltage from a sensor. This could be done digitally with a counter and comparator, or using some kind of reasonably fast microcontroller design.

Continued on page 117

The filter company

Positive Notch Filters

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Positive "G" Series Filters

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Tom G. - FL Excellent Service, fast and quick. The only bad thing about this company is they don't sell Cable Boxes. We hope someday they will.....

David P. - CA We called around for help on filters and nobody had any answers except one, call the "THE FILTER COMPANY" we did and had an answer in a matter of minutes.....

John B. - IL Our company buys in the 100 quantities and needs a large selection of different channels. "The Filter Company" was the only one who was able to supply....

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Linear power supply module Size:

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115vac 50/60hz 271 ma. Output: 5vdc 5000ma. Price... \$20.00 ea. 5 for \$75.00



Modem Coupling TRANSFORMER PART # 671-8284 IMPEDANCE (OHM) PRIM. SEC 600 294

INSERTION LOSS: 3.15 dB @ 1 kHz DC CURRENT: 0 MA RESISTANCE (ohm) PRIM. SEC

156 145 DIELECTRIC RATING: 1000 VAC

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PRICE \$1.25 EA. 120VAC 3900 OHM 50/60HZ 15 AMP DPST - BLUE POLY. DUST COVER. SOCKET OR SOLDER TERMINALS. Size: 1-3/8"H x 3/4"w x 1"deep PRICE \$1.75 EA.

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CHECKER

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PROGRAM DISKETTES contains list of all commands, a file that contains list of initialization steps, and a text file that has the 68hc11 program

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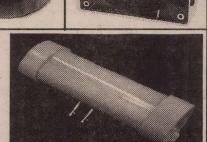
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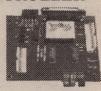
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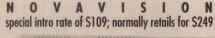
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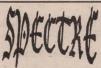
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Home Security with PCM and The Parallel Port



you how to create a simple home security system with a personal computer using PCM as the application framework. PCM, which stands for Pseudo Cooperative Multitasking, is an application framework, written in C, that helps you build structured control programs. The original PCM article appeared in the June '96 issue of *Nuts & Volts*, and also contains a short tutorial on different types of multitaskers and their design. Reading the original article will enhance your understanding

DB-25P Female 0 Strobe (out) Auto LineFeed (out) 0 14-2 0 Printer Error 0 15-(in) 3 0 0 16 Init Printer (out) 4 0 Select/Deselect (out) 0 17-Data (out) 0 0 18bits 0 - 7 0 0 19-0 O 20-0 0 21-0 Ground O 22-ACK (in) -10 0 (pins 18-25) O 23-Busy (in) 110 0 24 Paper Out (in) 12 0 0 On Line (in) -13 0 FIGURE 1

Port Address	Without mond	With mono
LPT1	0x378	0x3b
LPT2	0x278	0x378
LPT3	n/a	0x278
FIGURE 3	F	FIGURE 2

- d 0:0408 0000:0408 78 03 00 00 00 00 00 00

of PCM.

The PC parallel port is used for the application program's inputs and outputs, utilizing a homebuilt parallel port interface. The code presented writes directly to the hardware, bypassing DOS

by Jeff Stefan

and BIOS. The parallel port interface is simple, useful, and easy to program. All of the input scanning and output processing is written using Turbo C, and is by no means exclusive to the PCM application framework. It can be used in any Turbo C program to control the parallel port.

Controlling the Parallel Port

The PC parallel port is designed to connect to a printer via a DB-25P female connector mounted on the PC side, and a Centronics 36-pin male connector at the printer.

Figure 1 shows the parallel port pinout.

There are 12 outputs and five inputs. The 17 I/O lines are controlled and monitored by three registers labeled: Data, Status, and Control. In most of the PC parallel port literature, the lines assigned to the Data register have a D prefix. For example, data line zero is referred as D0, data line one is D1, and so on. The Status register lines have a S prefix (S0 through S7), and the Control lines have a C prefix (C0 through C7).

The installed parallel ports (LPT1, LPT2, and LPT3) and their base addresses are

located in the PC's BIOS interrupt vector table. These addresses are standard from machine to machine, and are easy to find. This is called the I/O Base Address. The address for LPT1 resides at vector table address 0000:0408 hex or 0x408

(this is how C likes to see hexadecimal numbers). The contents of this address depend on the type of display adapter connected to the PC. If a PC has a monochrome adapter, the parallel port addresses change. Figure 2 is a table showing the addresses for both the non-monochrome and monochrome printer port addresses.

One way to determine the parallel port addresses on your machine (chances are it's 0x378 for LPT1), is by using the DOS program debug. At the C:\>

prompt, type debug to start the program, then enter the dump command d 0:0408. This displays the printer addresses, with the output

appearing as in Figure 3.

The address is displayed as it appears in raw memory. It looks backwards, because it is. The low byte, 78 hex, is displayed first, then the high byte, 03 hex, is displayed next. This is called Little Endian byte ordering (Motorola chips use Big Endian byte ordering). It works out with the Intel architecture that this is the most convenient way to load the microprocessor's registers from memory.

For all practical purposes, just look at the second byte and put it in front of the first when determining the physical address. Believe it or not, the names Big and Little Endian come from the book *Gulliver's Travels*, by Jonathan Swift. In *Gulliver's Travels*, the Big Endian and

Little Endian political parties argued whether it's better to open soft boiled eggs from the big end or the little end!

Another way to find the parallel port addresses is to write a short C program using the MK_FP

macro. MK_FP means make far pointer. This creates a full address from the memory segment and offset values passed in as the macro arguments. The following short program, getport.c, gets the address of LPT1, the first parallel port, using MK_FP.

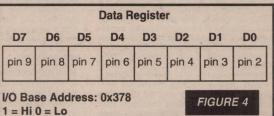
* getport.c * gets first parallel port I/O Base Address

#include <stdio.h> #include <dos.h> void main(void)

unsigned int PortAddress;
PortAddress = *(unsigned far *)MK_FP(0x40,0x008);
printf("LPT1 is at: [0x%2X]\n",PortAddress);

In this article, the I/O Base Address used is 0x378.

Now that you know how to find the I/O Base Address, you need to know how to read from and write to the registers. Turbo C, like most C



compilers, offer four functions to read and write to the PC's input/output registers. These functions are:

int inportb(int PortAddress); /* Read a byte (8 bits) from I/O port PortAddress */

int inport(int PortAddress); /* Read a word (16 bits) from I/O port PortAddress */

void outportb(int PortAddress, unsigned char Data); /* Write a byte (8 bits) to I/O PortAddress */ void outport(int PortAddress, int Data); /* Write a word (16 bits) to I/O) PortAddress */

The Data Register

The data register is an eight bit read/write register. The data register lies right at the I/O Base Address (0x378), as shown in Figure 4.

To write to the Data Register, you need to use outportb(0x378,Data);. The Data parameter value writes to the individual bits you need to turn on or turn off. The easiest way to write to a register is to create a set of #define statements to set the individual bits and pass them as the Data arguments:

#define CLEAR_DATA_BITS 0x00
#define SET_DATA_D0 0x01 /* binary 0000 0001 */
#define SET_DATA_D1 0x02 /* binary 0000 0010 */
#define SET_DATA_D2 0x04 /* binary 0000 0100 */
#define SET_DATA_D3 0x08 /* binary 0000 1000 */
#define SET_DATA_D3 0x08 /* binary 0001 0000 */
#define SET_DATA_D5 0x20 /* binary 0010 0000 */
#define SET_DATA_D6 0x40 /* binary 0100 0000 */
#define SET_DATA_D6 0x40 /* binary 0100 0000 */
#define SET_DATA_D7 0x80 /* binary 1000 0000 */

To set bit D0, us outportb(0x378,SET_DATA_D0);.

That's all there is to it. To set more than one bit at a time, create a unique #define. For example, if

Status Register								
S7	S6	S5	S4	S3	S2	S1	SO	
pin 11	pin 10	pin 12	pin 13	pin 15	?	?	?	
S1 Undefined S4 S2 Undefined S5			S3 S4 S5	Error On Line	1 0 1 Out 1	1 = True 1 = True 1 = True 1 = True		
FIGURE 5			The second second	Busy		= True		

int AlarmActive = 0;

int AlarmTripped= 0;

FIGURE 6	Status Register AND Masks							
	S7	S6	S5	S4	S3	S2	S1	SO
	1	1	1	0	0	?	?	?
AND Mask Value:	0	1	0	0	0	0	0	0
Result:	0	1	0	0	0	0	0	0

you want to set bits zero and two, this comes out to be a hexadecimal five.

The Status Register

simultaneously.

The Status Register is used to read inputs, and is

addressed at I/O Base Address +1. Software can only read this register. Any attempt to write to the Status Register is ignored.

The Status Register is shown in Figure 5.

The contents of the Status Register is d using the inportb(PortAddress) function. A n call is StatusRegState =

inportb(0x378+1);. Status Bit 7, S7, is different. S7 is inverted, so when a zero appears at pin 11, the input is considered True, and when a one appears, the input is considered False.

The contents of the entire status register is read using the inportb() function. In order to examine individual bits, a mask value must be used. This is similar to the way the SET_BIT defines are used to set individual output bits. The masks used are called AND masks, as shown in Figure 6.

If only one bit is set, as in the AND Mask Value in Figure 6, then that is the only bit that is effectively compared. Note that S5 is set and read, but is ignored when ANDed with the AND Mask Value. The AND Masks defines are as follows:

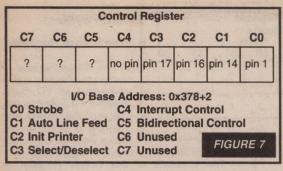
#define READ_STATUS_S7 #define READ_STATUS_S7
0x80 /* binary 1000 0000 */
#define READ_STATUS_S6
0x40 /* binary 0100 0000 */
#define READ_STATUS_S5
0x020 /* binary 0010 0000 */
#define READ_STATUS_S4
0x10 /* binary 0001 0000 */
#define READ_STATUS_S3
0x08 /* binary 0000 1000 */

An AND mask with S7 effectively inverts the output. If the switch assigned to S7 is open, this produces a one at the S7 input pin 11, and a zero appears at the S7 register bit. This special case is handled by considering a zero output from the AND mask as true (the switch being open), and a one output as false (the switch being closed). This is a constraint of the parallel port hardware, and cannot be changed. Bits S1, S2, and S3 contain garbage and cannot be used.

AND Mask Values are called using the following form:

	se #define SET_DATA_HEX5 0x05 0000 0101 */ to set the two l				
Filename: secure.c Description: model security system based on PCM Copyright (c) 1996 Jeffrey M. Stefan	int Result = 0; int AllClosed = 0; int FrontDoor = 0; /* S7 bit */ int BackDoor = 0; /* S6 bit */ int Window1 = 0; /* S5 bit */ int Window2 = 0; /* S4 bit */ int Window3 = 0; /* S3 bit */				
#include <stdio.h></stdio.h>	unsigned char StatusRegState = 0;	*****			
#include <strilib.h> #include <string.h> #include <crype.h> #include <conio.h></conio.h></crype.h></string.h></strilib.h>	* These variables are required to get the system time. time_t timer,first,second; struct tm *tblock;				
#include <time.h> #include <dos.h></dos.h></time.h>	* main : main() acts like a task dispatcher of				
/**************	***************************************	*****			
* Use this define to define the maximum tasks in your application. * Make sure that the number of tasks you run are at least equal to * the number defined here, otherwise unpredictable (bad) things * will happen. There are three tasks in this system.	main() { int nRetVal = 0; /* Return value from c int nCtr = 0; /* Function a(task) are				
#define MAX_TASKS 3	int (*fp)(); /* Function (task) poin	ter */			
* The following defines are used for task return identifiers. The individual tasks may return something to the main loop. Using these defines identify the task.	int *afp[] ={ /* array of function (tas nTimeMenuTask, nPortTask, nMonitorTask, NULL	nPortTask, nMonitorTask,			
#define TASK1_RET 1 #define TASK2_RET 2 #define TASK3_RET 3	clrscr(); CursorOff();				
/**************************************					
* Function (task) prototypes. These functions should return ints * In order to inform main that it has finished processing. For * example, nTimeMenuTask should return TASK1_RET, which is defined * as 1.	* Dispatch Tasks using infinite for loop. for(;;)				
/*********/ int nTimeMenuTask(void); int nPortTask(void); int nMonitorTask(void);	fp = afp[nCtr]; /* Get the ta nRetVal = (*fp)(); /* execute task nCtr++; /* Increment task counte	and get return value */			
void CursorOff(void); void CursorOn(void);	if(nCtr==MAX_TASKS) /* If last	task was run, start over */			
* Parallel Port Addresses.This is important! If you have a monochrome * monitor, the DATA_PORT address should be 0x3ab instead of 0x378.	nCtr=0;				
#define DATA_PORT 0x378 /* LPT1 Port Address for Data */ #define STATUS_PORT DATA_PORT + 1 /* LPT1 Port Address for Status */ #define CONTROL_PORT STATUS_PORT + 1 /* LPT1 Port Address for Control*/	* The following switch statement * from the task that just returned * structure, signal event, or act c * area will be application specific	The task may fill a data n some other data. This			
* AND Mask Defines to read and write individual bits	switch(nRetVal) {				
#define SET_DATA_D1 0x02 #define SET_DATA_D2 0x04	* Process the data that (none in this application)	t nTimeMenuTask worked or tion)			
#define SET_DATA_D3 0x08 #define SET_DATA_D4 0x10 #define SET_DATA_D5 0x20	break; case TASK2_RET:				
#define SET_DATA_D6 0x40 #define SET_DATA_D7 0x80	/****************	t nPortTask worked on.			
#define READ_STATUS_S7 0x80 #define READ_STATUS_S6 0x40 #define READ_STATUS_S5 0x20	break;				
#define READ_STATUS_S4 0x10 #define READ_STATUS_S3 0x08 #define READ_STATUS_S2 0x04 #define READ_STATUS_S1 0x02 #define READ_STATUS_S0 0x01	* (none in this applica	t nMonitorTasl worked on.			
/*************************************	break;				
int Duration = 0; int AlarmActive = 0	break; } /* switch */ } /*for loop */				

} /*for loop */
} /* main */



```
StatusRegState = inportb0x378+1); /* get the status port register info */
S6Bit = (StatusRegState & READ_STATUS_S6); /* check if S6 is set */
if(S6Bit)
printf("Back Door is Open!!");
eise
printf("Back Door is Closed");
To handle S7, use this:
StatusRegState = inportb0x378+1); /* get the status port register info */
S7Bit = (StatusRegState & READ_STATUS_S7); /* check if S7 is not set */
printf("Front Door is Open!!");
 printf("Front Door is Closed"):
```

break:

nState = 3;

gotoxy(30,9);

printf(" gotoxy(1,1);

if(!kbhit())

inCmd = getch(); switch(inCmd)

Enter Duration

break:

break;

clrscr():

= 0;

case 0x31:

/*****

* GO!

case 0x32:

* Cancel

case 0x33:

clrscr() gotoxy(30,3); printf("=====

case 2:

```
AlarmActive = 0;
                                          AlarmTripped= 0;
                                          Result
                                          nState
                                          break:
                            * exit System
                           case 0x33:
                                          nState = 4;
                           default:
                                          nState = 0;
                                          break:
             } /* switch */
return(TASK1 RET):
 Set Alarm. The tasks are temporarily suspended here
* in order to get new duration and time mark.
                                     ----\n"):
=\n"):
gotoxy(30,6);

printf("= 1. Enter Escape Time =\n");

gotoxy(30,7);

printf("= 2. Go =\n");

gotoxy(30,8);

printf("= 3. Cancel =\n");

gotoxy(30,9);
                                           =====\n");
              gotoxy(30,11);
printf("Enter Duration:(min): ");
scanf("%d",&Duration);
Duration = Duration * 60l;
              gotoxy(30,11);
printf("
                                                         ");
              gotoxy(1,1);
nState = 2;
              nState = 0;
              first = time(NULL);
AlarmActive = 1;
               clrscr();
              Duration
                             = 0:
               AlarmActive = 0;
              Result
nState
                            = 0:
```

```
In the PCM home security system application,
the S6Bit monitors the back door, and the S7Bit
monitors the front door.
```

The Control Register

The Control Register is designed to send control commands to a printer, and is addressed at I/O Base Address + 2. The Control Register is shown in Figure 7.

Two of the interesting bits in the Control Register are C5, the Bi-directional Control Bit, and C4, the Interrupt Enable Bit. On newer PCs, the

parallel port data bits are read/write, instead of write only. Whether data is read or written depends on the state of C5. If C5 is set to a one, then data is output. If C5 is zero, then the outputs are effectively disconnected, and the data lines may be used for inputs. The Interrupt Enable Bit, C4, allows an external device to generate an interrupt request (IRQ). If this bit is set, then an interrupt is generated on the low transition of the Status Register ACK input signal, C6. Usually this interrupt is set by jumpers to IRQ5. To use the interrupt, an interrupt service routine must be written, the address must be entered into the interrupt vector table, and the interrupt mask register must be enabled for IRQ5. Most of the older PC parallel ports do not have Bidirectional data lines, and almost none of the PC operating systems utilize the Interrupt Enable Bit. Our PCM application does not use the Control Register.

Building the Parallel Port Interface

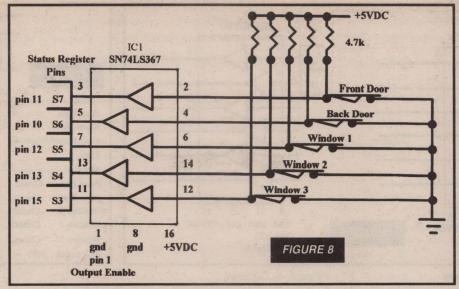
Building the Parallel Port Interface is straightforward, and is built using three SN74LS367 hex buffers and five 4.7K resistors. This is a bare bones interface that you will probably want to expand. Why not just use the parallel port alone to directly control inputs and outputs? Without some type of buffer, you run the risk of frying your parallel port. Using just the PC's internal +5 VDC power, the data outputs can only source 2.6 mA. Remember, the standard parallel port printer cable is only six feet long. I've seen them run about 20 feet, but that's really pushing it. Do unprotected! not run Buffering the inputs and outputs is simple and using sensible, and SN74LS367 TTL hex buffer chips is a cheap, easy, and reliable solution. The only shortcoming is that an external +5 VDC power supply is required, but the

```
nTimeMenuTask: This task gets the system time, displays, then
            dispatches menu selections
***********************
int nTimeMenuTask()
static int nState = 0;
int lnCmd = 0;
   The static nState variable remembers where the task left off.
  *When the task is called again by the dispatcher in main loop,
* the switch statement will place the program counter at the lase.
  * known state. This is effectively a cooperative context switch.
  switch(nState)
          case 0:
                    /****************
                    * Get and Display System Time
                    nState = 1;
                    /**************
                    * System Time Calls
                    timer = time(NULL):
                    tblock = localtime(&timer);
                    if(AlarmActive)
                               if(Result <= Duration)
                               second = time(NULL);
                               Result = difftime(second, first);
          return(TASK1_RET);
          break:
          case 1:
                     /***************
                     * Change nState to whatever new state is required.
                     gotoxy(30,3);
                    gotoxy(30,4);
printf("= Main Menu =\n");
gotoxy(30,5);
printf("=
                    printf("
                                   =====\n"):
                                      ============\n"):
                    gotoxy(30,6);
printf("= 1. Set Alarm =\n");
                    ====\n"):
                     gotoxy(1,1);
                     /*********************
                     get menu selection and
                     * check for shutdown request
                     nState = 0;
                     if(kbhit())
                               InCrnd = getch();
                               switch(InCmd)
                                         * Set Alarm
                                         case 0x31:
                                                   nState = 2:
                                                   break:
                                          Clear Alarm
```

added expense of a surplus power supply is well worth it. If you rely on the internal five volts from the parallel port for power, you could damage or destroy your port if something goes wrong.

Figure 8 shows the switch inputs for the simple security system, running through a SN74LS367 hex buffer. The switches are held low when closed, and pulled high when open.

Figure 9 shows how to connect the Data Lines (D0-D7) to the hex buffer chips. The Control Register outputs can be wired the same way. This PCM application doesn't use any outputs, and simply emits the PC's bell when the system is armed and violated.



```
break:
                                                                                                                            else
                                        default:
                                                      nState = 0:
                                                                                                                                          gotoxy(20,17);
printf("CLOSED");
                                                      break;
                          } /* switch */
             return(TASK1_RET);
                                                                                                                             Display Back Door status
            break;
                                                                                                                            if(BackDoor)
            case 3:
                          * Change nState to whatever new state is required.
                                                                                                                                          gotoxy(20,18);
printf("OPEN ");
                                                                                                                            else
                           return(TASK1_RET);
                                                                                                                                          gotoxy(20,18);
printf("CLOSED");
                          break:
            case 4:
                                                                                                                            /****************
                           * Exit Task Here
                                                                                                                             * Display Window1 status
                          gotoxy(1,1);
printf("** System Shutdown **");
CursorOn();
                                                                                                                            if(Window1)
                                                                                                                                          gotoxy(20,19);
printf("OPEN ");
                           exit(0);
nState = 0;
                           return(TASK1_RET);
                                                                                                                             else
                                                                                                                                          gotoxy(20,19);
printf("CLOSED");
             default:
                           break;
  } /* switch */
                                                                                                                              Display Window2 status
  return(TASK1_RET);
                                                                                                                                          gotoxy(20,20);
printf("OPEN ");
 nPortTask: parallel port control task.
int nPortTask()
                                                                                                                             else
                                                                                                                                          gotoxy(20,20);
printf("CLOSED");
static int nState = 0;
  switch(nState)
     case 0 :
              * Read inputs from Status Register
                                                                                                                             if(Window3)
                                                                                                                                          gotoxy(20,21);
printf("OPEN ");
             StatusRegState = inportb(STATUS_PORT);
              /* Get the status bit state
             FrontDoor = (StatusRegState & READ_STATUS_S7);
BackDoor = (StatusRegState & READ_STATUS_S6);
Window1 = (StatusRegState & READ_STATUS_S5);
Window2 = (StatusRegState & READ_STATUS_S4);
Window3 = (StatusRegState & READ_STATUS_S3);
                                                                                                                                          gotoxy(20,21);
printf("CLOSED");
                                                                                                                             * Set state of AllClosed
              * Display Front Door status
              if(!FrontDoor)
                                                                                                                                           AllClosed = 1;
                           gotoxy(20,17);
printf("OPEN ");
```

```
* Display Window3 status
if(FrontDoor && !BackDoor && !Window1 && !Window2 && !Window3)
```

The Home **Security System**

The PCM Home Security program is named secure.c. The display interface, shown in Figure 10, is intentionally very simple, and you are encouraged to customize and enhance it. The main display page is also shown in Figure 10.

The Main Menu selections allow the user to set and clear the alarm system (which is simply the bell output on the PC). Figure 11 shows the Set Alarm menu. The user first sets the escape time which delays the alarm system until everyone is out of the house. The alarm is

triggered when Go is selected, and aborted when Cancel is selected. The display returns to the Main Menu screen, and the Time Remaining (sec) value starts counting down. If the doors and windows are not closed by the time the value counts down to zero, the alarm sounds. If the alarm is active and any of the doors or windows are opened, the alarm will sound. To deactivate the alarm system, press menu selection 2, Clear Alarm.

Again, the program features are intentionally kept to a bare minimum. Keeping it lean allows you to put in the features you want, without being burdened with taking out things you don't like, or have little use for. This program is designed for easy expansion.

The secure.c

Program

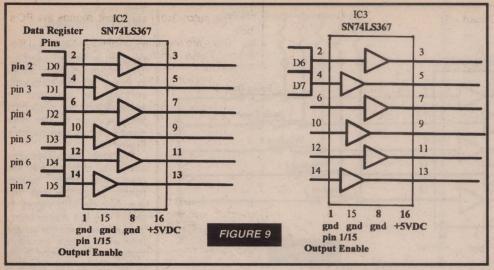
The Secure PCM application contains three tasks:

nTimeMenuTask(void); int nPortTask(void); int nMonitorTask(void);

The nTimeMenuTask gets the system time, displays and dispatches menu selections. The PC's system time is monitored by using the calls:

timer = time(NULL); tblock localtime(&timer);

The Time Remaining value in the Status display area



if(Result <= Duration)

```
Main Menu
                     1. Sct Alarm
2. Clear Alarm
3. Exit
 FIGURE 10
 Status
- Time/Date
                             : Wed June 19 17:48:06 1996
Alarm State
                             : Active
- Time Remaining (sec)
                             : 60
Monitor -
- Front Door : Closed
Back Door : Closed
- Window 1 : Closed
Window 2 : Closed
Window 3 : Closed
Raw Status: [7F]
```

```
if(AlarmActive)
                      AllClosed = 0:
           nState = 0;
           return(TASK2_RET);
break;
           default :
  break;
} /* switch */
  return TASK2_RET;
  nMonitorTask: Status Display Task
int nMonitorTask()
static int nState = 0;
int TimeDif
   switch(nState)
   case 0
            .
/**********************
            * Change nState to whatever new state is required.
           nState = 1:
           gotoxy(5,11);
printf("— Status —
gotoxy(5,12);
printf("- Time/Date
                                      : %s\n", asctime(tblock));
           gotoxy(5,13);
if(AlarmActive)
                      printf("- Alarm State
                                                  : Active\n");
                      printf("- Alarm State
                                                 : Inactive\n");
           gotoxy(5,14);
if(AlarmActive)
                      TimeDif = Duration - Result;
                      printf("- Time Remaining (sec): %2d\n", TimeDif);
            else
                      printf("
           gotoxy(5,15);
if(AlarmActive && TimeDif <=0 && !AllClosed)
                      AlarmTripped = 1;
                      printf("- Siren
                                                     \n");
           if(AlarmTripped)
                      gotoxy(5,15);
printf("- Siren
                                               : ALERT!\n");
                      putch(0x07);
```

determined by the following code:

```
second = time(NULL);
Result = difftime(second, first);
                   return(TASK3_RET);
         case 1:
                    * Display switch status then return to
                    * state 0 to monitor alarm.
                    nState = 0;
                   gotoxy(5,16);
printf("— Monitor—
gotoxy(5,17);
printf("- Front Door : \n");
                                                                                   -\n");
                   gotoxy(5,18);
printf("- Back Door: \n");
gotoxy(5,19);
printf("- Window 1:\n");
                   gotoxy(5,20);
printf("- Window 2 : \n");
                   gotoxy(5,21);
printf("- Window 3 : \n");
                   gotoxy(5,22);
printf("- Raw Status : [%2X]",StatusRegState);
                    gotoxy(5,23);
                   return(TASK3_RET);
                   break:
        default:
        break;
/* switch */
        return TASK3_RET;
        CursorOff: turns off cursor using BIOS screen interrupt 0x10.
        void CursorOff()
        union REGS regs;
                    regs.x.ax = 0x0100;
                   regs.x.bx = 0;
regs.x.cx = 0x3200;
                    int86(0x10,&regs,&regs);
```

CursorOn: turns on cursor using BIOS screen interrupt 0x10.

void CursorOn()

union REGS regs;

regs.x.ax = 0x0100; regs.x.bx = 0; regs.x.cx = 0x0607;

int86(0x10,®s,®s);

The variable first, which holds a previously stored time value, is set by the first = time(NULL) line of code in the GO command switch statement. Calling the difftime() function this way is a standard way of comparing time values, and is often used to benchmark algorithm execution time.

The nPortTask reads and processes the parallel port inputs. The Status Register is read by StatusRegState = inportb(STATUS_PORT);. The state of the individual bits is checked by:

FrontDoor = (StatusRegState & READ_STATUS_S7);
BackDoor = (StatusRegState & READ_STATUS_S6);
Window1 = (StatusRegState & READ_STATUS_S5);
Window2 = (StatusRegState & READ_STATUS_S4);
Window3 = (StatusRegState & READ_STATUS_S3);

Each of the status bits are examined and displayed by a series of if-else statements, starting with the Front Door switch.

```
* Display Front Door status
******************
if(!FrontDoor)
{
    gotoxy(20,17);
    printf("OPEN ");
}
else
{
    gotoxy(20,17);
    printf("CLOSED");
}
```

```
Set Alarm

1. Enter Escape Time
2. Go
3. Cancel

Enter Duration: (min):
```

FIGURE 11

******** Display Back Door status if(BackDoor) gotoxy(20,18);

printf("OPEN ");

else

gotoxy(20,18); printf("CLOSED");

Notice that the state of the Front Door switch in the if statement is different than the Back Door switch. Remember that S7 in the Status Register is inverted, so !FrontDoor checks to see if the variable FrontDoor is zero. If it is, the door is opened. This is the opposite of the remaining switches.

The variable AllClosed is one of the variables used to set the alarm. If all of the doors and windows are closed, then AllClosed is set to one. If any door or window opens, AllClosed is set to zero.

if(FrontDoor && !BackDoor && !Window1 && !Window2 && !Window3)

AllClosed = 1; else AllClosed = 0:

The last task in the system is nMonitorTask. This task receives and monitors the switch state and sets the alarm. The code that turns on and off the alarm is:

if(AlarmActive && TimeDif <= 0 && !AllClosed)

AlarmTripped = 1:

else

printf("- Siren \n"):

if(AlarmTripped)

gotoxy(5,15); printf("- Siren putch(0x07);

: ALERT!\n");

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bell There are two small functions at the end of the

The putch(0x07) statement sounds the PC's

listing. These functions are CursorOff() and CursorOn(). These two functions use the Turbo C int86() function to execute a BIOS call to turn off and turn on the PC's cursor.

The program listing is pretty short, just a little over 600 lines. Like the original PCM program, type it in, compile it, build the interface, and watch it run. Run it through a debugger to watch how it switches in and out of tasks and how it reads the Status Register inputs. Like a lot of you, I spend a large amount of time writing programs and tinkering with small computers. Writing a useful program is fun, but it's a downright thrill when it interacts with the outside world. Use the parallel port code and interface in your own projects, and consider using PCM as a structured control program framework. Some suggested applications using PCM and the parallel port include a sprinkler control system, a garage door control and security system, and a home weather station. So dust off and fire up that old computer. There's still some life left in that old XT! NV

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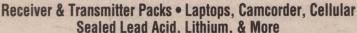
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THE SOLDERLESS

by Don Wilcher

Using Oscillators

My first encounters with oscillators occurred while taking an electronics circuits analysis course. In that class, I learned about sinusoidal waveform theory, positive feedback, amplifiers, and tuned circuits. All this information was digested by performing experiments in the lab, building, and analyzing oscillator cir-

The circuits that were breadboarded in the lab consisting of Colpitts, Hartley, and Armstrong networks. The frequencies of these circuits were recorded and compared to the calculated values. In addition to learning about these transistor circuits, we'd even experiment with a 555 Timer astable circuit.

Well this month, you will be able to experience the excitement that I've experienced by experimenting with a couple of oscillators. In the following paragraphs, we will investigate the basic components of an oscillator and see how to use this waveform generator in some useful applications. So without further delay, let's examine the make-up of a basic oscillator.

Oscillator Basics

An oscillator is a device or circuit that produces sinusoidal or other repetitive waveforms. Generally speaking, oscillators can be divided into three groups: (1) common sinusoidal (sinewave) oscillators, which produce and output waveforms that are sinusoidal; (2) common relaxation oscillators or multivibrators which produce nonsinusoidal (sawtooth, square, and pulse) waveforms; and (3) miscellaneous signal sources whose operation is based on principles and techniques less commonly encountered in projects like those discussed here in this column, such as "gas ionization oscillators," "electron-coupled oscillators," and "negative resistance oscillators."

The make-up of most oscillators is amplifiers with positive feedback. Positive feedback is where the output signal is fed back to the input of the electronic cir-

In using this design scheme, the oscillator's amplification must be greater than unity (1), feedback must be positive, and magnitude of the feedback signal must be high enough to sustain continuous oscillation. Another key component of the oscillator is its resonant frequency/sinusoidal characteristics.

If an inductor is wired in parallel to a capacitor and both these components are in series with a voltage source with a switch positioned between them to select one of the passive components, a dampening wave can be produced. If the voltage is across the capacitor, it will begin to charge and, with the switch in the inductor position, the voltage across the capacitor is released (discharged). If the switch is toggled between the capacitor and inductor at some pre-determined rate,

this charging and discharging effect will continue at a repetitive energy pattern. This pattern is a sinewave. The frequency of this simple passive circuit can be determined by using the equation:

Equation 1: $[f = 1/2 \times pi \times Squareroot (L \times C)]$

f is the resonant frequency of the tuned circuit in hertz pi equals 3.1412

L is the inductor in henries C is the capacitor in farads

Example Circuit Analysis

The oscillator circuit shown in Figure 1 will be used in a home security system for an audible alarm. What frequency would this circuit produce given a L of 10 henries and a C = 0.02 microfarads? Making the appropriate substitutions into the equation, the resonant frequency equals

 $fr = 1/2 \times 3.1412 \times squareroot (10H \times 0.02 uF)$ fr = 356 Hz

An observation of this equation will show that the resonant frequency can be increased by decreasing the value of L or C. If we go back to our example

Home Security Intruder Sensor Figure 1 Example oscillator circuit for design analysis problem.

oscillator circuit in Figure 1, by changing the value of 0.02 uF to 0.002 uF the resonant frequency equals

 $fr = 1/2 \times 3.1412 \times squareroot (10H \times 0.002 uF)$

This equation is a very useful tool for doing quick analysis on oscillator circuits such as the Hartley or the Colpitts. These two circuits use the LC-tuned network tied together with a transistor amplifier. Figure 2 shows the schematic circuits for the Hartley and Colpitts' oscillators. In observing these two circuits, some common differences are the number of passive components used and their placement. The major difference between the Hartley and Colpitts is the method of feedback.

In a Colpitts oscillator, the feedback is obtained across a portion of the total capacitance instead of a portion of the coil as in the Hartley. The resonant frequency of the Colpitts can be varied by changing either C1, C2, or both. In the Hartley, changing C or L will allow the resonant frequency to vary.

Now that the basics of oscillators have been discussed, let's take a closer look at the Hartley Oscillator via a "hands-on" project.

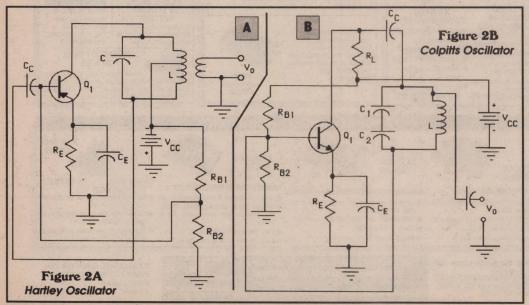
An Experimental Hartley Oscillator

The circuit in Figure 3 is a Hartley Oscillator. The tuned circuit establishes the resonant frequency for the oscillator. The 0.02 microfarad capacitor and the primary coil of a PCB transformer create the tuned circuit. The resonant frequency measured with a Fluke 87 Digital Multimeter (DMM) was 268 Hz. This particular multimeter has frequency measurement capabilities so that's why I chose to use this instead of a frequency counter. (Well, the real reason is that I was too lazy to get the frequency counter from off the shelf in the lab.)

The waveform of this oscillator is shown in Figure 3B. The reason behind this weird-looking signal is the piezo's internal resonant frequency intermixing with the Hartley's output. Another interesting item to point out is the "L" value from the circuit is missing. The reason for this missing component value is to illustrate the usefulness of the "fr" calculator shown in Equation 1. By doing some algebraic manipulation, we get the final equation to calculate "L"

Equation 2: $[L= 1 / C \times 4 \times pi^2 \times fr^2]$

making the appropriate substitutions into this equation:



THE SOLDERLESS BREADBOX

 $L = 1 / 0.02 uF \times 4 \times 3.1412^2 \times 268 Hz^2$ L= 17.6 or 18 H

For those of you who are not familiar with computer equation notation, the "^" means square. With this "L" calculator, any value of inductance can be computed, knowing a constant capacitance value and a desired resonant frequency. To calculate a "C" value given a constant inductance and "fr" values, the equation would be:

Equation 3: $[C = 1 / L \times 4 \times 3.1412^2 \times fr^2]$

The piezo-buzzer provides an audio output and the 470K resistor establishes the proper bias for the 2N2222 NPN transistor amplifier. The feedback path starts from the transformer coil and is sent to the base via the collector-base junction of the transistor. By depressing the key, the output of the oscillator can be heard through the piezo-buzzer and is vibrated by the resonant frequency of the tuned circuit. One application that immediately comes to mind is a Morse Code Sender. By depressing the key appropriately, dots and dashes (messages) can be transmitted across the air. For for those of you who are into ham radio, this circuit will make a great practice tool for the communication's portion of the radio license exam.

Another application that this circuit can be used in is science demonstrations in automotive engineering. By adding a variable resistor to the Hartley Oscillator, a fuel sending unit simulator can be creat-

ed. But what is a fuel sending unit?

Fuel Sending Unit Simulator

Quite often, we engineers come across an engineering problem where we need to reproduce a problem or solution on the bench. Instead of devoting personnel, manhours, and equipment to solve a vehicle problem, a simulator will be created first to create the problem in a control environment. Simulators allow engineers to reproduce the physical environment of interest on the bench.

In terms of the fuel sending unit, the Hartley Oscillator can simulate the action of the device quite easily by adding a variable resistor. Basically, a fuel tank sending unit looks like a toilet bowl float with a variable resistor and is used to measure the amount of gasoline in a fuel tank of an automobile. Figure 4 shows a pictorial diagram of the fuel tank sending unit

Depending on the amount of gasoline in the fuel tank, the variable resistor will send an electrical signal to the fuel gauge. The fuel gauge will show how much fuel is in the fuel tank. The electrical signal is proportional to the amount of resistance in the circuit. Less resistance would indicate a full fuel tank, whereas an empty tank would have more resistance on the variable resistor/sending unit.

The simulator circuit in Figure 5 was derived from a science teachers class I conducted in July of this year. The organization that sponsored this training session is called the Detroit Area Pre-College Engineering Program (DAPCEP). DAPCEP helps to motivate minority students to pursue careers in science and engineering by allowing them to take Saturday enrichment courses at local universities and

In designing this circuit for the teachers, I did not have 50 analog meters to give to the teachers. What was available for this project were the piezo-buzzers on the 60-In-1 Electronic Lab kits from Radio Shack that I used for the weekend program. So, instead of having a fuel gauge for visual indication of fuel used, the piezo provided an audible indicator for fuel.

With the circuit in Figure 5 breadboarded, closing the switch and turning the shaft of the 1K variable resistor will increase the volume of sound heard through the piezo-buzzer. Less amplitude correlates to low fuel in the tank where more amplitude corresponds to a full fuel tank. The frequency of this simulator can be changed by changing the value of "C" in the tuned circuit if a more desirable tone is required for those of you with sensitive ears.

Breadboarding Techniques

The circuits described in this column were designed and built using the Radio Shack 60-In-1 Electronic Lab kit. The cost of the lab kit is about \$14.99. This kit is wonderful for trying out circuit concepts, testing design hypothesis, and science and engineering demos. All of the components are mounted on the board and wiring is accomplished using spring terminals and pre-cut wire. This particular kit does not have integrated circuits, but they can be added by using a solderless breadboard and wiring them to the pre-mounted components of the

This method of construction allows the novice to build some innovative gadgets and devices without spending a lot of money for parts. Plenty of Solderless Breadbox projects and experiments were derived from this kit, as

well as using the 200 and 300-In-1 kits I have in the lab. The applications of these kits are only limited by the user's imagination and ingenuity.

Further Thoughts And Experimentation

Well folks, there you have it! This topic only scratched the surface of what oscillators can do, as well as circuit types. The 555 timer is an excellent IC for creating one-shot and free running oscillators.

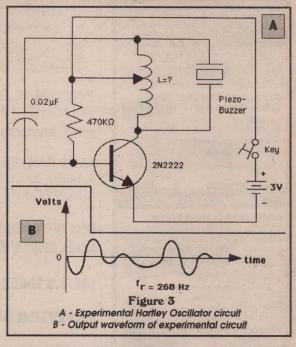
The 741 op-amp can be used to create sawtooth, sine, and squarewave signals for testing all types of analog circuits, digital circuits, and systems. One project that comes to mind is to create a function generator capable of producing triangle and squarewave

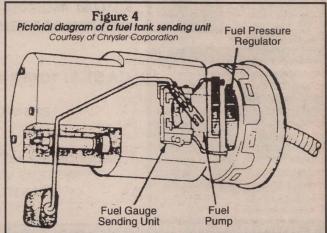
signals. This circuit is a standard device and can be found in any electronics hobbyist book in the public library

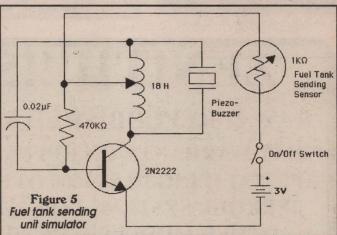
In addition, the Hartley Oscillator circuit described can be used to test those inductors lying around the bench. By attaching test leads to the circuit for connecting an inductor, this circuit will sound off if the inductor is good. This oscillator circuit lends itself to making a great and cheap audible continuity checker, as well. The circuit that is being worked on will provide a tone if the circuit is not broken. If the circuit being troubleshot is not working, the oscillator will sit in silence.

Last, upcoming school year, those of you who have children, now is the time to prepare for science fairs. This circuit has all of the makings for a good science topic on sound waves and measurements. By measuring the sound using a dB (deciBel) meter, and knowing the setting of volume control, a graph of dB vs control setting can be plotted. Well enuf said, see ya next month in the breadbox!!! NV

E-Mail address: MRDON@AOL.COM







Resources

The Art Of Electronics, 2nd Ed.: Paul Horowitz and Paul Winfield Hill, Cambridge Press, 1989.

Electronic Principles And Circuits: an Introduction to Electronics for the Technician: William E. Long and Paul L. Evans, John Wiley & Sons Inc., 1974.

Radio Shacks Science Fair 60-In-1 Electronic Project Lab: Cat. No. 28-261.



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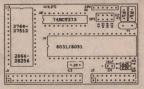
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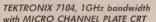
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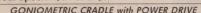
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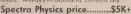
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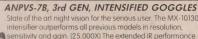


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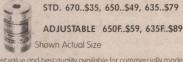


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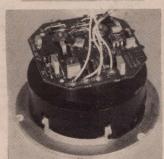
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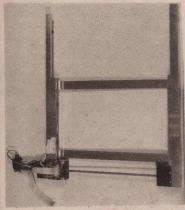


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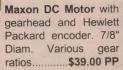
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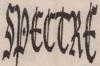
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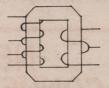


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by Karl Lunt

Three guys take on the 68hc12

n mid-June, Motorola announced its latest family of microcontrollers (MCUs), built around the newly-designed CPU12 core. The company targeted the first chip in this group, dubbed the 68hc812a4, for sampling by early July. Billed as a migration path from the 68hc11 to the 68hc16, the 68hc12 is packed with features ideal for robotics. Since most of the Seattle Robotics Society (SRS) members use 68hc11's in their robots, there was immediate interest in the new chip.

This month, I'll describe the journey that Kevin Ross, Marvin Green, and I took in developing a low-cost hacker tool set for the 68hc12. I'll also introduce you to several valuable sources of info on this chip, and give you a brief outline of the power this new device can provide in your robot designs.

A look at the 68hc12

I don't want to spend a lot of space on describing the 68hc12; for full details, contact your local Motorola distributor and ask for a copy of the MC68HC812A4 Technical Summary (MC68HC812A4TS/D). While you're at it, get a copy of the CPU12 Reference Manual (CPU12RM/AD), which serves as a programmer's reference manual for using the instruction set.

If you can't wait, or don't have access to a Motorola distributor, you can download pdf files of both documents from links reachable from http://129.38.232.2/index.html (yes, I know it's not a very user-friendly address). Just browse through the 68hc12 links and you'll get there. Please note that these are very large files; the file for the *Technical Summary* holds 180 pages of text and drawings, and is an exact duplicate of the printed document distributed by Motorola.

If you have an Acrobat reader (also available from the Motorola web site), you can actually view this document on your screen before printing it. This makes a very handy reference source, though I had some problems with my version. It acted as if it was using an imported font, rather than a font already on my machine, and the characters within the font got mixed up. This resulted in plausible, but wrong numbers for items such as pin numbers or timing info. The printed document, however, came out per-

I applaud Motorola for making these documents available in this format, especially as they followed the formal announcement of the chip by only a day or so. The documents contain enough information to start designing a 68hc12 system, even

though at the time, the real chips weren't available and the printed docs weren't shipping.

The first difference you notice between the 68hc812a4 and its older cousin, the 68hc11, is speed. Not only does this chip run on a crystal of up to 16 MHz, its internal CPU clock runs at half the crystal frequency, rather than using a quadrature system like the 68hc11. So right off, you're

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looking at a 4x speed boost by going to the newer device.

The 'a4 carries much more on-chip memory than the 68hc11 variants, as well. This first chip contains 1K of static RAM and 4K of EEPROM; plenty of room for fairly strong projects. It also contains an expanded set of 68hc11 peripherals, so you veteran 'hc11 users will feel right at home moving up.

For example, the 'a4 carries two asynchronous serial ports (SCI) and a synchronous serial port (SPI). Both have been enhanced over the same peripherals found in the older 'hc11. The 68hc12 SPI contains additional logic supporting a bidirectional serial interface. This means that a 68hc12 can send and receive data using only a clock line and a single data line, rather than the two data lines previously required. Note, however, that it appears all SPI transfers are still forced to use eight-bit exchanges.

Motorola also enhanced the two SCI channels over that found on the 68hc11. These SCI peripherals can perform parity calculations in hardware, and use separate modulus counters for baud rate generation. The latter gives you much more flexibility in defining baud rates. Oddly enough, given a crystal of 16 MHz, the fastest baud rate this chip can support is 38.4K baud.

The 68hc11's timer/counter system has always been one of its strongest points. Long-time readers of this column know well how to use it for generating pulse-width modulation (PWM) signals with zero software overhead. But the basic 68hc11 only supports this feature on its four output-com-

vos with no software overhead.

In the timer/counter area, Motorola has really beefed up the versatility of the pulse accumulator. It now uses a 16-bit counter and you can insert a /256 or a /65536 prescaler, if desired. On the downside, the 'a4 still only carries one pulse accumulator.

The 'a4 has an improved real-time interrupt (RTI) system, as well. Many of my projects use the 68hc11's RTI system to control down-counting timers, for timing motor motions or actuator delays. But using the typical 8 MHz crystal on a 68hc11, the fastest RTI signal I can get is 4.1 msecs. This is fine for most uses, but sometimes I need to divide a second into finer slices. The 'a4 provides seven different RTI rates, and using a 16 MHz crystal can get the RTI time slice down to 1.024 msecs.

The 'a4 also has an eight-channel A/D system much like that of the 68hc11. In the 'a4, these A/D channels provide eight-bit resolution, but the I/O registers for providing the A/D data all occupy two bytes. I asked a Motorola representative if this meant that future versions of the CPU12 family would offer 10-bit or higher resolution on the A/D ports, but he replied "No comment." I noted, however, the faintest of smiles on his face, so I'm looking for more resolution down the road.

Motorola did address one complaint I've long had with the 68hc11 A/D system. The 'a4 lets you read all eight A/D channels with one command, and to pull the results from eight separate registers. This eliminates the convoluted folding of eight A/D channels into four result registers, as used in

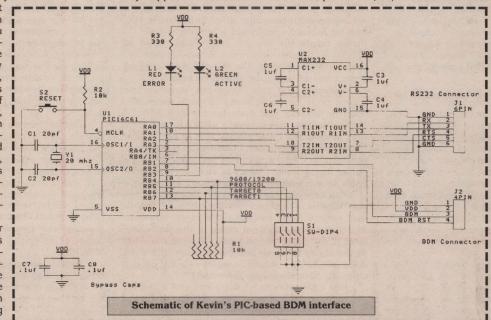
the 68hc11

The CPU12 family also offers features not seen on the generic 68hc11. For example, the 'a4 MCU has up to 24 I/O lines with key wake-up interrupt ability. Key wake-up means that the MCU can generate an interrupt if the state changes on one of these special input lines. You can use this wake-up interrupt feature to make a keypad that doesn't need to scan the matrix continuously; it just wakes up whenever the user presses a key.

The 'a4 also supports a non-multiplexed address and data bus, for memory expansion. The CPU12 architecture permits addressing of up to 5 Mbytes of external memory, using a windowing system. This is not as clean a layout as that used by the

68hc16, since some of the larger address space gets folded into windows in the chip's 64K linear address space. Still, it eliminates all of the hardware gyrations used by most 68hc11 large-memory designs.

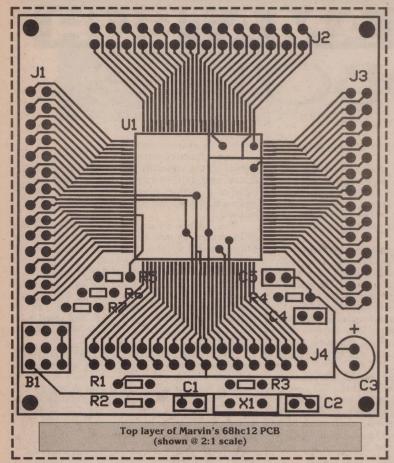
Along with the increased address space, the 'a4 has seven programmable chip select lines. These I/O lines provide chip select signals for



pare (OC) channels, while the remaining three channels can only handle input compare (IC) functions

The 68hc812a4 has eight timer/counter channels, any of which can be configured for IC or OC operation. You can still slave any of these OC channels to timer 7, which acts as a master timer channel. This would let you run seven hobby ser-

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whatever address space you declare, simplifying the addition of external memory or I/O devices.

Motorola has abandoned the 68hc11's bootstrap loader in favor of a background debug module (BDM) similar to that used on the 68hc16 and 68332 chips. The BDM provides the main entry way to the 'a4, providing support for downloading programs, dumping memory and registers, and doing single-step, trace, and breakpoint operations. The CPU12 BDM system is very different from that used in the other two chips, and it presented a substantial obstacle in getting our first project running; more on this later.

Hidden inside the CPU12 are even more surprises. Motorola has made substantial improvements to the programmer's model, adding some very powerful indexing modes to not only the X and Y registers, but to the S (stack) and PC registers, as well. These enhancements should help quiet the grumblings from those programmers who moved to the 68hc11 from the 6809 and felt the pain of the 68hc11's feeble indexing modes. Of course, software types are rarely satisfied, and many will ask why Motorola didn't throw in a second stack register or another accumulator. Still, the CPU12 represents a big upgrade to the 68hc11 family, and I'm looking forward to working with the new model.

And Motorola added support for fuzzy logic and digital signal processing (DSP) on this chip. Several special CPU12 opcodes permit you to set up a fuzzy logic engine in only 50 or so instructions. The docs go into considerable detail on how to set up and use these instructions, and Jim's article, cited above, includes much more information. The DSP instructions, while lacking the strength of a full-power DSP controller, will get you started on DSP programming and will solve some problems previously beyond the reach of the 68hc11.

In all, this first member of the CPU12 family is a large win for those who have outgrown the 68hc11. Readers of this column should prepare for a shift toward 68hc12 projects, though I intend to keep supporting my 68hc11 tools such as SBasic.

The first 68hc12 project

Motorola's announcement fired up a lot of people in the SRS and on the Internet. Marvin, Kevin, and I immediately grabbed the available docs and started checking out what was obviously the future in small robots. Each of us focused on a different aspect of this new chip.

Marvin Green, developer of the 68hc11 BOTBoard and BOTBoard-2, used his documentation to start the design of a singlechip PCB; sort of a BOTBoard for the 68hc12. Kevin Ross, with his expertise in PCbased software and his considerable Windows experience, took up the challenge of a PC-based front end for the BDM. And I zeroed in on PCbased tools, such as an assembler, for writing new 68hc12 programs

once Marvin and Kevin finished their tasks.

We quickly realized that we just didn't have enough information to complete our individual tasks. Using the Internet, I was able to locate a Motorola engineer, Jim Sibigtroth, who proved to be a valuable 68hc12 resource. Jim is a senior member of the technical staff at Motorola's Advanced Microcontroller division in Austin, TX, and he recently published a lengthy article on Fuzzy Logic for the 29 July 1996 issue of Electronic Engineering Times. When Jim read my plea for assistance on the 68hc12 development project we had begun, he offered his help.

Motorola, via Jim Sibigtroth, donated the source code for the in-house absolute assembler used to develop test code for the first 68hc12 chips. The code was originally written for use on a Unix system, but I was quickly able to port it over to the DOS environment. I cleaned up a few bugs, removed an Easter egg or two, and prettied up the output it generates. Thanks to Motorola's support, I now have a working absolute 68hc12 assembler.

You can use this tool, known as as12, to create Motorola .s19 object files from 68hc12 assembler source files. Because this is an absolute assembler, you don't get some of the fancy features of the expensive relocating assemblers, such as the ability to make linkable modules. But as12 does offer many strong features lacking on the older Motorola shareware assemblers.

For example, as 12 supports the #define, #include, and #ifdef constructs familiar to C programmers. as 12 also provides an extensive macro capability, useful for defining a set of commonly-used functions.

I didn't lack for sample assembler language source code to use in testing as12. The CPU12 chips are source-compatible with 68hc11 assembly. Note the term "source-compatible," not "object-compatible." You can assemble a 68hc11 source file and get code that will run on a CPU12 device, but you cannot take a 68hc11 executable

file and expect it to run on a 68hc12 chip.

This happens because Motorola kept the assembler mnemonics the same for both chips, but revamped much of the underlying opcodes. For example, the 68hc11 instructions dealing with the Y-register all begin with a \$18 pre-byte. Thus, a LDY #\$a5 assembles into the bytes \$18 \$CE \$00 \$a5, where the instruction LDX #\$a5 becomes \$CE \$00 \$a5. This means your 68hc11 programs pay a one-byte penalty for each use of the Y-register.

On the 68hc12, however, Motorola re-arranged the opcodes so the penalty disappears. Instructions using the X- and Y-registers assemble to the same number of bytes. One obvious fallout of this re-arrangement means that 68hc11 object files won't run on the 68hc12.

You can check out the new as12 assembler by accessing my web site at http://www.seanet.com/~karllunt — Note that this is a zip file for running on a PC. Those of you with Macs, take heart. Porting of this assembler to the Macintosh environment is underway, and a Mac version of as12 should be ready by the time you read this.

My next task was to get my hands on some actual 68hc12 chips. After a few calls to Motorola, I reached their samples desk at 1-800-635-8293. I signed up for two samples of the 'a4, which arrived by mail a few days later. Interestingly, these chips carried the part number PC68HC812A4. The PC prefix marks these as prototype devices, with designs even more unstable than the XC, or experimental chips. This means these chips are using silicon dies with a design that may well be unreliable or poorly documented. As it turns out, the PC chips I received had some serious design flaws in them, which hampered our efforts. But we only discovered that later.

Now, however, I was the proud owner of two 68hc12 chips, each housed in a 112-pin thin quad flatpack (TQFP) package. This package design uses leads only 0.5 mm wide on 0.65 mm centers. Marvin's efforts to design his board were slowed because we just couldn't get definitive information on lead arrangment and spacing. Eventually, however, he received enough information to complete the layout for holding the chip.

Marvin also added traces for the necessary resistors, capacitors, and other passives to use the 'a4 in single-chip mode. He included the same three-pin power connector and four-pin serial connector on this board as used on his BOTBoard products. And he even added four layouts for 28-pin dual-row headers, used to bring each pin of the 'a4 out to the board edge for probing or connection.

This last feature really makes his board design flexible. On one of his prototype boards, I installed four dual-row male IDC headers. I then built up a Radio Shack experimenter board with matching female connectors. Now I can just plug my 68hc12 PCB into a matching experimenter board, moving the MCU from board to board as needed. Or, you could solder the female connectors into the 68hc12 board. This would let you plug wires into the connector pins and out to a prototyping board, so you could quickly build up test circuits around a 68hc12.

Marvin ran his ideas past Kevin and I many times, and between the three of us we eventually hit a design that Marvin liked. He then packaged up his design and shipped it to Alberta Printed Circuits (APC) for etching. APC, as usual, did a first-rate job and had the boards back to Marvin in a couple of days.

While Marvin was getting his PCB design finalized, Kevin was up to his armpits in the BDM interface. His task would have been far simpler if Motorola had stuck with the BDM design used in the 68hc16 and 68332 devices. Those chips use a three-wire interface that contains a clock line, a data-in line, and a data-out line. The clock line,

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controlled by the host, means that the PC talking to the MCU dictates the rate at which the data transfer occurs. Thus, you have no problems with synchronizing the timing between the host and target systems.

For some reason, Motorola dumped this design on the 68hc812a4, opting instead for a one-wire interface. Here, the target and host exchange data over a single wire. The timing for the exchange is dictated by the target MCU, in that the host must use a data transfer rate that is based on

the MCU's crystal. Unfortunately, the host does not have visibility as to the transfer rate, since there is no clocking signal in the BDM interface.

After much discussion among us, Kevin chose to build a BDM interface around a PIC 16c84. He needed to use a 20 MHz PIC device, to keep up with a 'a4 using a 16 MHz crystal. We had built our

with a 'a4 using a 16 MHz crystal. We had built our first boards using 8 MHz crystals, however, so Kevin could get by with a 10 MHz PlC, available from Digi-Key for less than \$10.00. Refer to the accompanying schematic of Kevin's PlC-based BDM interface. Note that the schematic calls out a PlC 16c61, which is basically a one-time program-

mable version of the 16c84.

The PIC 16c84 is highly cool, since it comes with 1K of EEPROM to hold your code. It doesn't have a lot of on-chip peripherals such as serial ports, but the bloody thing runs so fast that you can bit-bank serial data with little trouble. Plus, Microchip's web site at http://www.microchip.com is crammed with technical info, application notes, and source code for plenty of finished projects and subsystems. You can also find the latest versions of Microchip tools, such as the MPASM PIC assembler.

Kevin designed his BDM interface to run from the PC's serial port. The first cut at his PIC electronics used 5 VDC and a four-wire cable to the target 68hc12 board. This cable supplies connections for ground, +5 VDC, BDM data, and a reset line. The latter wire is needed because Kevin's PIC uses it to reset the target board and control what mode the target enters after reset.

As you can see from the accompanying schematic, there just isn't very much to this BDM interface board. Kevin added some LEDs for status display, a MAX232 to provide the needed RS-232 level shifting, and a DIP switch to select one of several operating modes on power-up. For example, one switch tells the PIC that the target system runs

with either an 8 or 16 MHz crystal.

Most of the hard work in Kevin's design took place in the PC and PIC software. Kevin created a very nice Windows-based program that acts somewhat like a modem program, exchanging data and commands with the PIC BDM interface. As the PIC receives each command, it translates it into the proper BDM sequence and sends it to the target. If the PIC needs to handle a response from the target, it collects the data and translates for passing back to the PC host.

Putting it all together

Our mutual efforts converged one Friday afternoon at my home in Bothell, WA. Marvin had driven up from Portland with several copies of his first PCB, and Kevin had brought much of his software design tools over from nearby Woodinville. With our gear spread out on a table in my living room, we began the task of integrating all our efforts.

While Kevin worked on completing the PIC software, Marvin started building up one of his 68hc12 boards and I began breadboarding a PIC BDM interface. It took a couple of hours, but I finally had the electronics ready to fire up. Marvin, in the meantime, had all of his 68hc12 board soldered except for the one device none of us wanted to touch, the MCU itself. The fine pitch on its leads just looked like too much for us to handle, given our relatively crude tool set.

We solved this problem with typical software ingenuity; we went out for Chinese food. By the time we returned, we had a plan of action. And since these were my chips, I was elected to start the soldering operation.

I began by spreading a very thin film of Kester paste flux on the traces destined to hold the 68hc12 chip. I intentionally used the thinnest possible film, since I only wanted enough paste to stick the chip into place. I then positioned the chip onto the board, taking care to align the dot on the chip's case with pin I on Marvin's board.

If you build such a board of your own, pay close attention to this alignment procedure. The text on the chip's surface, giving the Motorola logo and part number, gets printed in a random orientation, and you cannot use the printing for chip alignment. Always use the alignment dot when placing the chip on the board.

Next, I worked very carefully to gently shift the chip on the trace layout, aligning the chip's leads with the matching traces. This took a little time, as the flux only keeps the chip from sliding, but lets you push it around fairly easily. Once I had the board resting on my work surface and I was satisfied with the alignment, I was ready to solder the leads.

I cleaned the tip of my grounded soldering iron, then added a thin film of solder to the tip. Be sure to use a grounded-tip iron to minimize risk of static damage while soldering the chip in place. Then, using only the soldering iron and no other solder, I gently touched the trace at one of the chip's corners. The solder plating on the trace melted and fused with the film of solder on the iron's tip, then wicked underneath the pin. When it cooled, the lead was soldered. I then moved to the diagonally opposite pin and repeated the operation. This left the chip tacked in place at two corners, essentially immobilizing it.

Take care during these steps not to touch the chip or its leads. If you do, you risk misaligning the chip and soldering it in the wrong place at the same time. Once soldered by two or more adjacent pins, you will have a tough time removing the chip

and repositioning it.

With the chip locked in place, I started at a third corner and began soldering the remaining pins. I still soldered only to the traces, letting the melted solder wick underneath the pins to complete the connection. We took turns on this phase and when done, I went over all connections with a magnifying glass to check for opens or shorts. Satisfied that we had actually soldered this baby in place accurately, we were ready for some power.

We connected Kevin's BDM electronics to the PC and to the 68hc12, then applied power in the form of four Renewal C-cells. Using a logic probe, we were able to verify that the MCU's clock was operating properly. This sparked a lot of grinning and congratulations, since it meant that Marvin had designed the correct trace layout for the many power and ground pins, as well as the passives needed to get the chip running.

When Kevin entered a command on his PC screen and actually got a response from the BDM interface, there was more stomping around and high-fiving. We were having a great time celebrating these small steps, because it had taken so

much hard work just to get this far.

However, we eventually sobered up when we realized that the PIC BDM interface was not returning the proper responses to our commands. Evidently, the MCU was either not running or the PIC interface had some kind of timing problems. We spent an hour or more trying different fixes, but we finally called it quits around 2:00 AM that Saturday morning.

During the next week, Kevin worked with Keith Payea, another SRS stalwart, to verify the timing on his PIC interface. He

cleaned up a few small problems, but still could not make the MCU respond properly. In desperation, I once again turned to the Internet, posting a request for help to the 68hc11 listserver. And once again, Jim Sibigtroth came through. Thanks to his intimate knowledge of this chip's inner workings, he was able to solve our problem. It turns out that "first silicon," that is the experimental devices we were using, had a flaw in the BDM logic. This flaw inserted an additional /2 in the BDM timing chain, making the chip's interface run at half the advertised speed.

Kevin added code to his PIC firmware to compensate for this timing flaw and the MCU "woke up" and began talking to the PIC via the BDM interface! Immediately, Kevin was able to write data to RAM, to download simple programs, and to single-step through those programs. We had done it! Using nothing but cheap hacker tools and the club's expertise, we had put together a 68hc12 development system.

What's next?

This kind of project just grows without bounds. Marvin is already making plans for selling his 68hc12 PCB; contact him directly for pricing and availability. Kevin has also talked about selling PIC chips already programmed with the firmware for his BDM interface; I've supplied his E-Mail address elsewhere. As I've already mentioned, you can get a copy of the current version of the as12 assembler from my web site.

In the meantime, I've already started work on converting tiny4th to run on the 68hc12 MCU; the code should be finished by the time you see this. Check on my web site for a copy of the new compiler. Porting SBasic to the 68hc12 will take considerably more time, but I will work on it whenever I can.

And then there are all the robots to build using this new chip. The small physical size and huge I/O capabilities mean that you can pack a lot of robot smarts into a very small package. Previous designs that called for one or two added chips can now be done using just this one device. And Marvin's 68hc12 BOTBoard is only slightly larger than its 68hc11 cousin, so you won't pay a large price in real estate for using the newer device.

If you think that 4K of EEPROM might limit you in your designs, then stick around until the fourth quarter, when Motorola is scheduled to ship first silicon on the 'b32 version of the 68hc12. This device will contain 32K of flash EPROM. As I understand, it is targeted to come out in an 80-pin QFP package, which will make it even smaller than most 68hc11s. The smaller lead count means that Motorola will sacrifice some of the I/O; check the Motorola web sites for details on this new chip. **NV**

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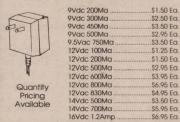
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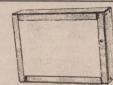
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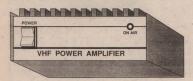
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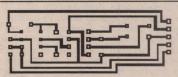


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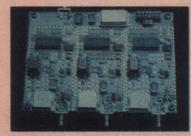
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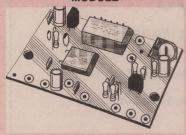
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Continued on page 110

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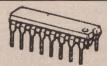
RADAR SETS Wanted. WWII 1970 X-band, military, civilian working or not, books, TMs, parts. SHF, Box 10215, Pittsburgh, PA

WANTED: HIGHEST prices paid for Western Electric tubes and equipment, also wanted 2A3, 10, 45, 50, KT-66, KT-88, 5691, 5692, 6550(st), and Altec 604B/C speakers. Don Singerhouse, PO Box 768, Maywood, NJ 07607. Phone/Fax: 201-843-8682.

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MILITARY TRANSISTORS & DIODES & INTEGRATED CIRCUITS WANTED. ELECTRONIC MATERIAL INDUSTRIES 818-842-3539, FAX 818-845-3613.

WANTED: MILITARY "MILES" systems. Parts, transmitters, receivers, helmet and chest sensors, test sets, manuals, etc. BPB Assoc., 1977 High St., Longwood, FL 32750. 407-330-3408, FAX 407-330-3498.

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Events

SEPTEMBER 1996

CA - LIVERMORE - Swapmeet. Las Positas College. Noel Anklam 510-447-3857 CA - SANTA ROSA - Computer Show. Sonoma Co. Fairgrounds. MarketPro. 415-456-6730

SEPTEMBER 4-5-6-7

NY - ROCHESTER - Antique WA Conf. Marriott. Joyce 607-739-5443

SEPTEMBER 7

CA - OAKLAND - Computer Show. Convention obert Austin. 1-800-346-0100 CA - SANTEE - ARC of El Cajon Ham, Computer &

Electronic Swapmeet. Santee Drive-in. 619-561-0052 CA - STOCKTON - Computer Show. Civic

CA - STOCKTON - Computer Show. Clivic Auditorium. MarketPro. 415-456-6730 FL - PLANTATION - Computer Show. Holiday Inn. Directions: 954-472-5600. Dealers: 954-792-663 FL - POMPANO BEACH - Computer Show & Sale. Palm-Aire Hotel. Narisaam 1-800-436-6707

IL - ROLLING MEADOWS - W9DXCC DX Convention & Banquet. Holiday Inn. 708-343-1696 IN - SPENCER - Owen Co. ARA Hamfest. Kathryn Smith 812-829-2140

IN - SUMMIT CITY - Hamfest. Allen Co. Fairgrounds. Fort Wayne RC. Bill or Becky 219-471-5657

KS - TOPEKA - NE KS ARC Hamfest. Expocentre Ag Hall. Robert Nall 913-267-9399

MA - WILMINGTON - Computer Show. Shriner's.

KGP Productions 908-297-2526

MA - FALMOUTH - FARA Auction. Methodist Church, Jo 508-548-1121

ME - WINDSOR - Ham/Electronic Flea Market. Fairgrounds. Frank 207-623-9217

MI - SOUTHFIELD - Computer Show. Southfield Pavilion. 313-283-1754

MN - WALKER - Hamfest. Cass-Hubbard ARC. Bill Beebe 218-547-3147

PA - ALLENTOWN - Computer Show & Sale Merchant's Square. Peter Trapp Shows. 607-369-

PA - UNIONTOWN - Hamfest. Uniontown ARC clubhouse grounds. Carl 304-594-3779

PA - ERIE - Hamfest '96. Radio Assn. of Erie. Chris

Robson 814-474-1211

SEPTEMBER 7-8

AL - HUNTSVILLE - Computer Show, Von Braun Civic Center. ComputerShow 770-907-6225 CANADA - MANITOBA - AUSTIN - Hamfest. Manitoba Amateur Radio Museum. 204-728-2463
CA - OAKLAND - Robert Austin Computer Show. Convention Center. 1-800-346-0100 CA - VENTURA - Computer Show. Ventura Fairgrounds. MarketPro. 415-456-6730

KY - LOUISVILLE - Hamfest, Computer Show, & State Convention. Kentucky Fair & Exposition Center. 812-294-4905

CA - SACRAMENTO - Computer Show. Scottish Rite Center. MarketPro. 415-456-6730

CA - SAN DIEGO - Computer Show. Scottish Rite Center. MarketPro. 415-456-6730

IA - DUBUQUE - Hamfest/Radiofest/Computer

Expo. Dubuque Co. Fairgrounds. Loren Heber 319-556-5755

Dile - JOLIET - Hamfest '96. Inwood Recreation Center. Bolingbrook ARS. 708-759-7005 FL - FORT LAUDERDALE - Computer Show. Sheraton Inn. Directions: 954-772-7770. Dealers: 954-792-6639

FL - WEST PALM BEACH - Computer Show & Sale. Palm Beach Airport Hilton. Narisaam 1-800-436-

MA - DARTMOUTH - Radio & Electronic Fleamarket. Clubhouse. Bill 508-996-2969

MA - SEEKONK - Computer Show. Johnson & Wales Inn. Northern Computer Shows 508-744-8440 MD - GAITHERSBURG - FARFEST '96

Montgomery Co. Agricultural Center. 301-490-3188 MI - LANSING - Computer Show. Holiday Inn South Convention Center. 313-283-1754

NY - BINGHAMTON - Computer Show & Sale. Kalurah Temple, Endicott. Peter Trapp Shows. 607-

NY - WOODBURY - Ham/Electronic Flea Market.

LIMARC. Mark 516-796-2366
OH - FINDLAY - Hamfest & Computerfest. Hancock

PA - PHILADEL PHIA - Computer Show, Butler Farm Showgrounds, Roe Airport, 412-282-1183
PA - PHILADEL PHIA - Computer Show, Holiday Inn Sports Complex, Micro-Mart 609-924-2344

SEPTEMBER 13-14-15

IL - PEORIA - Superfest 1996, ARRL National Convention. Downtown Peoria Civic Center. Peoria

he Events Calendar is a free service limited to electronic events such as computer shows, hamfests, flea markets, etc. If your organization is sponsoring an event and would like a free listing, contact us at least 60 days prior to the event. Include your flyer, estimated attendance, name of the person to contact, and phone number.

Complimentary issues are available upon request for distribution to your attendees. A street address for UPS is required.

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All listing information should be sent to

Nuts & Volts Magazine Events Calendar

430 Princeland Court Corona, CA 91719

Phone 909-371-8497 Fax 909-371-3052

E-mail events@nutsvolts.com

ateur Radio Club. 309-692-3378 **WI - MADISON -** Computer Sale/Show, Dane Co. Expo Center. Blue Star Productions. 612-788-1901

SEPTEMBER 14

CA - FONTANA - Inland Empire ARC Amateur Radio & Electronics Swapmeet. A B Miller High School. Bill 909-822-4138 eves

CA - LOS ALTOS HILLS - Electronics Flea Market. CA - LUS ALTOS HILLS - Electronics Free Market.
Foothill College, Parking Lot A. 408-734-4453
CA - MONTEREY - Computer Show. Conference
Center. MarketPro. 415-456-6730
CA - SAN FRANCISCO - Computer Show. Cow
Palace. Robert Austin. 1-800-346-0100
CA - SAN JOSE - Computer Show. Santa Clara Co.
Fairgrounds. MarketPro. 415-456-6730

CA - SANTEE - ARC of El Cajon Ham, Computer &

Electronic Swapmeet. Santee Drive-in. 619-561-0052 CT - PAWCATUCK - Electronic Auction.

Neighborhood Center. Bill 401-596-5849
IN - INDIANAPOLIS - AGI Computer Fair.
Indianapolis Events Center. 317-299-8827
LA - PRAIREVILLE - Ascension ARC Hamfest.
Shane Dugas 504-673-8396

MA - TOPSFIELD - Computer Show. Fairgrounds. Northern Computer Shows 508-744-8440 MI - KALAMAZOO - Computer Show. County Fairground. 313-283-1754

NY - BALLSTON SPA - Hamfest '96. Sarat

County Fairgrounds. Saratoga County RACES Association. Lenny 518-885-4933 NY - SYRACUSE - Computer Show & Sale. NY

State Fairgrounds, Horticultural Bldg. Peter Trapp Shows, 607-369-2796 OH - CLEVELAND - Computer Show & Sale.

Cloverleaf Trading Center. Peter Trapp Shows. 607-369-2796

SEPTEMBER 14-15

CA - SAN FRANCISCO - Robert Austin Computer Show. Cow Palace. 1-800-346-0100 FL - MELBOURNE - Hamfest. Platinum Coast ARS.

Allen Hudson 407-773-9658

SEPTEMBER 15

CA - LANCASTER - Computer Show. Antelope
Valley Fairgrounds. MarketPro. 415-456-6730
CA - MODESTO - Computer Show. Centre Plaza at
Red Lion. MarketPro. 415-456-6730
CT - NEWTOWN - Western CT Hamfest. Edmond
Town Hall. Candlewood ARA. John 203-438-6782
IL - ORLAND PARK - Orland Park Computer Show.

Civic Center. 24-Hour Hotline 708-974-3123

KY - LOUISVILLE - Computer Fair South. Executive
West Hotel. MEKA Enterprises 812-333-9300

MA - CAMBRIDGE - MIT RS & Harvard Wireless Club Hamfest. Nick Alternburnd 617-253-3776

ME - AUGUSTA - Computer Show. Civic Center. Northern Computer Shows 508-744-8440

MI - FARMINGTON HILLS - Computer Show Activity Center. 313-283-1754

MI - MT. CLEMENS - L'Anse Creuse ARC Hamfest. Ralph Irish 810-731-6760

NE - OMAHA - AK-SAR-BEN ARC Hamfest. Dave

ARCO SP.4930 NJ - HAMILTON TWP. - Fallfest '96. Tall Cedars of Lebanon picnic grove. DVRA. 609-882-2240 OH - AKRON - Computer Show & Sale. Tadmor Shrine Temple. Peter Trapp Shows. 607-369-2796 WV - WHEELING - Triple States RAC Hamfest Ralph McDonough 614-546-3930

SEPTEMBER 17

OH - TOLEDO - Computer Show. Masonic Great Hall. 1-800-893-SHOW

SEPTEMBER 20

NJ - PISCATAWAY - Ham & Electronics Auction. North Stelton Firehouse. Piscataway ARC. Marty

908-574-2873

SEPTEMBER 20-21-22

WA - SEATTLE - Digital Communications Conference. Quality Inn Seattle Airport. ARRL & TAPR. 817-383-0000

WI - WAUSAU - Computer Sale/Show. Wausau/Marathon Co. Park. Blue Star Productions. 612-788-1901

SEPTEMBER 21

AR - MOUNTAIN HOME - Twin Lakes ARC

Hamfest. Phil Waters 501-425-7406

CA - BAKERSFIELD - Computer Show. Red Lion Inn. MarketPro. 415-456-6730

CA - SEBASTOPOL - Swapmeet/Auction. Holy Ghost Hall. Sonoma County Radio Amateurs. Rick 707-575-4455

FL - ST. PETERSBURG - Computer Show. Days Inn. Directions: 813-522-3191. Dealers: 954-792-

MA - FALL RIVER - Computer Show. White's of Westport. Northern Computer Shows 508-744-8440 ME - LINCOLN - Hamfest. Ella Burr School. BARC. Sylvia Cockburn 207-732-5185 MI - LIVONIA - Computer Show. Livonia Elks Hall.

NE - LEXINGTON - Consignment Auction. Knights of Columbus Hall. Randy 308-987-2312

NJ - PENNSAUKEN - Computer Show. South
Jersey Expc Center. KGP Productions 908-297-2526
NY - HAMBURG - Western NY Section Convention.
Harold Smith 716-424-7184

OK - ALTUS - Hamfest. Altus Area ARA. Mart Dennis, POB 73, Altus, OK 73521

OR - PORTLAND - Ham Fair. Pay N Pak Building, Delta Park, 503-257-4822

Detta Park. 303-227-4622 RI - FORESTDALE - Ham/Electronic Flea Market & Auction. VFW Hall. Rick 401-725-7595 VT - RANDOLPH CENTER - Central VT ARC Hamfest. Tom Girardi 802-244-7836

SEPTEMBER 21-22

AK - ANCHORAGE - Hamfest, Kincaid Park Outdoor Center. ARRL. Robert Wilson 907-271-5304 CA - VALLEJO - Computer Show. Solano Co.
Fairgrounds, MarketPro. 415-456-6730
IL - GRAYSLAKE - Radio Expo '96. Lake County
Fairgrounds. Chicago FM Club. 708-457-0966
VA - VIRGINIA BEACH - Hamfest. Virginia Beach Pavilion. 804-HAM-FEST

SEPTEMBER 22

FL - NEW PORT RICHEY - Hamfest/Computer Show. Recreation Center. Ron 813-376-6575 FL - ST. PETERSBURG - Computer Show. Best Western Mirage. Directions: 813-525-1181. Dealers: 954-792-6639

MI - ADRIAN - Adrian ARC Hamfest. Brian Sarkisian

MI - DEARBORN - Computer Show. Civic Center. MI - FLINT - Computer Show, Holiday Inn. Gateway

Centre. Five Star Productions. 810-230-0988

MO - COTTLEVILLE - Swapfest. St. Charles Co.
Community College. St. Peters ARC. 314-723-4200

NH - PORTSMOUTH - Computer Show. Frank Jones Center. Northern Computer Shows 508-744-8440

NJ - MT. HOLLY - Hamfest/Computer Show. Mt. Holly Armory. South Jersey Radio Association. 609-596-7749

OH - CLEVELAND - Hamfest Assn. of Cleveland WI - MADISON - Hamfest, Dane Co. ARES, Bob Aspinwall 608-222-3444

SEPTEMBER 27-28

KS - WICHITA - Hamfest. Wichita ARC State Convention. Sonjia Cowart-Clover 316-284-0262

FL - MELBOURNE - Computer/Electronics Show. Civic Auditorium. Bill Arflin 407-777-9692 WA - WALLA WALLA - Hamfest. SE. WA Section Convention. Walla Walla Valley ARC. 509-525-2529

SEPTEMBER 28

AL - ANNISTON - Hamfest. City Meeting Center.

AAERC. Harold 205-435-4833

CA - CONCORD - Computer Show, Centre Concord. MarketPro. 415-456-6730

CA - SAN DIEGO - Computer Show. Scottish Rite

CA - SAN DIEGO - Computer Show. Scotush Nit Center. MarketPro. 415-456-6730 CA - YUBA CITY - Computer Show. Yuba-Sutter Fair. MarketPro. 415-456-6730 GA - GAINESVILLE - Hamfest. Lanierland ARC. Jerry Adams 770-967-2445 IN - TERRE HAUTE - Computer Fair. Vigo Co.

Fairgrounds. 812-442-0110 LA - ALEXANDRIA - Hamfest. Central LA ARC. Jon

MI - HUDSONVILLE - Hamfest. Grand Rapids ARA. os 616-698-7022

MI - GRAND RAPIDS - Super Swap '96. Unity Christian High School. Dave DeVos 616-698-7022 MI - TAYLOR - Computer Show. Democratic Club Hall 313-283-1754

MN - CHISHOLM - Hamfest. Midrange ARC. Dan

Weinberger 218-492-4501
NY - HORSEHEADS - Elmira International NY - HORSENEADS - CITING INTERNATIONAL Hamfest/Computerfest. Chemung County Fairgrounds. Elmira ARA. Dave 607-589-7495 TN - COOKEVILLE - Hamfest. Cookeville Rptr. Assn. & TN Tech ARS. Rich King 615-526-7171

SEPTEMBER 28-29

AL - MOBILE - Hamfest. Mobile ARC. Bill Cleveland

KY - LOUISVILLE - Computer Fair. Kentucky Fair & Expo Center. 513-263-3378

NC - CHARLOTTE - Computer Show. Metrolina Expo Center. ComputerShow 770-907-6225 NJ - MORRISTOWN - Computer Show. National Guard Armory. KGP Productions 908-297-2526

SEPTEMBER 29

CA - ANTIOCH - Computer Show. Contra Costa Fairgrounds. MarketPro. 415-456-6730 CA - CONCORD - Computer Show. Centre Concord. MarketPro. 415-456-6730

CA - OXNARD - Computer Show. Community Center. MarketPro. 415-456-6730

CA - SANTA ANA - Swapmeet. ACP Parking Lot. Mary Russo 714-558-8813

IL - GLEN ELLYN - Computer Show & Sale. College of DuPage, 708-940-7547 MA - FRAMINGHAM - Fleamarket & Exams

Framingham High School, FARA, 508-435-0564 MI - MADISON HEIGHTS - Computer Show. UF &

CW Hall. 313-283-1754

MI - MIDLAND - Computer Show. Holiday Inn. Five Star Productions. 810-230-0988

NH - NASHUA - Computer Show. Sheraton Tara Hotel. Northern Computer Shows 508-744-8440

NY - YONKERS - Electronic Flea Market. Lincoln High School. Metro 70cm Network. 914-969-1053 OH - ASHLAND - ARRL Hamfest. Ashland Area

ARC. David Fike 419-289-1082

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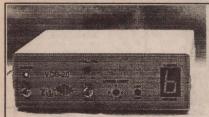
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ELECTRONICS

• • • • • • • • • • • • With TJ Byers

In this column, I answer questions about all aspects of electronics, including computer hardware and software. This column doesn't replace the Tech Forum that you've grown to love and support. Instead, it will supplement it, so feel free to participate as always with your questions and answers. You can reach me on America Online at TJBYERS, on the Internet at TJBERS@aol.com or by snail mail at Nuts & Volts Magazine, 430 Princeland Ct., Corona, CA 91719.

Relay Contact Resistance

Q. Concerning electromechanical relay contacts, I would like information about the so-called "dry contact" effect. I hear it occurs when a set of contacts switches less than a minimum current that causes contact surface deterioration that leads to increased contact resistance. Do all relay contacts exhibit

> Joseph F. Richmond Joppa, MD

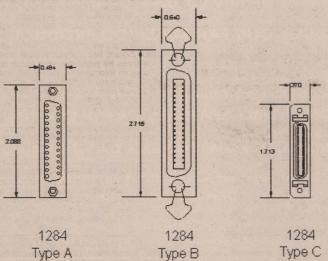
A. I think what you're asking about is the lack of arcing that often cleans relay contacts. When a switch or relay breaks contact, there's a residue arc "connection" that most engineers try to squelch using snubber circuits. DC voltages are the hardest to squelch, and can actually do damage to the contacts by burning them out. AC voltages, by their very nature, automatically extinguish the arc because the voltage has to pass through zero on its way from positive to negative. When the switching current is very low, you don't get arcing, which can cause the relay contacts to oxidize ("rust") that, over time, increases the contact resistance. For these applications, they make a wetted relay. A wetted relay uses mercury on its contacts. Because mercury is a liquid, a new surface is formed each time the contacts break, which prevents oxidation and provides lowers contact resistance.

IEEE Printer Cables

Q. A lot of the new printers sold today tell you, for best results, to use a IEEE printer cable. While shopping at a national computer store chain, I found a wide variety of printer cables, ranging in price from \$6.00 for a standard parallel cable to \$18.00 for a bidirectional cable to over \$30.00 for an IEEE cable. All these cables have a DB25 male connector on one end and a 36-pin Centronics connector on the other. Can you tell me what the difference is besides the price?

Charles E. Heisler K3VDB Red Lion, PA

A. It used to be so simple, didn't it? The \$6.00 cable is the type you'd use with an old dot-matrix printer. In the last few months, printer vendors have started shipping printers that talk back to the PC, telling it things like it's out of paper or ink. For these, you need a bidirectional cable. The IEEE 1284 cable is double-shielded to prevent EMI (electromagnetic interference), and allows you to send data to your printer at 2 MBs.



If that's not confusing enough, the IEEE 1284 standard also specifies a mini, 36-pin "Centronics" connector (above) that forces you to look at the back of the printer before you buy a printer cable. Will an IEEE cable make your computer run faster? Not likely unless you've bought a Pentium system with a color laser printer in the last month.

Dot-Laser Printer Problem

Q. I bought a Panasonic KXP-1624 dot-matrix printer in need of repair. At first it was totally dead, but I was able to coax it back to life. Unfortunately, it's totally confused. The self-test works okay, but documents print in gibberish. I have new printer drivers that Panasonic sent to me, and have installed them several times over with the same results. Any suggestions?

> George R. Adam Rio Vista, CA

A. The last time I repaired a printer (a year ago), it cost me \$80.00, and that was for a dot-matrix printer that needed a new "motherboard" which I had to install myself. If I had to do it again, I wouldn't. Why? Because you can buy an ink jet printer for about \$200.00 — and they're a lot faster! Perhaps, with my editor's blessing, we can have a future article that takes a hard look at who's who in the ink jet printer market.

Fluorescent Ballast

Q. I'm interested in converting all of my fluorescent light fixtures from the bulky, noisy, short-lived, and generally poor quality ballast transformer type, to the quiet, cheap, and long-lived capacitor ballast type. I need to know several things before I embark on this journey. Is the principal of capacitance reactance for florescent tubes the same as inductance reactance, or are there other factors that come into play when using capacitors? Is there anything out of the ordinary when choosing a capacitor for this type of work other than the AC rating and duty cycle?

Bieber, CA

A. Let's start at the beginning. Capacitors are a lot different than inductors in that one has leading current and the other has lagging current, and the two have to be in sync to equalize the power factor. Since a fluorescent tube is basically a capacitive device, you need an inductor to balance the equation. Fluorescent lamps also require a high-voltage kick to get them started — a kick that the ballast provides. The ballast provides power for the filaments, too. In other words, a capacitor ballast won't work. What you're probably thinking of as a "capacitor" ballast are actually electronic ballast, which are found in newer fixtures. They're essentially a switching power supply in a metal case. BTW, the reason ballast transformers are "short-lived" is not the fault of the transformer, but the fault of a bad lamp. When you see a fluorescent tube flickering, replace it now. What happens is that as the tube dies it draws a lot of current, which overheats the ballast.

Laser Printer Woes

Q. I have a Panasonic KXP-4420 laser printer that was working fine. Now all of the sudden, it doesn't work. When I turn it on, it goes through the warm-up process then gives the error message "Call service - E31." I'm curious if I REALLY need to call for service. Is this something I can correct?

Jeff Aust via Internet

A. Unfortunately, no. This warning means the fuser assembly heater isn't working. That's the part that makes the laser printer a laser printer. It could be the heater itself (\$200.00) or a problem with a sensor on the controller board (\$150.00) - plus labor. What I would do is replace your defunct laser printer with a color ink jet printer, like Cannon's BJC4000 (\$249.00).

Door Ajar Indicator

Q. I used to have a Sears brand garage door opener that had a provision for sending a low voltage to an open door indicator. Instead of using the large, ugly, Sears indicators, I ran low-voltage wiring to a retro-junction box and, with a resistor to drop the 24 volts, used a flashing LED. I recently replaced the Sears opener with a new Genie brand garage door opener, but it lacks this feature. I would like to use the garage door indicators I have installed in the house without rewiring them. Is there a way, other than using X-10 devices, of utilizing a low-voltage open door indicator with the Genie door opener that would allow me to use my existing low-voltage indicators?

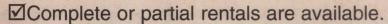
Dean Olson Omaha, NE

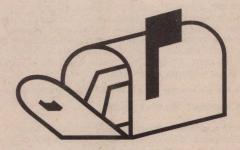
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MA - BOXBORO - ARRL Convention. Boxborougi Woods Hotel & Conference Center. 617-334-5770 MS - BILOXI - ARRL Hamfest. B.J. Byrd 601-832-3249

OCTOBER 6

CA - LIVERMORE - Swapmeet. Las Positas College Noel Anklam 510-447-3857

IN - BEDFORD - Hamfest. Lawrence Co. 4-H Fairgrounds. Dewy Sowders 812-279-5546 IN - HUNTINGTON - Hamfest. Huntington County ARS. Ray Tackett 219-786-0057

MA - TAUNTON - Computer Show. Holiday Inn. Northern Computer Shows 508-744-8440 NY - WOODBURY - Hamfest. Long Island Mobile ARC. Mark Nadel 516-796-2366 OH - SPRINGFIELD - Hamfest. Clark Co. Fair-

grounds. Independent Radio Assn. 513-568-4070 **PA - WARRINGTON -** ARRL Hamfest. Mt. Airy VHF Radio Club. Paul Drexler 609-538-1687

OCTOBER 11-12-13

AZ - MESA - ARRL Convention, Sheraton Mesa Hotel & Conference Center. Scottsdale ARC. 602

CA - BAKERSFIELD - Hamfest. Bakersfield ARA. Ed Harlander 805-589-4163 MN - ST. PAUL - Computer Sale/Show. State

Fairgrounds. Blue Star Productions. 612-788-1901 **TX - HOUSTON -** South Texas Section Convention. Darrell Glueck 713-544-4829

OCTOBER 12

CA - FONTANA - Inland Empire ARC Amateur Radio & Electronics Swapmeet. A B Miller High School. Bill 909-822-4138 eves

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CO - GOLDEN - Computer Show/Swap Meet. Jefferson Co. Fairgrounds. Greg 303-444-2664 FL - TAMPA - ARRL Hamfest. Egypt Temple ARA Joseph Fara 813-933-2322

Joseph Fara 813-935-222
ID - NAMPA - Hamfest & Computer Fair, Nampa Civic Center, Jay Diyer 208-467-3245
IN - INDIANAPOLIS - AGI Computer Fair, Indianapolis Events Center, 317-299-8827
GA - AUGUSTA - Hamfest & Computer Show.

Evans Middle School. ARC of Augusta. 706-860-

MI - ALPENA - Hamfest, Thunder Bay ARC, Tom Twite 517-354-2469 **NJ - TEANECK -** Hamfest, Bergen ARA, Bruce

Lemken 201-967-7041

WA - BREMERTON - Hamfest. North Kitsap ARC. Robert Tomas 360-638-1659

OCTOBER 12-13

CA - OXNARD - Computer Show. Community

Center. MarketPro. 415-456-6730
CA - VALLEJO - Computer Show. Solano Co. Fairgrounds. MarketPro. 415-456-6730
KY - LEXINGTON - Computer Fair. Heritage Hall.

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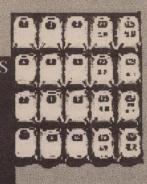
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OCTOBER 13

CT - DURHAM - Hamfest, Computer Show & State Convention. Fairgrounds. Kit 860-873-2978, Bill Wawrzeniak 203-269-8252

IL - OAK BROOK TERRACE - Hamfest. Oak Brook Terrace. Chicago ARC. George 312-545-3622 MD - COLUMBIA - Hamfest. Howard County Fairgrounds. Columbia ARA. 410-964-4378 MI - FLINT - Computer Show. Holiday Inn, Gateway Centre. Five Star Productions. 810-230-0988 MI - MASON - Ham Fair, Ingham Co. Fairgrounds, LCDRA & CMARC, Jeff Oberg 517-393-4713 NJ - LINCROFT - Shore Area Hamfest & State Convention. Al Allen 984-495-3246 NY - QUEENS - Hall of Science ARC Hamfest, NY Hall of Science parking lot. Arnie 718-343-0172

OH - LIMA - Hamfest. Northwest Ohio ARC. 419-

OCTOBER 15

OH - TOLEDO - Computer Show. Masonic Great Hall. 1-800-893-SHOW

647-6321

OCTOBER 18-19

FL - STARKE - Hamfest. Bradford Area ARC. Lev Francis 352-473-3511

OCTOBER 18-19-20

CA - CONCORD - PACIFICON '96. Mt. Diablo ARC. Richard Schulze 510-932-6125

GA - ATLANTA - Grove Communications Expo. Larry Van Horn 704-837-9200

CA - BAKERSFIELD - Computer Show. Red Lion Inn. MarketPro. 415-456-6730

CA - OAKLAND - Robert Austin Computer Show. Convention Center. 1-800-346-0100

CA - SANTEE - ARC of El Cajon Ham, Computer & Electronic Swapmeet. Santee Drive-in. 619-561-0052
PA - SENECA - Hamfest. Ft. Venango Mike & Key Club. Mary Housholder 814-437-2036

TN - GRAY - ARRL Hamfest. Bristol, Kingsport, & Johnson City ARCs. Wendell Messimer 423-928-4407

AL - HUNTSVILLE - ComputerShow. Von Braun Civic Center. 707-907-6225

FL. WEST PALM BEACH - Hamfest. South Florida Fairg Expo Center. Vi Kiekenapp 407-585-9074 MA - MARLBOROUGH - Computer Show. Royal Plaza Trade Center. KCP Productions. 908-297-2526 TX - EL PASO - International Hamfiesta. Ron Love 915-595-3371

TX - HOUSTON - Gulf Coast Ham Convention. 1-800-231-3057

OCTOBER 20

CA - LANCASTER - Computer Show, Antelope Valley Fairgrounds. MarketPro. 415-456-6730 CA - STOCKTON - Computer Show. Civic Auditorium. MarketPro. 415-456-6730 CO - GOLDEN - Hamfest. Jefferson Co Fairgrounds. Joe Dickinson 303-771-9577 IL - ORLAND PARK - Orland Park Computer Show Civic Center. 24-Hour Hotline 708-974-3123

IL - SALEM - Hamfest. Centralia Wireless Assn.

KY - LOUISVILLE - Computer Fair South. Executive West Hotel. MEKA Enterprises 812-333-9300 MA - CAMBRIDGE - Hamfest, MIT RS & Harvard Wireless Club, Nick Alternburnd 617-253-3776 MI - KALAMAZOO - Hamfest. Kalamazoo Co. Fairgrounds. 616-657-4482

MI - WARREN - Swap-N-Shop. OLR Conference Center. Kevin 810-636-3414

NY - MELVILLE - Hudson Division Conv./Hamfest. Huntington Hilton Hotel. JoAnn Colletti 516-399-

NY - MELVILLE - LI NY RCARC Ham/Electronics Flea Market. Steve 516-584-5909

PA - GREENSBURG - Hamfest. Greensburg Hose Company #1. Foothills ARC. Al 412-523-3727 PA - SELLERSVILLE - Hamfest. Sellersville Fire House. Linda Erdman 215-679-5764

OCTOBER 22

OH - LIMA - Computer Show. Holiday Inn Lima. 1-800-893-SHOW

OCTOBER 25-26-27

OK - KINGSTON - Hamfest. Texoma Hamarama Assn. Charles Bilbay 405-924-0687

OCTOBER 26

CANADA - QUEBEC - LONGUEUIL - Hamfest. Francois Drien 514-672-9994

CA - SANTA ROSA - Computer Show. Sonoma Co.

Fairgrounds. MarketPro. 415-456-6730 FL - PORT ST. LUCIE - Hamfest '96. Port St. Lucie

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MA - PLYMOUTH - Ham/Electronic Flea Market.

Mayflower RC. Jon 508-746-0162 MN - ST. PAUL - Hamfest & Computer Expo. St. Paul Civic Center. Twin Cities FM Club. 612-535-

MO - ST. LOUIS - Halloween Hamfest. West County Tech. Dennis 314-351-3568

NH - NASHUA - Ham/Electronic Flea Market. Res. Ctr. Church. NE Antique RC. 617-923-2665

OR - RICKREALL - Swap-Toberfest, ARES/RACES Convention. Polk Co. Fairgrounds. 503-585-5924 SC - ANDERSON - Hamfest. Anderson Radio Club. J. Walter Sanders 864-964-9097 SC - SUMTER - Hamfest. Sumter Co. Exhibition

Center. Mike Dunlap 803-481-4611

MA - WALTHAM - Photographica. Ed 617-965-0807 NJ - HACKENSACK - Computer Show. FDU/ Rothman Center. KGP Productions. 908-297-2526 TN - EAST RIDGE - Hamfest. Chattanooga ARC. Louise Carter 423-821-4043

OCTOBER 27 CA - CONCORD - Computer Show. Centre Concord. MarketPro. 415-456-6730

CA - FRESNO - Computer Show. Fres Fairgrounds. MarketPro. 415-456-6730 CA - SAN DIEGO - Computer Show. Scottish Rite Center. MarketPro. 415-456-6730

IA - DES MOINES - Hamfest. Tikva Tracers ARC.

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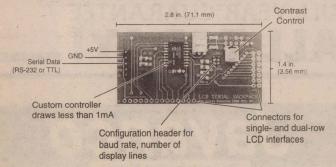


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IN - LEBANON - Hamfest. Boone Co. & Clinton Co.

ARCs. Michael Ottinger 317-482-1866

MA - SOUTHWICK - Ham/Electronic Flea Market.

HCRA Rec. Ctr. Barry 413-747-7010

MD - WESTMINSTER - Computer & Hamfest Carroll Co. Ag Center. CCARC. George 717-632-1621
OH - MARION - HamFiesta & Computer Show.

Marion Co, Fairgrounds. MARC. Karen 614-499-3565
PA - NEWTOWN - Hamfest. Bucks Co. Community
College. Penn Wireless Assn. Steve 215-752-1202

OCTOBER 29

OH - COLUMBUS - Computer Show. Aladdin Temple. 1-800-893-SHOW

NOVEMBER 1996

NOVEMBER 2

CA - SANTEE - ARC of El Cajon Ham, Computer & Electronic Swapmeet. Santee Drive-in. 619-561-0052 FL - SORRENTO - Hamfest & Electronic Expo. East Lake Chamber of Commerce. Tony 352-360-1380

IL - GODFREY - ARRL Hamfest. Lewis & Clark RC. Brent Urfer 618-498-4865

NY - HEMPSTEAD - Computer Show. Hofstra University, Field House. 908-297-2526 OK - ENID - Hamfest. Hoover Bidg., Garfield Co. Fairgrounds. Fred N5QXL, Jeff EnidARC@AOL.COM

NOVEMBER 2-3

CA - OAKLAND - Robert Austin Computer Show. Convention Center. 1-800-346-0100 CA - SACRAMENTO - Computer Show. Cal Expo. GA - LAWRENCEVILLE - ARRL Hamfest. Alford Memorial RC. Randy Bassett 770-410-3989
OH - CINCINNATI - Computer Fair. Cincinnati
Gardens. Mark Hanslip 513-263-3378
TN - NASHVILLE - Computer Show. State
Fairgrounds Computer Computer Show. Fairgrounds. ComputerShow 770-907-6225

TX - ODESSA - Hamfest. Ector Co. Coliseum Exhibit Bldg. C. West Texas ARC, Robert 915-335-

NOVEMBER 3

CA - LIVERMORE - Swapmeet. Las Positas College Noel Anklam 510-447-3857

MI - MIDLAND - Computer Show. Holiday Inn. Five Star Productions. 810-230-0988 OH - CANTON - ARRL Hamfest. Massillon ARC. Jack Cale 216-477-8261

PA - TREVOSE - Photographic Show. Radisson Hotel. OMM Productions 610-527-5903

NOVEMBER 9

AL - MONTGOMERY - Hamfest & Computer Show Garrett Coliseum, South Alabama State Fairgrounds.
Montgomery ARC. Phil 334-272-7980
CA - FONTANA Inland Empire ARC Amateur
Radio & Electronics Swapmeet. A B Miller High
School. Bill 909-822-4138 eves

FL - TITUSVILLE - ARRL Hamfest. Titusville ARC.

Clyde Lee 407-267-9375
IN - INDIANAPOLIS - AGI Computer Fair. Indianapolis Events Center. 317-299-8827

LA - WEST MONROE - Hamfest. Conv. Center.

Becky Ketchell 318-343-3530
PA - HERSHEY - Hamfest. Hershey Armory. Harold

SC - MYRTLE BEACH - Radio/Computer Show

High School. Dave Berry 803-248-9401

NOVEMBER 9-10

GA - DALTON - Computer Show. NW GA Trade & Conv. Center. ComputerShow 770-907-6225 NJ - EDISON - Computer Show. NJ Convention Expo Center, Raritan Center Expo Hall. 908-297-

NOVEMBER 10

CT - BRANFORD - SCARA Flea Market. Branford Intermediate School. 203-483-0856

NOVEMBER 15-16-17

IA - DES MOINES - Computer Sale/Show. State Fairgrounds. Blue Star Productions. 612-788-1901

GA - WATKINSVILLE - ARRL Hamfest. Athens Radio Club. David Alper 706-354-1952

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NOVEMBER 16-17

CA - SAN FRANCISCO - Robert Austin Computer CA - SAN FRANCISCU - RODERT AUSIIN COMPUTER Show. Cow Palace. 1-800-346-0100 CA - SANTEE - ARC of El Cajon Ham, Computer & Electronic Swapmeet. Santee Drive-in. 619-561-0052. IN - FORT WAYNE - Hamfest. Allen Co. War Memorial Coliseum Expo Center. Don 219484-1314

MA - METHUEN - Computer Show. Valley Expo
Center. KGP Productions. 908-297-2526

NOVEMBER 17

IL - ORLAND PARK - Orland Park Computer Show. Civic Center. 24-Hour Hotline 708-974-3123

KY - LOUISVILLE - Computer Fair South. Executive West Hotel. MEKA Enterprises 812-333-9300 NC - BENSON - JARSFest '96. American Legion Complex. 919-894-3352, 919-894-3100 NY - POUGHKEEPSIE - Hamfest. Mt. Beacon ARC.

NOVEMBER 20

FL - ST. PETERSBURG - ARRL Hamfest. Pelican Chapter #128 (QCWA). Don Bice 813-347-2702

NOVEMBER 23

CA - OAKLAND - Robert Austin Computer Show. Convention Center. 1-800-346-0100

NOVEMBER 23-24

FL - TAMPA - Radio/Computer Conv. FL Expo Park, Expo Hall. Jean 813-525-5178 NC - CHARLOTTE - Computer Show. Metrolina Expo Center. ComputerShow 770-907-6225 PA - FT. WASHINGTON - Ft. Washington Expo Center. KGP Productions 908-297-2526

NOVEMBER 24

CA - SANTA ANA - Swapmeet. ACP Parking Lot. Mary Russo 714-558-8813 IL - GLEN ELLYN - Computer Show & Sale. College of DuPage. 708-940-7547

NOVEMBER 29-30

OH - SHARONVILLE - Computer Show. Sharonville Convention Center. 1-800-672-3472

NOVEMBER 29-30-DECEMBER 1

GA - MARIETTA - ComputerShow. Cobb County Civic Center. 770-907-6225

NOVEMBER 30

IL - LITCHFIELD - Amateur TV Club Banquet. Ariston Restaurant. Scott Millick 217-532-3837 IN - EVANSVILLE - Winter Hamfest. Vanderburgh Co. Fairgrounds. Neil 812-479-5741

NOVEMBER 30-DECEMBER 1

NJ - PENNSAUKEN - Computer Show. South Jersey Expo Center. KQP Productions 908-297-2526 OH - COLUMBUS - Computer Fair. Veterans Memorial. Mark Hanslip 513-263-3378

DECEMBER 1996

DECEMBER 1

CA - LIVERMORE - Swapmeet. Las Positas College.

Noel Anklam 510-447-3857

KY - ERLANGER - Computer Show. Holiday Inn,
Cincinnati Airport. 1-800-672-3472
MI - FLINT - Computer Show. Holiday Inn, Gateway
Centre. Five Star Productions. 810-230-0988

DECEMBER 6-7

AR - HOT SPRINGS - Arkansas DX Convention. Dave Jacques 501-760-1181

CA - SAN FRANCISCO - Computer Show, Cow Palace, Robert Austin. 1-800-346-0100 CA - SANTEE - ARC of El Cajon Ham, Computer & Electronic Swapmeet. Santee Drive-in. 619-561-0052 In - InDIANAPOLIS - AGI Computer Fair. Indianapolis Events Center. 317-299-8827 LA - MINDEN - ARRL Hamfest, Minden ARA, Jimmy

DECEMBER 7-8

CA - SAN FRANCISCO - Robert Austin Computer Show. Cow Palace. 1-800-346-0100

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EXPER

elcome! This month, we'll take a look at the recent developments and price reductions of diode lasers.

This is an Intermediate Level project. Years ago, one milliwatt helium-neon lasers were the only type within the experimenter's budget, being about \$100-\$200 for a system, depending upon what type you chose to buy. And before that era, they were only available for use in a laboratory. Nowadays, we have five milliwatt diode lasers that operate in the visible spectrum which can be purchased for \$33.00 right out of the Digi-Key catalog {1}, and for less from other suppliers. Digi-Key offers lasers at 670 nM, which is very visible, and a ruby-red color rather than the orange-red of a He-Ne. They also sell infrared laser diodes that operate at 780 nM, which is barely visible. In addition, they are carrying visible laser diode modules which have their own drive circuitry inside. A couple of these are quite pricey, ranging from \$75.00 to \$200.00 in single quantities.

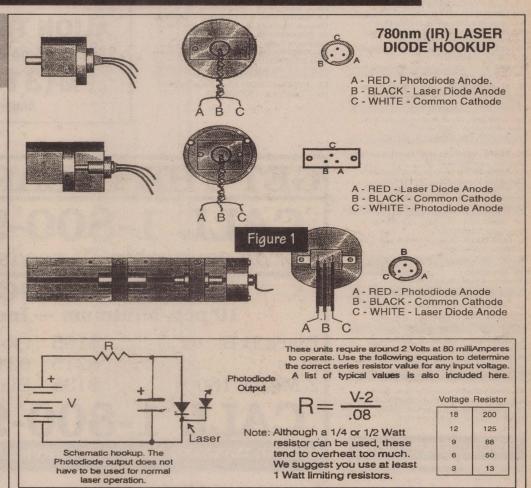
Diode lasers are different from He-Nes and other lasers because of their power requirements and beam profile characteristics. Sans a lens, a diode laser beam spreads out rapidly, losing power at a short distance. This is called ASTIGMATISM. The neat thing about a diode laser is its ability to be powered by just a couple of 1.5 volt batteries. However, I prefer to use a nine-volt battery or regulated power supply to allow more of a "cushion" in the circuit. Nevertheless, there are plenty of laser pointers and modules on the market which use two small 1.5 volt, common batteries. They have integrated circuit regulated power supplies inside that make the adjustment automatically for the lower voltages without potential for destruction of the laser diode.

Speaking of destruction, a laser diode is VERY sensitive to stray static electricity charges. If you work with these, you must take proper measures to eliminate static. One is a wrist strap grounded through a one megohm resistor to a utility (earth) ground. This can be homemade, such as from aluminum foil you place around your wrist. Be sure and use the resistor and keep away from directly grounded objects like if you have such an arrangement with your metal worktable. Since, otherwise, if your soldering iron or power supply developed an internal short, you could be electrocuted if you were connected between the faulty equipment and an earth ground. But this applies to all electronic equipment.

Next comes the soldering iron. It, too, must be grounded. In this case, you can either purchase an already grounded iron, or add a ground wire that runs from the iron to a utility ground. If you intend to work

on a metal worktable, that should be grounded too. Your diode(s) want to be kept in their protective static-proof containers until you're ready to assemble and have taken the above precautions. If you think I'm being fussy, consider this true tale: A fellow told me about walking across a rug, touching a then \$100.00 diode, and blowing it out! Once the diode is installed in your circuit, it will be far less

[1] I am not affiliated with Digi-Key, nor any other company mentioned here, but often mention D-K in articles because of their excellent selection of products and quick service. Request their free catalog, and you'll never regret it.

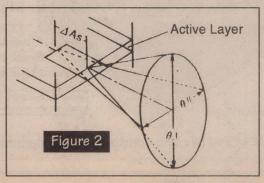


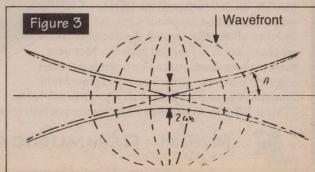
likely to be bothered by static. Keep it away from high voltage, though. Use $0.1~\mathrm{uF}$ bypass capacitors on the circuit board across the power supply rails and you should have no problems.

Soft Start Circuit

Diode lasers are also subject to voltage SPIKES which can ruin them. Use a 47 to 100 uF capacitor across the power supply rails so that the laser can start running without the possibility of voltage spikes. Figure shows one simple circuit to satisfactorily drive a laser diode without incorporating the feedback internal diode in the circuit. This is courtesy of MWK Industries. They sell several laser diode types and are listed in the Sources box. I have modified the circuit minorly by adding a capacitor (value is not specified; use your own judgement and available parts) across the power supply rails as shown, thus increasing the survival percentage of the laser diodes. Several common 780 nM diode configurations are shown; these are typical of what one finds in CD players, but the circuit also applies to visible

Should you want a more sophisticated set-up, which will cost you more, contact Meredith Instruments who appear in Sources. It is interesting to see what





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Table 1 Some Visible Laser Diode Wavelengths

Wavelength in nanometers	s Color
633	Orange-red
635	Orange-red
640	More reddish
650	Red
655	Red
670	Quite red
680	Darker red
685	Darker red
690	Ruby red
750	Dark ruby red
780	Barely visible
785	Barely visible
790	Practically invisible
800	Invisible

Note: There are many, many types of laser diodes available. From a few milliwatts to 150,000 watts (on a pulsed basis). Most of these emit in the infrared spectrum, which doesn't concern us at the moment. To list them all would take two or three columns!

diode manufacturers and vendors offer. Much is now pre-assembled using surface mount technology. This has the advantage of short lead lengths and using an advanced technology which combine to lessen the potential damage to diodes, increase efficiency, and lessen power consumption.

If you want a first-class experimenter kit, you may

want to contact LNS Technologies, who is also in Sources. While their kit isn't the cheapest, the superb manual and Heathkit style of their engineering makes the assembly and usage a pleasure. Plus, they offer an inexpensive repair service if you have a problem.

Optical Astigmatism

Figure 2 shows the astigmatism of a laser diode and its beam geometry. Lensing is essential for a tight beam. Diode lasers use what is called an ASPHERIC lens to achieve tight (or any) collimation. That is a lens which combines several lens geometry configurations in one unit. Aspheric lenses may be either plastic or glass. The plastic ones are cheaper, of course. Lenses are available from several dealers in the Sources list. Investigate who has what, and their current availability.

Note the geometry of the beam pattern as depicted in Figure 3. The BEAM WAIST is also different from He-Ne and other gas lasers. Clearly, we have a peculiar profile as

compared to a He-Ne, hence the need for aspheric lensing.

Laser Diode Structure

Figure 4 shows the basic structure of a laser diode. Note the interspersed layers of different

semiconductors. It is fascinating that the precise arrangement of these materials provide internal mirrors, where the laser beam is amplified and bounced from mirror-to-mirror. After many bounces, which amplifies the weak laser light, output eventually becomes a usable laser beam. Note the tiny size of the diode laser chip; uM is an abbreviation for one-millionth of a meter, so these are pretty small.

Figures 5 and 6 show different structures of the layered semiconductor material that make up two different diode lasers. Figure 6 is particularly interesting because it shows convergence of three beams into one. This is not unusual, for the more powerful lasers have several beams that converge into one single beam.

Integrated Circuit Driver ICs

There are several IC drivers available on the market. Most I've seen are made by Sharp. Surplus dealers often carry these along with a little resistor and capacitor kit so you can assemble your own power supply. Perhaps you'd like to explore the possibility, and it would make a great class project. We'd like to get into these in a future column. If you like the idea, drop us a note at the Laser Experimenter's P.O. box listed elsewhere in this column. Well, that's it for now. We shall see you all next time! NV

Closing Quote:

"I am here to support the assertion that light of every kind is itself an electrical phenomenon — the light of the sun, the light of a candle, the light of a glowworm." — Heinrich Rudolph Hertz

German physicist (1857-1894)

SOURCES

MWK Industries, 1269 West Pomona, Unit 112. Corona, CA 91720; ph. (909) 278-0563, FAX: (909) 278-4887. Hours: 8-5 PST weekdays, 10-3 Saturdays. Free catalog. Source for new and used laser equipment, supplies, and optics. Carries a wide variety of He-Ne, some laser diodes, and other lasers suitable for the experimenter hobbyist and professional.

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Midwest Laser Products, P.O. Box 262, Frankfort, 60423; ph: (815) 464-0085, FAX: (815) 464-0767. Free catalog. Carries new and used laser equipment and inexpensive start-up systems. Also currently carrying ruby laser, army tank rangefinder units.

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Society for Amateur Scientists, 4951 D. Clairemont Square, Suite 179, San Diego, CA 91777; ph/FAX: (619) 239-8807. Internet: scarlson@sas.org Organization specifically started for gathering amateur scientists of all kinds together. Not a hobbyist organization unless you're serious

The Electronic Goldmine, P.O. Box 5408, Scottsdale, AZ 85261. Free catalog. Ph. (602) 451-7454, FAX: (602) 451-9495. SMC components and a wealth of high-tech experimenter electronics, components, and project kits. Check out their new, expanded catalog of bargains!

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Amateur TV Quarterly, 1545 Lee Street, Des Plaines, IL 60018. Voice: (708) 298-2269, FAX: (708) 803-8994. Publisher of a magazine and several books geared to amateur TV. Mentioned here because of the general interest to experimenters and the parallel to our laser video endeavor. Contact them for the current pricing of a sample issue and brochure.

Small Parts, Inc., 13980 N.W. 58th Court, P.O. Box 4650, Miami, FL 33014-0650. A valuable resource for construction materials, tools, and hard-to-find items such as telescoping tubing, sheet metal, metric hardware etc. Free catalog. Catalog requests: (305) 558-1255.

LNS Technologies, 20993 Foothill Blvd., Suite 307 Hayward, CA 94541 Inquiries: (510) 886-9296, orders: (800) 886-7150. Visible laser diodes and a driver kit, also carries a line of high-technology electronic kits. Accepts major credit cards. Catalog \$1.00.

Marlin P. Jones & Associates, Inc., P.O. Box 12685, Lake Park, FL 33403-0685; ph. (407) 848-8236, FAX: (800)-4 FAX YES. Carries some surplus He-Ne and diode lasers, power supply kits, and surplus optics. High voltage wire. Many interesting electronic items both new and surplus. Free catalog.

Geo-Monitor, 65 Washington Street, #400, Santa Clara, CA 95050; ph. (408) 749-6770, FAX: (408) 249-3780. Publishes a technical newsletter about earth science, and may be useful to those interested in using lasers to record seismic activity.

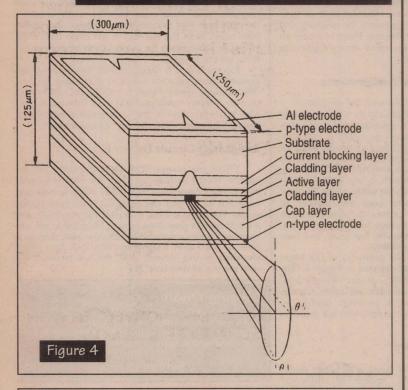
Edmund Scientific Company, 101 East Gloucester Pike, Barrington, NJ 00807-1380; ph: (609) 573-6250, FAX: (609) 573-6295. Free catalog. Supplier of a wide variety of educational and scientific equipment including lasers and an extremely comprehensive line of optics.

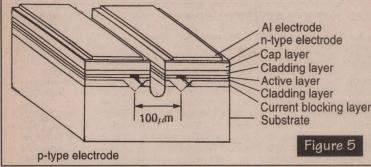
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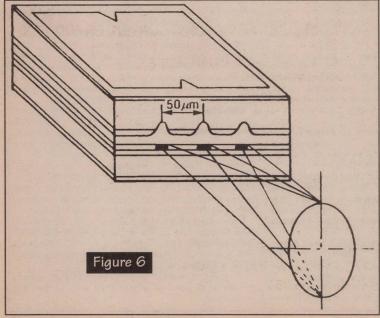
Be sure and mention Nuts & Volts when you contact the above suppliers so your request goes to the right department.

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Continued from page 93

New Products

TD-5 module has several new features, including crystal control for precise tone accuracy and a convenient DIP switch for tone selection without test equipment. General information in the manual helps you install the module in any transceiver. It is available either as a kit for only \$39.00 or factory wired and tested for only \$59.00.

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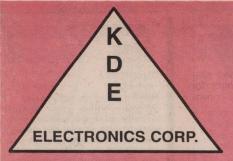
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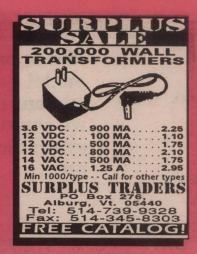
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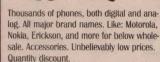
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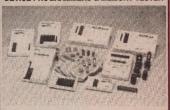
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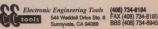
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MICROWAVE EXCITEMENT AT 10,000 MHz

by Gordon West

he amateur radio 10,000 MHz microwave band is called "X-BAND." It's also known as "10 giga-hertz" (GHz), actually pronounced 10 "jig-a-hertz." Look it up in the dictionary. The first letter "G" is soft like the letter "J!"

The amateur X-band extends from 10,000 to 10,500 MHz. This is an incredible 500 MHz of spectrum that we need to use, or quite possibly LOSE. You see, every now and then, Federal and private radio spectrum prospectors will tune up certain areas of VHF, UHF, and microwave frequencies and take a spectrum analyzer "snapshot" to prove band occupancy or radio band dis-use. If the X-band is grossly under-used by amateur radio operators, it's relatively easy for the private sector to call this to the attention of the Federal Communications Commission and come up with a plan to take over the ham-unused band and sell it off making gazillions of dollars for both the Feds, as well as privateers looking for more frequencies. And between you and me, I hope they don't take a snapshot today because the ham activity on 10 GHz isn't all ate on the amateur radio X-band, with no licensing and operating restrictions other than a valid ham ticket, with or without code, Technician level or higher. For you non-technical hams, you could pick up the Tech ticket in about three days by studying my no-code-plus book available at all Radio Shack stores.

Ten thousand megahertz (10,000 MHz) is not unreasonably high for quality, medium-range and sometimes long-range microwave communications.

Wide-band FM voice
Narrow-band SSB voice
Spread-spectrum voice
Amateur wide-band television
Amateur television repeaters
FM and SSB repeaters
CW propagation beacons
Weak signal SSB and CW tropospheric ducting

At 10,000 MHz, well above the marine radar bands and well above most home microwave oven frequencies, two-way SSB contacts well beyond 1,000 miles have been established. In August of

1994, armchair copy communications on 10,000 MHz were conducted between XE2/N6XQ in Baja, CA up to WB6CWN way up near the San

tain-top stations, and up to 400 miles and more during X-band contests when participating hams bring out their big microwave dish antennas and go for the best range possible from the very best location.

Up on 10 GHz, location is everything. Since we are transmitting on microwave frequencies which are identical to X-band radars, things in the way will block great range capabilities. Two X-band stations transmitting a fraction of a watt into modest horn antennas might only work a couple of miles with homes and trees in between their microwave path. If both stations went to their roof tops, range could expand out to 10 or 20 miles. If one station went to the top of a hill, and another station went to the roof of an office building, range could expand out to 50 miles. If both stations went to big mountains, X-band communications could conceivably span to the line-of-sight distance between both mountains in relatively calm, warm air.

A good way to approximate microwave range in miles is to take the square root of the building or mountain height in feet, and the square root of the other station's height in feet, and add these two numbers together for your estimated communications range at 10,000 MHz. For instance, a microwave operator running equipment atop a 100-foot building would give you a line-of-sight-to-the-horizon range of 10 miles. This is calculated by the square root of 100 feet of height equaling 10. If the other microwave station has driven up to the top of a 5,000 mountain peak, their line-of-sight-

Small, but power-

ful 10,000 MHz

horn antenna.

range would be about 70 miles. Now add 70 miles plus 10 miles, and 80 miles is your minimum microwave communications range. Since atmospheric refraction adds about 20 percent additional range, you could probably stay in touch out to about 100 miles separated. During periods of tropospheric ducting where microwave signals get caught up in an upper inversion layer, the range could be as much as 200 or more miles. Maybe even 300 miles over Maybe more.

You are probably beginning to wonder why I'm not talking much about microwave power levels. The reason I don't talk about them much is because the power levels aren't much! One hundred milli-

watts (0.1 watt) is quite common at these frequencies. One-tenth of a watt, on a line-of-sight basis, could carry well over several hundred miles with the right type of antenna and a good amount of antenna gain. More about antennas later.

One hundred milliwatts is the operating level of most start-up microwave systems that are pur-





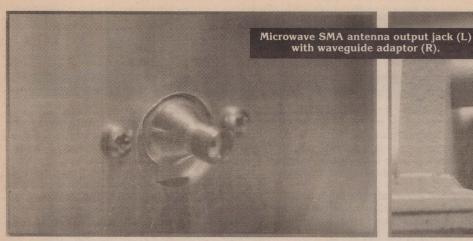
spanning over 1,124 miles — earning these two microwave ham operators a place in the X-band record book.

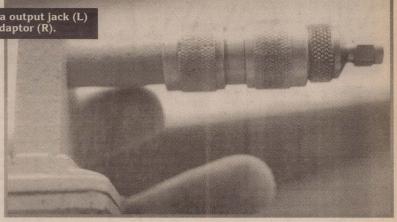
Francisco Bay area,

operators a place in the X-band record book. Typical communications range on X-band is about 50 miles for low-level stations, 100 miles for moun-

that much.

Hopefully this article will help. It takes only a no-code Technician class license or higher to oper-





chased over-the-shelf, ready-to-go, or put together in modules. Some microwave home brewers are working down at the 10 milliwatt levels, using automatic door opener gunnplexers within the heart of their X-band system. Ten milliwatts will go just about as far as 100 milliwatts on a line-of-sight

Where higher power counts is when trying to jam a signal through trees, clouds, rain, and taking a bounce off of a nearby mountain or building. Power levels approaching one watt work nicely in these applications. But to obtain one watt output at 10,000 MHz, relatively expensive power amplifiers called "traveling wave tubes" may need to be employed.

Power levels at 100 milliwatts are readily available in 10 GHz pre-assembled or kit systems. For over 15 years, a fully assembled, tuned, and tested 10 GHz FM transceiver is offered by Advanced Receiver Research (P.O. Box 1242, Burlington, CT 06013; 203-582-9409). You can order the equipment with either 10 milliwatts or higher output levels, and this is plenty to get you on the air.

The ARR transceiver comes with its own plastic horn antenna; VXO for sliding up and down the band; and full-duplex, crystal-clear, wide-band FM communications to link up with other stations using ARR transceivers or homebrew gear.

All you need to do is add 12 volts and a simple crystal microphone, and you are set for action. The unit is ready to play, out of the box, and I have personally communicated well over 250 miles on FM at X-band using one of these low-power ARR trans-

Working another station for your first time on 10 GHz will require pre-planning. It's not that you can turn your set on, call a CQ, and instantly make a contact. Just won't happen. Even if you live on a hill. You need to set up a schedule, an approximate frequency, and precise antenna pointing in order to link up. Since we are up at microwave X-band frequencies, the antenna must be of the same polarization and aimed at the other station with which you are attempting to link up. Within 10 miles, bounces do occur. But beyond 10 miles, you need to take steady aim to lock onto the other station. The voltage-controlled oscillator (VXO) is similar to a "VFO," and your automatic frequency control is normally in the off position as you are tuning up and down the band ready to make contact with another station in the direction of your horn antenna. After a little bit of playing, you'll get the hang of tuning with one hand, swishing the horn back and forth with your other hand, and keeping an ear out for the FM carrier of the other station. Once you hear their carrier, they will hear your carrier, and both of you will be able to talk full duplex because your transmit and receive are separated by your 30 MHz intermediate frequency.

Once you begin to understand each other over the air, maneuver your antenna slightly left and

right, and zero in for the best signal. If you are more than 10 miles away, the signal will build and fade as clouds and warm/cool air masses refract and reflect your signals.

The Advanced Receiver Research 10 GHz amateur transceiver is a fun way to get started on microwaves. But this unit is not going to lead you to any DX record because wide-band FM is simply too wide and too noisy for microwave weak-signal work. When you have worked at least 20 other FM X-band stations, time to graduate up to single sideband and CW.

Ten gigahertz (10 GHz) SSB transceivers are not available "over the counter." But they do come pre-assembled into modules that you can pull together in just a few days.

SSB Electronics

124 Cherrywood Drive Mountaintop, PA 18707 717-868-5643 (Ask for Jerry.)

Down East Microwave

P.O. Box 2310, RR 1 Troy, ME 04987 207-948-3741 (Tell them Gordo sent you.)

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These three companies offer 10 GHz X-band transverters. The transverter must be driven by a multi-mode, two-meter or 10-meter transceiver. The SSB Electronics transverter works nicely with almost any type of two-meter, multi-mode transceiver when the two-meter set is run in the lowpower position. Transverters cannot run with more than one watt input, and SSB Electronics has a special circuit which keeps your two-meter SSB mobile or base station from overpowering the input.

The transverter will take 10,000 MHz microwave signals and downconvert them to 144 MHz or 28 MHz. When you go into transmit, the transverter up-converts your 28 or 144 MHz signal to the desired operating frequency on X-band. The SSB Electronics system includes pre-assembled modules that are "plumbed" with tiny RF rigid copper coax sections, fitted with SMA connectors to keep losses down. The antenna switch-over relay is designed for microwave frequencies with negligible loss on X-band. You can write or call SSB Electronics for a detailed look at the required modules for X-band and pictorial layouts on how the system can go together as a weekend project once you have "plumbed" all of the pre-assembled pieces together. The modules are from Europe, but each one is tuned and tested by Jerry at SSB Electronics.

The output of the X-band transverter terminates to a gold-plated SMA receptacle, and here is where your antenna system begins. On microwaves, coaxial cable only a few feet in length can quickly attenuate weak signals. Ten gigahertz (10 GHZ) waveguide is the answer, and SMA-towaveguide adapters are easily seen for a few bucks each at ham swap meets. They are also seen, brand new, for over \$100.00 in catalogs! I suggest you try the swap meets first!

What you are looking for is a "WR-70" transition which may accept the SMA or end-connector and transition to 7,000 MHz-12,000 MHz waveguide. I prefer a short section of flexible waveguide between the output of my X-band transverter to the horn or dish antenna I plan to use. A four-foot section of flexible military surplus waveguide can be seen selling at hamfests for under \$40.00, but if you price it out new expect to pay over \$100.00. Again, shop the swap meets.

Your antenna system at X-band may be a simple horn to get you started. Small plastic horns with a radio-reflective metallic coating are an inexpensive way to beam your signal out and receive other signals up to about 50 miles away. The horn is rather broad in its transmission and reception, and 10 or 15 degrees either side of "dead on" will still lead to reasonable communications. But for the more serious DX operator, a two-foot or four-foot dish is preferred.

The two-foot and four-foot parabolic dish antennas have a variety of opportunities to feed them. At my 10 GHz station, I feed from the back of the dish through cylindrical waveguide to a splash plate, and the round plate in front of the dish captures full illumination of the entire dish. Other operators may mount the transmit portion and receive section in front of the dish and have nothing behind. There are also offset feed systems, the popular "penny feed" system, and the opportunities are endless if you enjoy working with parabolic reflectors. Many times the circular waveguide is made from common plumbing hardware.

The two-foot dish at X-band frequencies requires you to be within five degrees of the distant X-band signal when relative signal strength is low. Four-foot and six-foot dishes require you to be within one degree of precise aiming. When signals are strong, multiple reflections make aiming relatively simple. Swing the dish through weak peaks and valleys; and when you get to one extremely strong peak, that's probably the true direction of the incoming signal. Maybe it's a bounce - maybe it's the real thing - but this is the way you want to maintain communications. Sometimes a bounce is better than what you thought was a direct shot.

X-band contacts are usually coordinated over the two-meter band. In Long Island, NY, there are regular VHF nets to encourage X-band equipment construction and X-band operation. It has been found that handheld signals on two meters approximate the results you are going to obtain at 10,000 MHz. If you can communicate on two meters with low-power handhelds, you can certainly make the grade on 10 GHz X-band. However, if you and the other station can't hear each other on two meters handheld, chances are you won't be able to make the X-band path work. I have never, never, NEVER found that I could establish communications on 10 GHz where I could NOT on a pair of two-meter handhelds. Nor have I ever found crystal-clear two-meter handheld communications that I couldn't easily do on 10 GHz.

Many microwave experts have put up X-band beacon stations specifically to attract newcomers to 10 GHz, and to assist newcomers in perfecting their antenna and receiver systems by having a reference signal on the air, 24 hours a day. The frequency 10,368.185 normally downconverts to 144.028 to begin your search: If your local X-band beacon is slightly higher in frequency, tune higher on your two-meter, multi-mode transceiver. If it's lower, tune lower. Most beacons run horizontal polarization. Most usually operate CW, and continuous beacon reception is easily possible up to 100 miles distance. Continuously transmitting beacons will allow you to fine-tune your system on receive; and when you have increased reception capabilities through antenna improvements, transmit range will also increase, too.

How Much Will It Cost?

The Advanced Receiver Research 10 GHz completely assembled set-up, Model TR10GA, running 10 milliwatts output, runs about \$500.00. Add 12 volts, plus an external speaker or headphones, provide a simple Radio Shack high-impedance mike, and you are on the air, full duplex. There are plenty of other operators out there on 10 GHz wide-

band FM, so you'll have plenty of folks to talk to.

If you're putting together the transverter system, expect to pay about \$1,000.00 for the necessary pre-assembled modules:

XLO-1 local oscillator module XRM-1 10.368 GHz receive downconverter XTM-1 10.368 GHz transmit upconverter HP8761 single-pole, double-throw RF switch SMA to 10 GHz waveguide adaptor Control relay Output TR relay

You will also need to score a multi-mode twometer or transmeter radio to act as the transmitter/receiver system that is necessary for the transverter. You can sometimes find used multi-mode equipment for as little as \$300-\$500. Even inexpensive 10-meter SSB sets converted from the CB radio service could work as your transmitter/receiver. Be sure to specify what frequency your system is running on, in order to obtain the right transverter modules.

Suitable waveguide will be "WG16," 29.8 mm by 39.4 mm, flexible, about five feet long, under \$100.00.

You can build your own dish antenna and save big bucks, or spend big bucks like \$500.00 for a perfect spun-aluminum 10 GHz dish. I would rather build my own, and the \$30.00 VHF/UHF manual from the Radio Society of Great Britain, available at all ham radio stores, offers multiple pages of microwave homebrew projects. The RSGB VHF/UHF manual, now in its Fourth Edition, is considered the \$30.00 "bible" for homebrew enthusiasts at microwave frequencies.

The American Radio Relay League sponsors 10 GHz contests on August 17 and 18, as well as on September 21 and 22. These weekends bring out more 10 GHz activity than any other time of

year. You should plan ahead and try to uncover which two-meter frequency will be used for coordination. Almost all coordination takes place on two meters. Ask on your local repeater where the 10 GHz X-band groups hang out. Each area has their own two-meter favorite frequency, almost always simplex. Remember, if you can easily communicate between you and the other station on two-meter simplex with a handheld, this is easy X-band communications range.

So give X-band a try. It is by no means the very highest microwave ham band — there are five others even higher — and ARR even builds a 24 GHz system once the bug strikes! And every time you see a small dish on a building top, chances are it's a commercial microwave system operating well up to 100 GHz. And if you still want to operate higher, read over the laser communications column in this same magazine and see what happens when microwaves turn into lightwaves.

Enjoy X-band with your amateur radio license.

Some sources for 10 GHz and microwave building are:

Holzman Electronics 108 Sawtell Ave. P.O. Box 2421 Brockton, MA 02405-2421

> Danielson Sales 508 South N St. Indianola, IA 50125

K-Mec 4862 Montgomery Rd. Ellicott City, MD 21043

TECH FORUM

Continued from page 65

In practice, you may actually find it easier to do what I did: use a simple analog ramp generator and comparator. Let me refer you to my favorite IC, the LM3900 quad current-difference amplifier. Get the data sheet, or consult a good cookbook.

It's easy, with just a few components, to build a ramp generator that ramps from nearly O VDC to nearly VCC with very good output linearity. The key is to have ramp reset every time the AC power line voltage crosses zero; again, you can use one section of the LM3900 for the zero-crossing detector. Now each time the power line sinewave goes from 0 to 180 degrees, your ramp will go from minimum to maximum. Use another section of the LM3900 to amplify (and, if necessary, invert) the DC signal from your sensor.

Finally, a section of the IC serves as a comparator. When the [amplified] sensor voltage equals the ramp voltage, the comparator output will turn on an opto-isolator. The output of the opto triggers your thyristor circuit and turns on the load. As the sensor calls for more power, the comparator will switch earlier in the power cycle, and more power will be delivered to the load. Purists will point out that a graph of power delivered, relative to turn-on time (in degrees past zero-crossing) is not linear, while the ramp is linear. Therefore, the control curve of sensor voltage, relative to power delivered, is NOT a straight line. That's true. BUT many sensors don't have linear output in the first place. And, in any case, the system will eventually reach equilibrium, settling down to the amount of output power you need to maintain the sensor at the desired setpoint.

At first glance, this circuit may not be seem glamorous, but it has advantages. The chip count is lower than with a pure digital circuit. There's no code to develop, as with a micro-controller approach. Adjustments are as near as your nearest trimpot. And finally, it will work with a minimum of debugging.

Greg Miller State College, PA

ANSWER TO #8966 - AUGUST 1996

Seems like someone is always asking how to change apples into oranges or make do with something through extensive modifications. I won't say it's not possible to modify a two meter rig for aviation band, but there is a big difference besides frequency, namely that aviation bands are AM, while two meter is FM. The answer to this question, besides changing crystals, would be to re-tune the front end and output stages, possibly requiring component changes. It's a big jump. But even if that were accomplished, converting from FM to AM is not practical, especially on a nice, compact solid-state rig. Then there's the issue of licensing how small is this foreign country, and would such a non-approved transceiver

> Richard M. Nelson Newport News, VA



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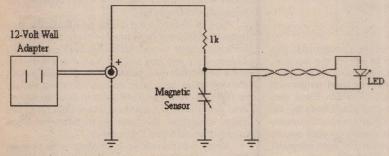
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distributors@nutsvolts.com URL — http://www.nutsvolts.com systems. Basically, it's a magnet/reed relay that's closes when the door is closed (you can get switches that open when the door shuts, but at higher cost). You can buy them at most hardware stores and any Radio Shack for about \$5.00. You'll also need a power source. I suggest a 12-volt AC-to-DC wall adapter, like those used for wireless phones and Ninetindo (Radio Shack has a good selection of them). The fact that you have a voltage dropping resistor makes this an easy fix. Here's how it looks:



Of course, you'll have to position the magnetic switch so that when the garage door is fully closed, the reed relay closes, too, and shunts the current from the LED to ground. You may need to adjust the value of your dropping resistor for the new voltage (halve it), but probably not.

Screen Goes Black At Boot

Q. I have a 486-50 PC that quit booting up right. I replaced the back-up battery, but still nothing. Sometimes I get a screen for a brief period, but then it goes black. I've tried delete, etc. Nothing works. Are there any programs/ methods I can use that does not require the use of a monitor to get my PC

G. Hall Mineral, VA

A. I have two thoughts on the subject. My first take is that you've changed the video somehow, and it's Windows that's going blank, not your PC. What I'd do is insert a floppy DOS boot disk, any version beyond 5.0 will work, change to the WINDOWS directory, and type SETUP. Now choose VGA from the video offerings, remove the diskette and reboot. If you can't get a DOS prompt using this method, try another video card and/or monitor. Second, have you installed a new adapter lately? 486-50 motherboards are very finicky because of the 50-MHz bus speed. All too often they crash at the least change. My suggestion is to backtrack, removing one card at a time - power off, of course until the problem is cured. Been there, had to do that.

Keyboard Error

Q. I just upgraded my computer's motherboard and CPU and have run into an annoying problem. When I turn on the computer, I get a "keyboard error or keyboard not present" error message along with an alarm signal. The problem goes away if I unplug and then plug in the keyboard and hit reset. If I turn on the computer with the keyboard unplugged, I still get the error message, but not the alarm. I think it's a keyboard problem. My current configuration is a Rhino 9 motherboard with Award BIOS and a Cyrix 6x86 P120+ CPU. The keyboard is made by IBM in 1986 and says model M on the underside label.



D. Deutscher via Internet

A. It's been a long time since I've seen one of those keyboards. Chances are you've spilled more than one cup of coffee on it, and it's survived as well as can be expected. I believe it's now time to retire it. I say that because the fact that you've also upgraded to a 6x86 CPU tells me you've also upgraded to Windows 95, in which case you'll want a Windows 95 keyboard, like the Nova 9500 from Maxi Switch (602-746-9378).

But there is one more thing to check first. Check in your BIOS setup (press Ctrl-Del on power up) and see how the BIOS recognizes the keyboard and

responds to errors.

Needs Obsolete Timer

Q. The LM322N timer IC is no longer available. Is there a direct substitute? Ramon Eller Pittsburgh, PA

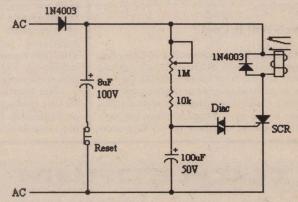
A. Nope. However, there are still some LM322N chips in the pipeline. I located three with my first phone call (415-457-0466); Digi-Key (800-344-4539) is out of stock, though. The LM122 is electrically identical. It comes in a metal can and will fit in your 14-pin DIP socket with some lead twisting. National Semiconductor recommends using the LM555 for new designs.

Diac Timer

Q. I've been using a UJT 4891 unijunction transistor to control an SCR in a timer circuit. Unfortunately, I can no longer find a source for them, and attempts at replacing it with a 2N6027 programmable unijunction transistor (PUT) have failed. Can you help?

Highland Park, IL

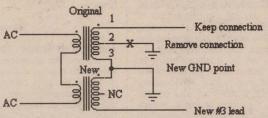
A. Unijunction transistors have been largely replaced with diacs because of their simplicity and low cost. You can buy them from Digi-Key (800-344-4539) for under a dollar. The semiconductor diac is a fullwave or bidirectional thyristor. It's triggered from the blocking state to the conduction state when the amplitude of the applied voltage exceeds the breakover rating of the diac (typically 30 volts). When combined with a capacitor, the diac can be used to trigger an SCR or triac. Here's a typical diac timer circuit that should work for your application.



400-Hz Surplus Operation

Q. Do you know of a circuit that will convert 60-Hz AC to 400-Hz AC, something that can be used with military surplus items that operate on 400 Hz? John Janeczko Haddon Heights, NJ

A. Here's a trick that we used use back in the 50s when W.W.II surplus was plentiful. Wire two 400-Hz power transformers in series, as shown below, and run them from 60 Hz. The catch is you need two identical power transformers, and you need to find a place to mount the new transformer in the surplus



BASIC Stamp Problem

Q. I'm using a Parallax BASIC Stamp 2 with a 4066 analog switch. The Stamp output drives the 4066 control voltage. The 4066 switches are wired in parallel with four keypad switches, so that I can control the keypad from the BS2. Three of the switches work fine, but the fourth doesn't switch reliably. I've checked the BS2 outputs and they're fine. If I apply five volts directly to the 4066 control line the switch works, but it won't always close under program control. Any ideas?

Andy Zaayenga via Internet

A. It sounds like the output pulse of the BS2 isn't wide enough for the 4066 to react. It could be slow rise time, too narrow a pulse width, or a bypass capacitor hung on that control line that's taking longer to charge than the rest. Check your program and increase the amount of time the output stays high, and see if that fixes the problem. If not, replace the 4066.

Reader's Tip

In the August issue, you sent someone to HP for a test instrument HP manual. My experience is that it's much cheaper to buy them from a used manual dealer. For example, I just bought an HP181 manual, for which HP wanted \$39.00, for \$15.00. Here's where to buy used manuals.

David DiGiacomo via Internet

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MAILBAG

Dear Mr. Byers:

Your comment to Kim White in Highland, Illinois about satellite pagers finding her anywhere in the world hasn't proven true for me in Madison, WI and two co-workers in Milwaukee. Our Skytel pagers would not work inside our buildings at all. A metal roof or concrete floors seem to stop the weak radio signal from the satellite. Pagers from local companies work much better.

Roland Peck Madison, WI

Dear Mr. Byers:

On page 118 of the August issue in the Electronic Q & A column, you gave the phone number for HP as 800-542-2844. I called the number and was informed by a gentleman that he was not HP and that he's not a happy camper because it costs him money for each wrong number. By the way, what is the right number?

John

Response:

I gleaned the document number and phone number from an HP Web site (they have more than one). A call to a live person at HP set the record straight. The correct numbers are 800-227-8164 if you have the document number and 916-783-0804 if you only have a model number. HP is in the process of correcting the error, and I assume an apology (at least) to Mr. Kane. I called both of the new numbers and verified them, as I usually do for most published numbers. From now on, I'll call them all!

TJ Byers Q & A Editor

Dear TJ:

I have a bone to pick with you on the subject of EDO SIMMs. In your Q & A column in the August '96 Nuts & Volts you state: "EDO memory sells for the same price as non-parity SIMMs, and can be used in all 72-pin non-parity systems." Well, according to my personal experience and my local computer guru, you are wrong - I was not able to use EDO SIMMs in my Acer 72-pin non-parity system. In the future, please refrain from making statements like this unless you've checked it out thoroughly for yourself.

Chuck Olson Washington Island, WI

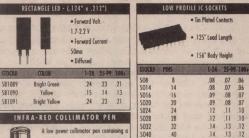
Response:

Actually, this answer came from an EDO supplier, so I took it at face value. Sorry your Acer didn't accept it. I don't know why. Repeated phone calls to Acer gave me no answers. It's possible that the 4 MB of built-in RAM is causing the problem, but I can't see how. I have tried it since, and EDO memory works in my old 486 motherboard. I wonder if other readers are experiencing similar problems? If so, please contact me.

TJ Byers Q & A Editor **NOKIA BAG**

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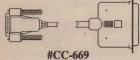
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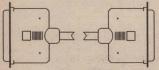
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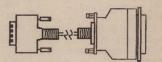
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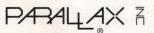


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